

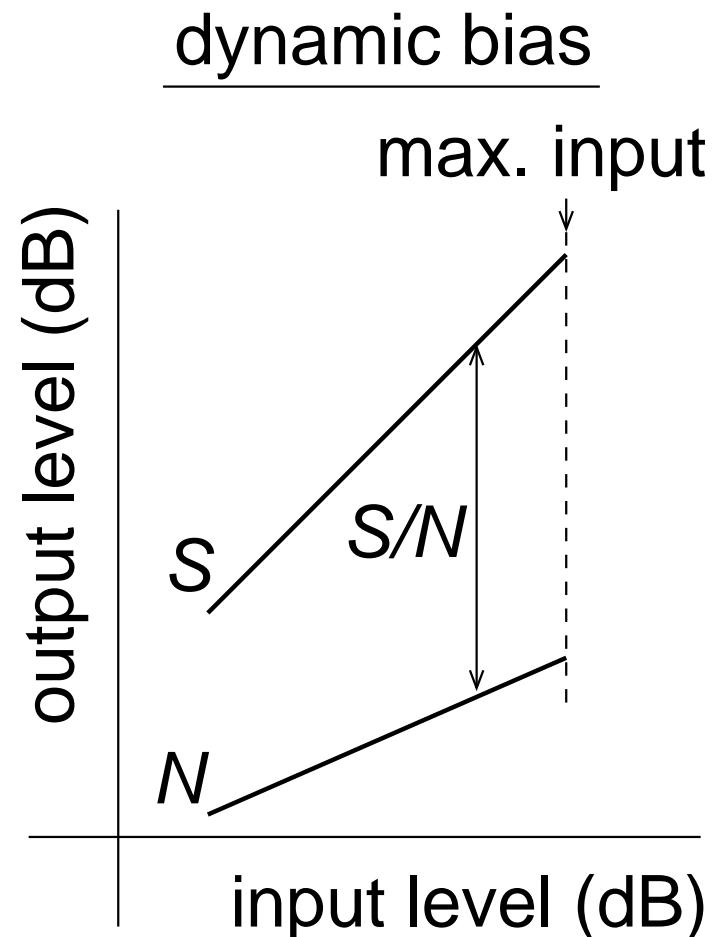
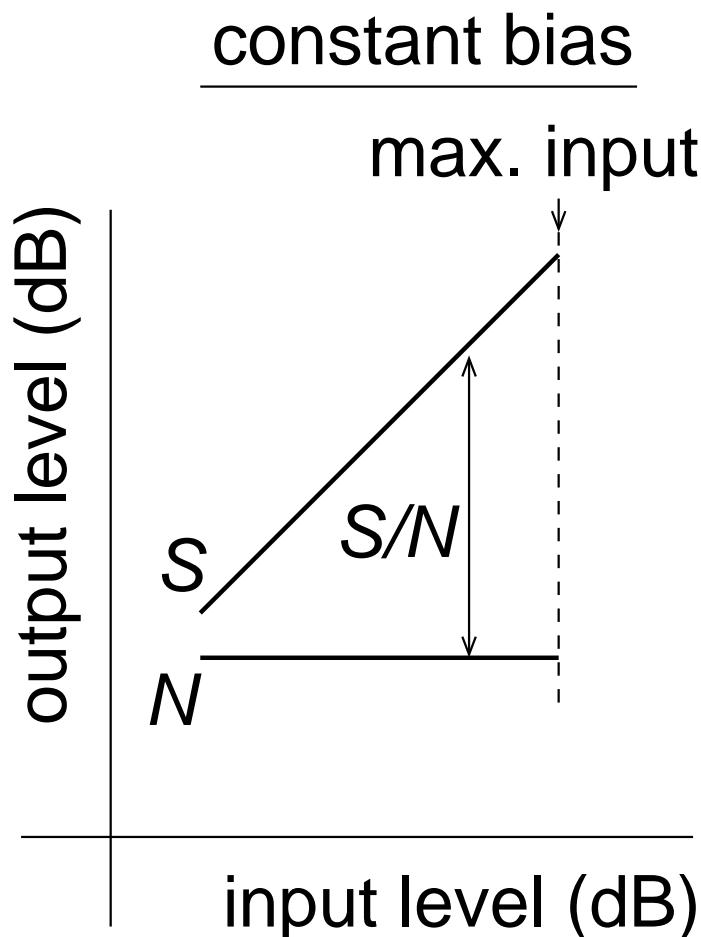
Dynamically Biased 1 MHz Low-pass Filter with 61 dB peak SNR and 112 dB Input Range

