

# ELL212 - Tutorial 2 Solutions, Sem II 2015-16

- 1) Assume the voltage and current at a point  $z$  on the transmission line to be of the form

$$V(z, t) = (V_0^+ e^{-j\beta z} + V_0^- e^{-j\beta z}) e^{j\omega t}$$

$$I(z, t) = \left( \frac{V_0^+}{Z_0} e^{-j\beta z} - \frac{V_0^-}{Z_0} e^{-j\beta z} \right) e^{j\omega t}$$

The unknowns  $V_0^+$  and  $V_0^-$  can be evaluated using the boundary conditions at the two ends of the TL

$$V_s \exp(j\omega t) = V(0, t)$$

$$V(l, t) = R_L I(l, t)$$

Therefore

$$V_s = V_0^+ + V_0^-$$

$$V_0^- = V_0^+ \left( \frac{R_L - Z_0}{R_L + Z_0} \right) e^{-2j\beta l}$$

Also,

$$V_L = V(l, t)$$

Therefore,

$$V_L = \frac{R_L}{R_L \cos(\beta l) + j Z_0 \sin(\beta l)} V_s$$

(Note that at low frequencies,  $\beta l = \frac{\omega l}{c}$  is very small and thus  $\cos(\beta l) \approx 1$ ,  $\sin(\beta l) \approx 0$ ,  $V_L \approx \frac{R_L}{R_L + 0} V_s \approx V_s$  and the transmission line behaves like a short)

- 2) This circuit is equivalent to

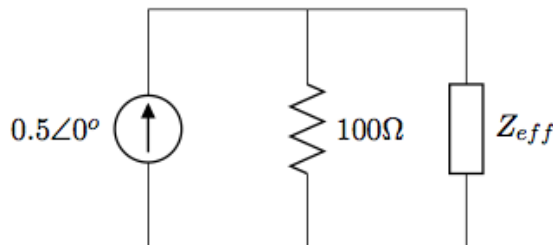


Fig. 1. Equivalent Circuit of Q2

$$Z_{eff} = 50 \left( \frac{25 + j50 \tan(360 \times 2.6)}{50 + j25 \tan(360 \times 2.6)} \right) = (22.3186 - j7.3812) \Omega$$

The current through the  $100 \Omega$  resistor ( $I_1$ ):

$$I_1 = 0.5 \left( \frac{22.3186 - j7.3812}{122.3186 - j7.3812} \right) A = (0.0927 - j0.02458) A$$

$$P_{R1} = |I_1|^2 \times 100 = 0.799 W$$

The voltage,  $V_s$  across the current source can be calculated as:

$$V_s = I_1 \times 100 = 9.27 - j2.458$$

$$P_s = \text{Re}[V_s I_s^*] = 9.27 \times 0.5 = 4.635 W \text{ and } P_{R2} = P_s - P_{R1} = 3.836 W$$

- 3) The equivalent circuit is shown in Fig. 2:

$Z_1$  is the equivalent impedance of the short circuit stub attached between A and ground.

Therefore

$$Z_1 = j Z_0 \tan(\beta d_1) = j 300 \tan\left(\frac{2\pi}{100} 10\right) = j 217.963$$

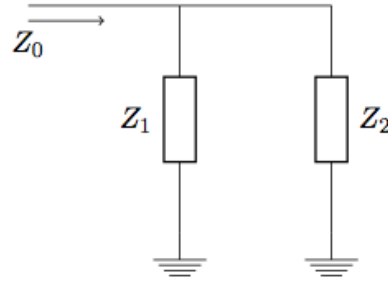


Fig. 2. Equivalent Circuit of Q3

Also

$$Z_0 = Z_1 || Z_2 = \frac{Z_1 Z_2}{Z_1 + Z_2}$$

$$Z_2 = \frac{Z_1 Z_0}{Z_1 - Z_0} = \frac{300 \times j217.963}{j217.98 - 300} = (103.64 - j142.658)\Omega$$

Since,

$$Z_2 = Z_0 \left( \frac{Z_L + jZ_0 \tan(\beta d)}{Z_0 + jZ_L \tan(\beta d)} \right)$$

$Z_L$  can be calculated as:

$$Z_L = Z_0 \left( \frac{Z_2 - jZ_0 \tan(\beta d)}{Z_0 - jZ_2 \tan(\beta d)} \right)$$

#### 4) Equivalent circuit

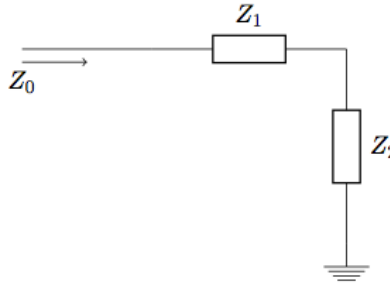


Fig. 3. Equivalent Circuit of Q4

$$Z_1 = jZ_0 \tan(\beta d_1)$$

$$Z_2 = Z_0 \left( \frac{Z_L + jZ_0 \tan(\beta d)}{Z_0 + jZ_L \tan(\beta d)} \right)$$

Thus

$$Z_0 = jZ_0 \tan(\beta d_1) + Z_0 \left( \frac{Z_L + jZ_0 \tan(\beta d)}{Z_0 + jZ_L \tan(\beta d)} \right)$$

Separating the real and imaginary parts of this equation:

$$1 = \frac{Z_L Z_0 (1 + \tan^2(\beta d))}{Z_0^2 + Z_L^2 \tan^2(\beta d)}$$

$$\tan(\beta d_1) = \frac{(Z_L^2 - Z_0^2) \tan(\beta d)}{Z_0^2 + Z_L^2 \tan^2(\beta d)}$$

which can be solved to obtain

$$\tan(\beta d) = \pm \sqrt{\frac{Z_0}{Z_L}}$$

$$\tan(\beta d_1) = \mp \left( \sqrt{\frac{Z_0}{Z_L}} - \sqrt{\frac{Z_L}{Z_0}} \right)$$