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Motivation

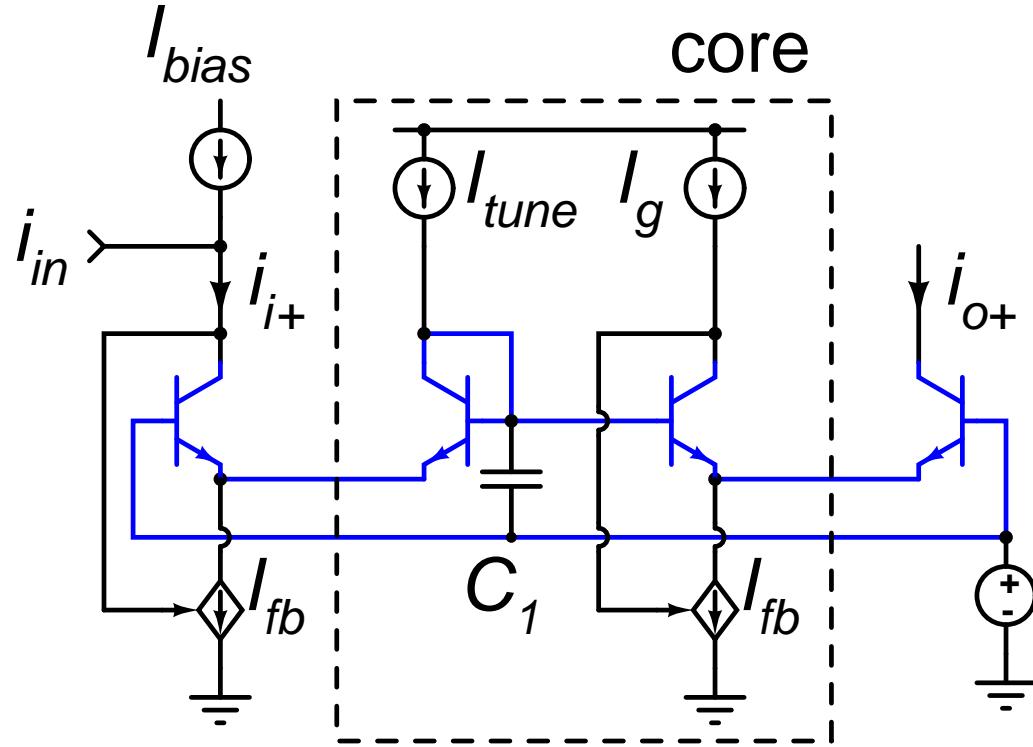


- Dynamically biased current mode filter:
 - 112 dB input range ($\text{THD} < -40 \text{ dB}$).
 - Quiescent power consumption of $575 \mu\text{W}$.

Outline

- Principle of distortionless dynamic biasing.
- Third-order Butterworth filter.
- Measurement results.
- Comparison.
- Conclusions.

First order log-domain filter



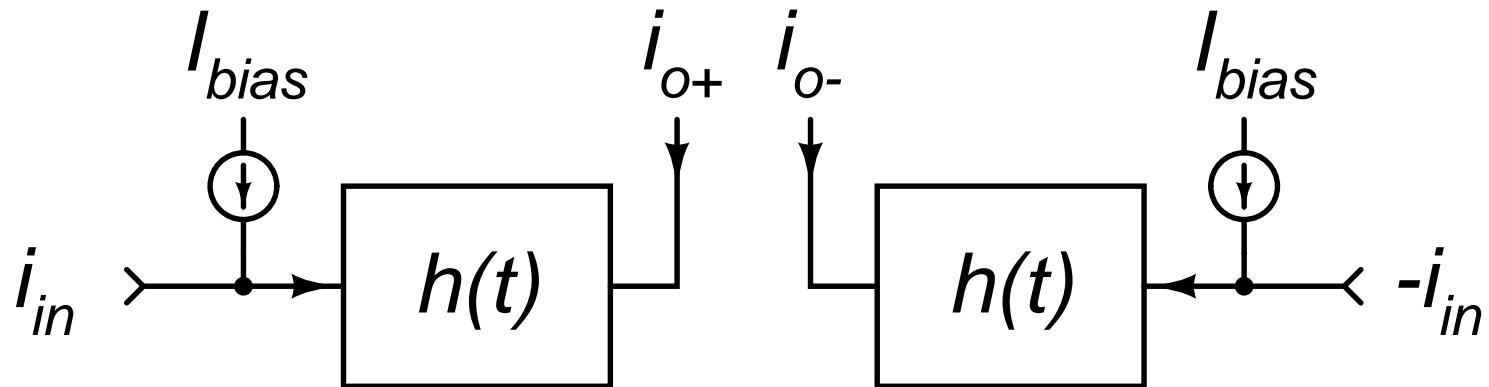
- Linear from i_{i+} to i_{o+}

$$\frac{I_{o+}(s)}{I_{i+}(s)} = \frac{I_g / I_{tune}}{1 + sC_1 V_t / I_{tune}}$$

[Punzenberger & Enz '97]

- Output noise $\sim I_{bias} + kI_{bias}^2$; Power diss. $\sim I_{bias}$.
- $I_{bias} \downarrow \Rightarrow$ Output noise \downarrow & Power dissipation \downarrow .
- But, filtered $I_{bias}(t)$ appears at the output.

Pseudo-differential operation

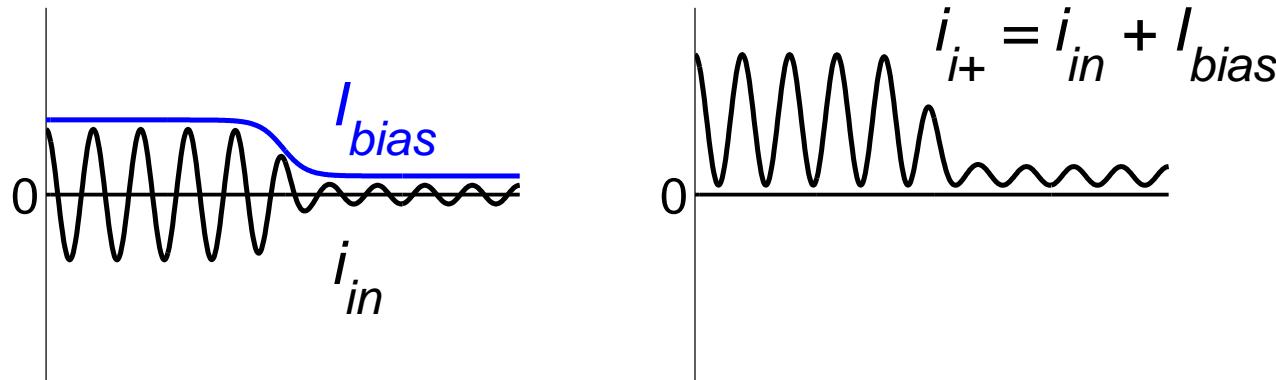


$$i_{o+} = (i_{in} + I_{bias}(t)) * h(t) \quad i_{o-} = (-i_{in} + I_{bias}(t)) * h(t)$$

$$i_{out} = i_{o+} - i_{o-} = 2i_{in} * h(t)$$

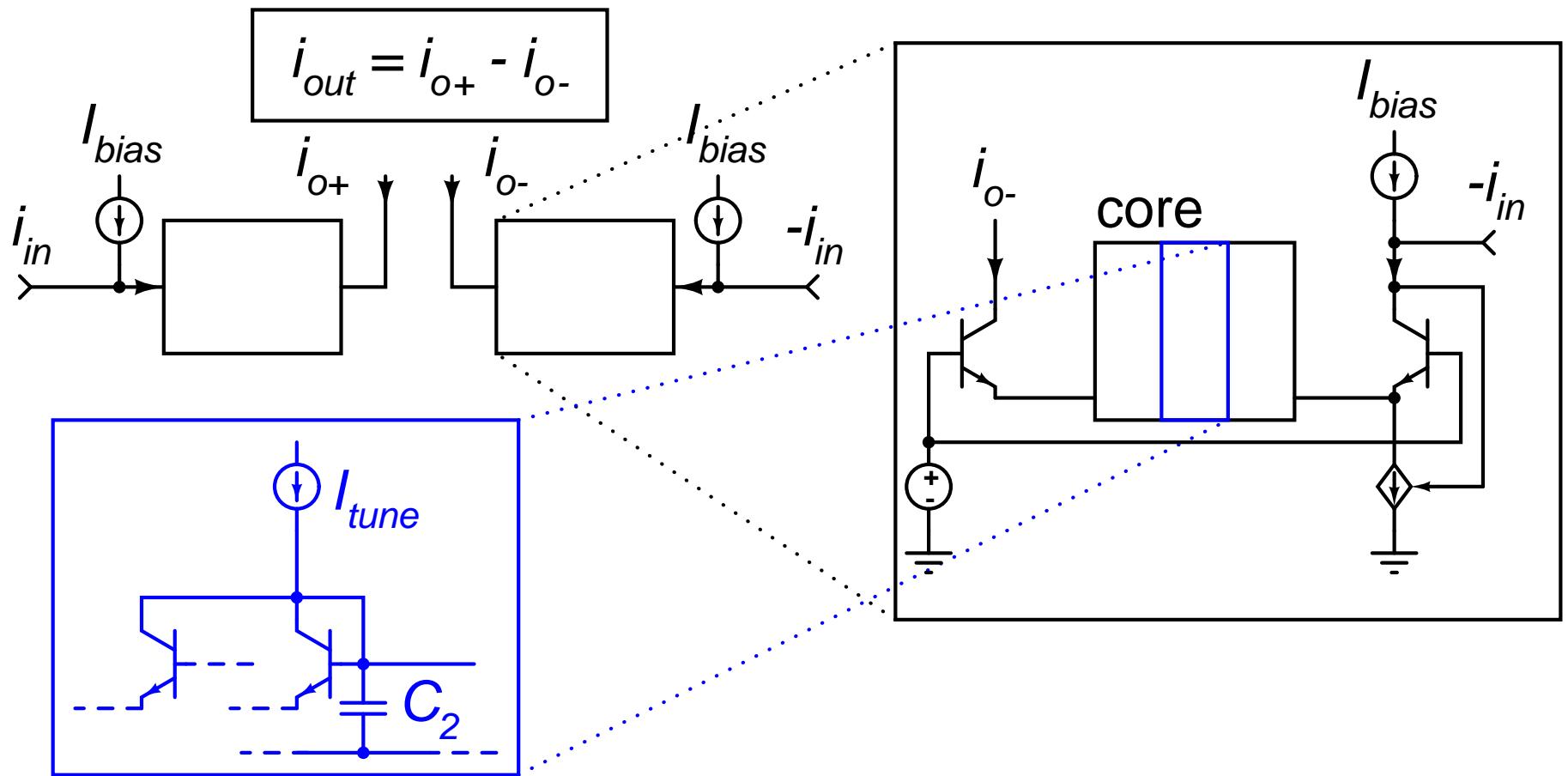
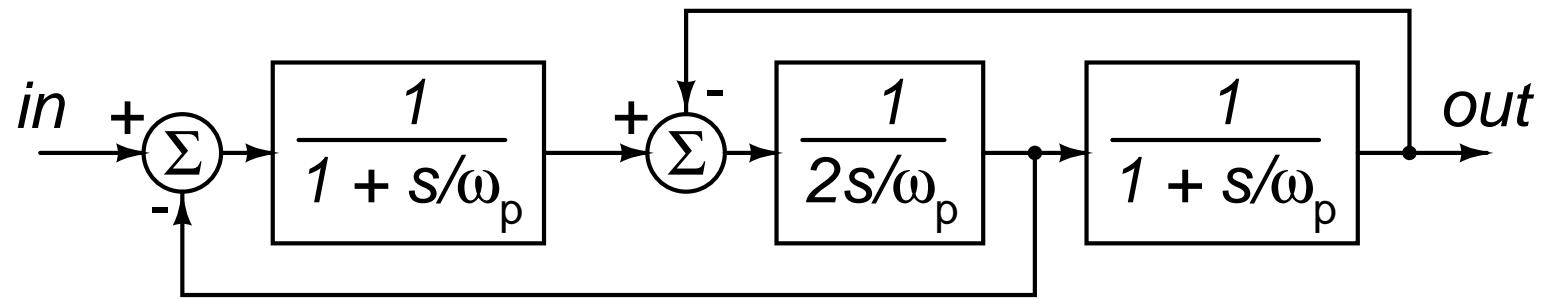
- Large signal linearity \Rightarrow cancellation of I_{bias} .

Dynamic biasing

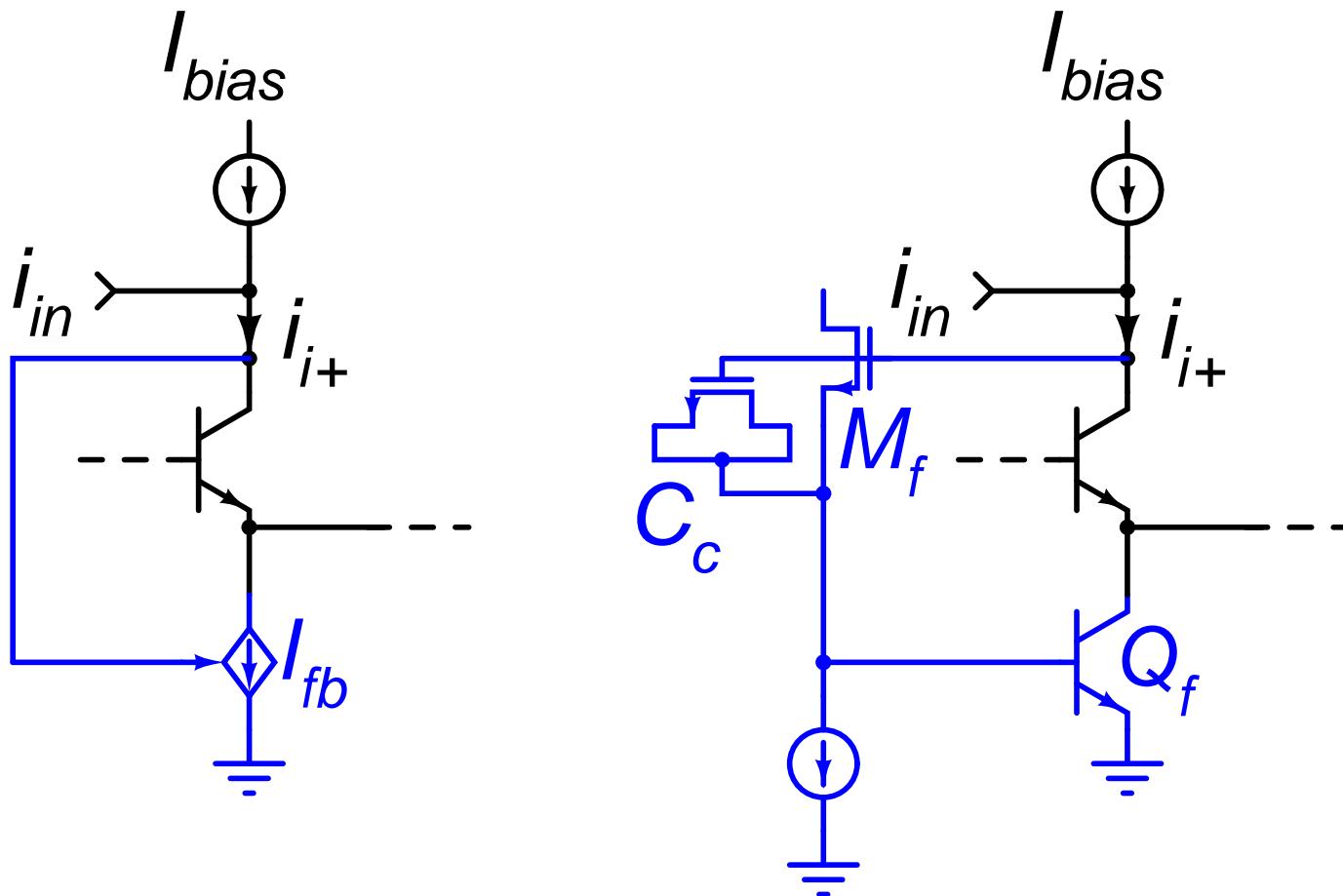


- Decrease I_{bias} for small input amplitudes.
- SNR improvement for small signals while maintaining linearity and time-invariance.
- Output noise depends on *total* input signal.

Third-order Butterworth filter

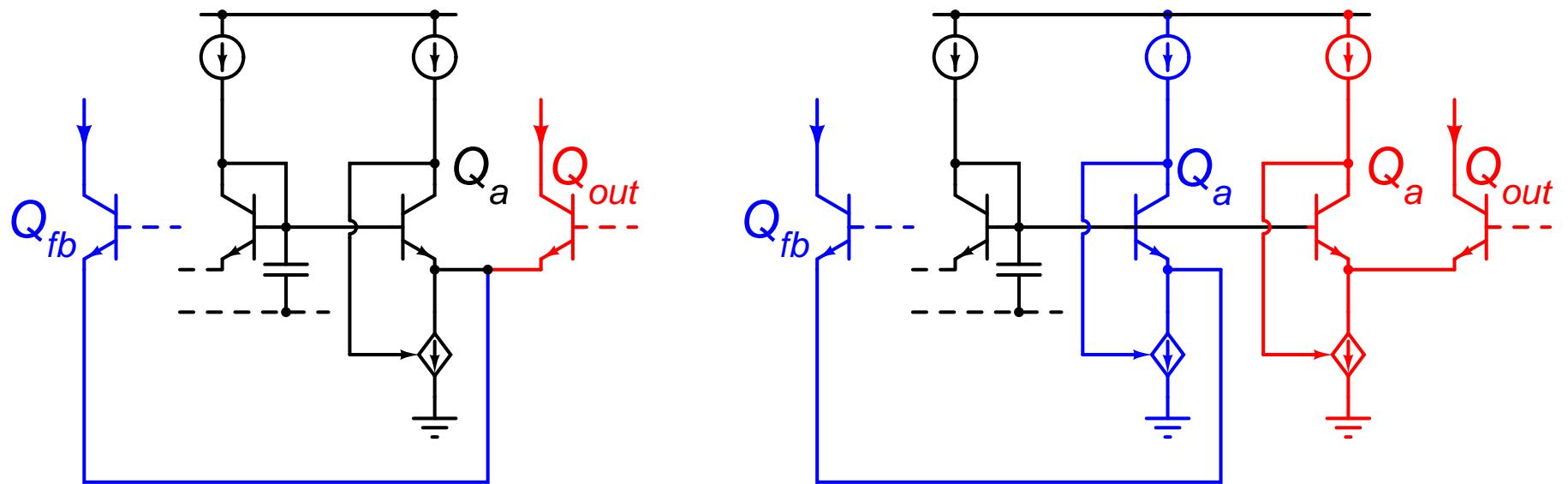
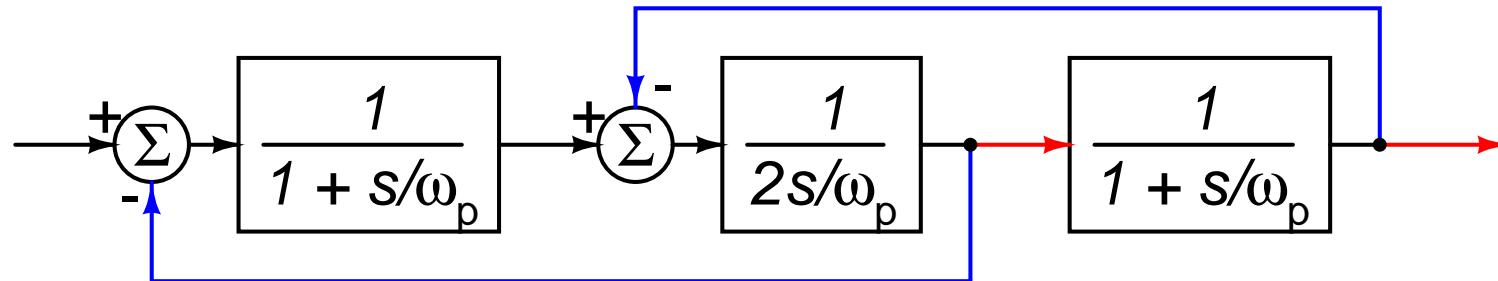


Minimizing distortion due to Early effect



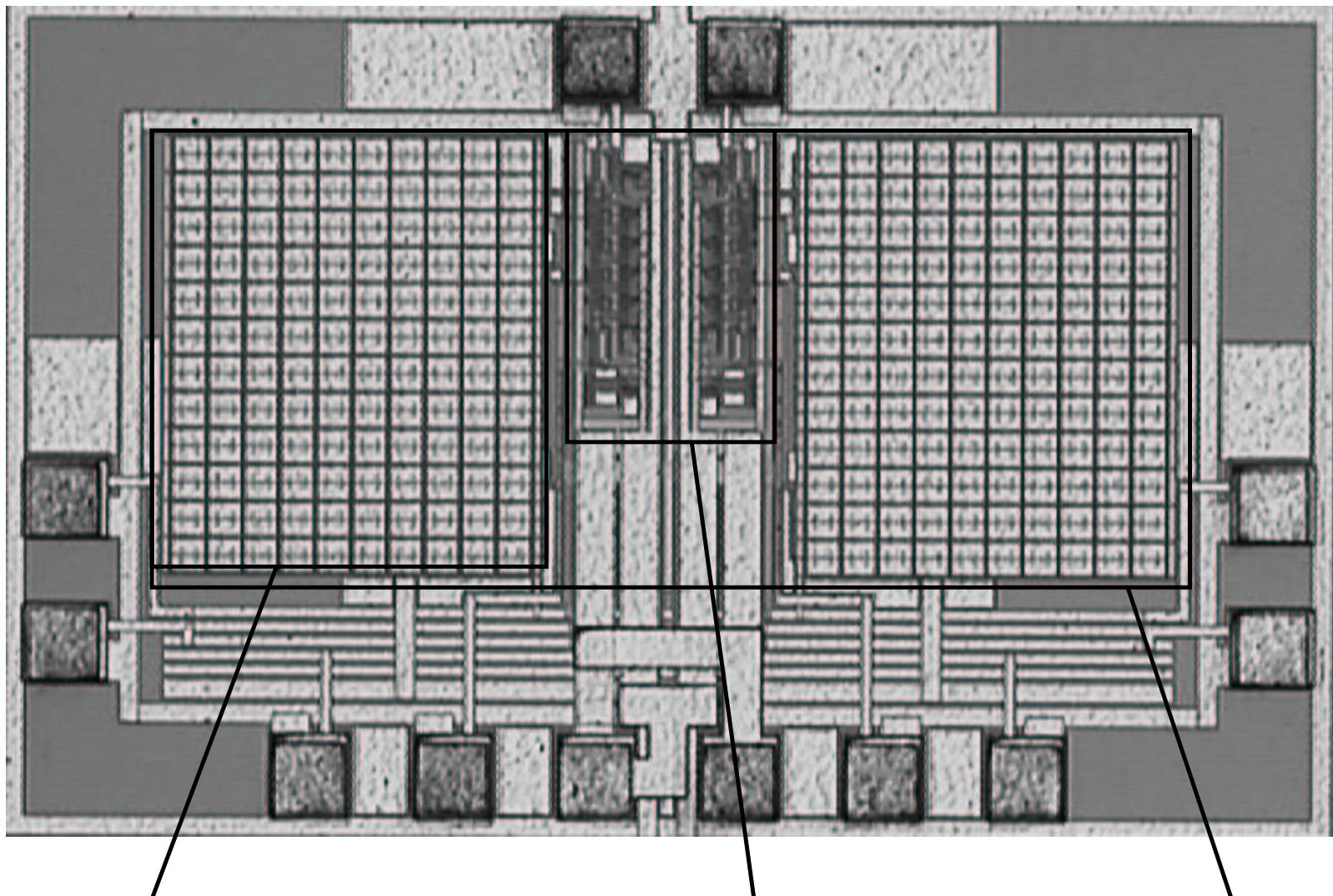
- Large g_m for I_{fb} to reduce distortion due to Early effect.
- C_c : pole-zero compensation.

Feedback paths in the filter



- Separate emitter follower for the feedback path ensures frequency response integrity.

Chip photograph



capacitor (0.19 mm^2)

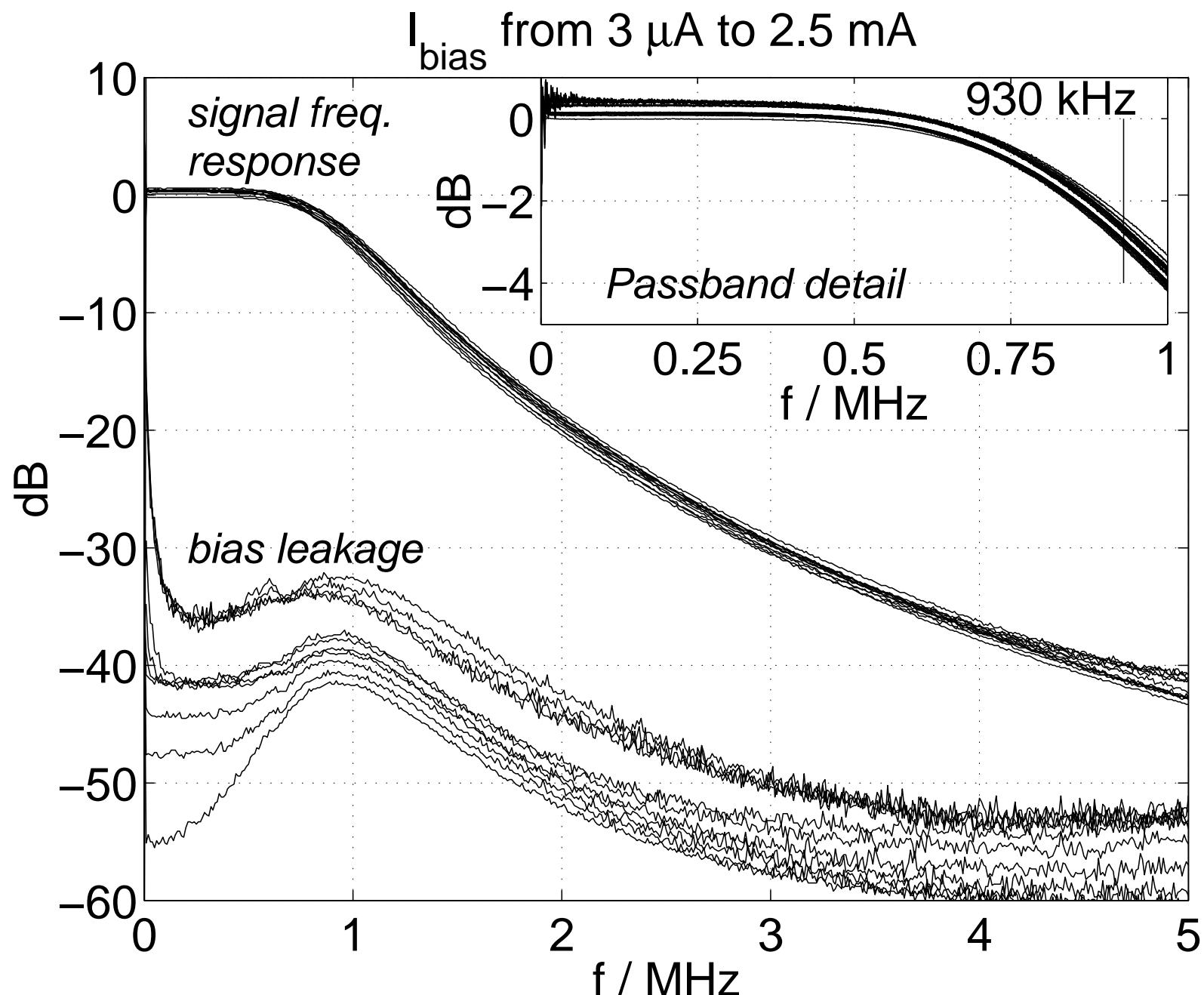
active circuit

0.52 mm^2

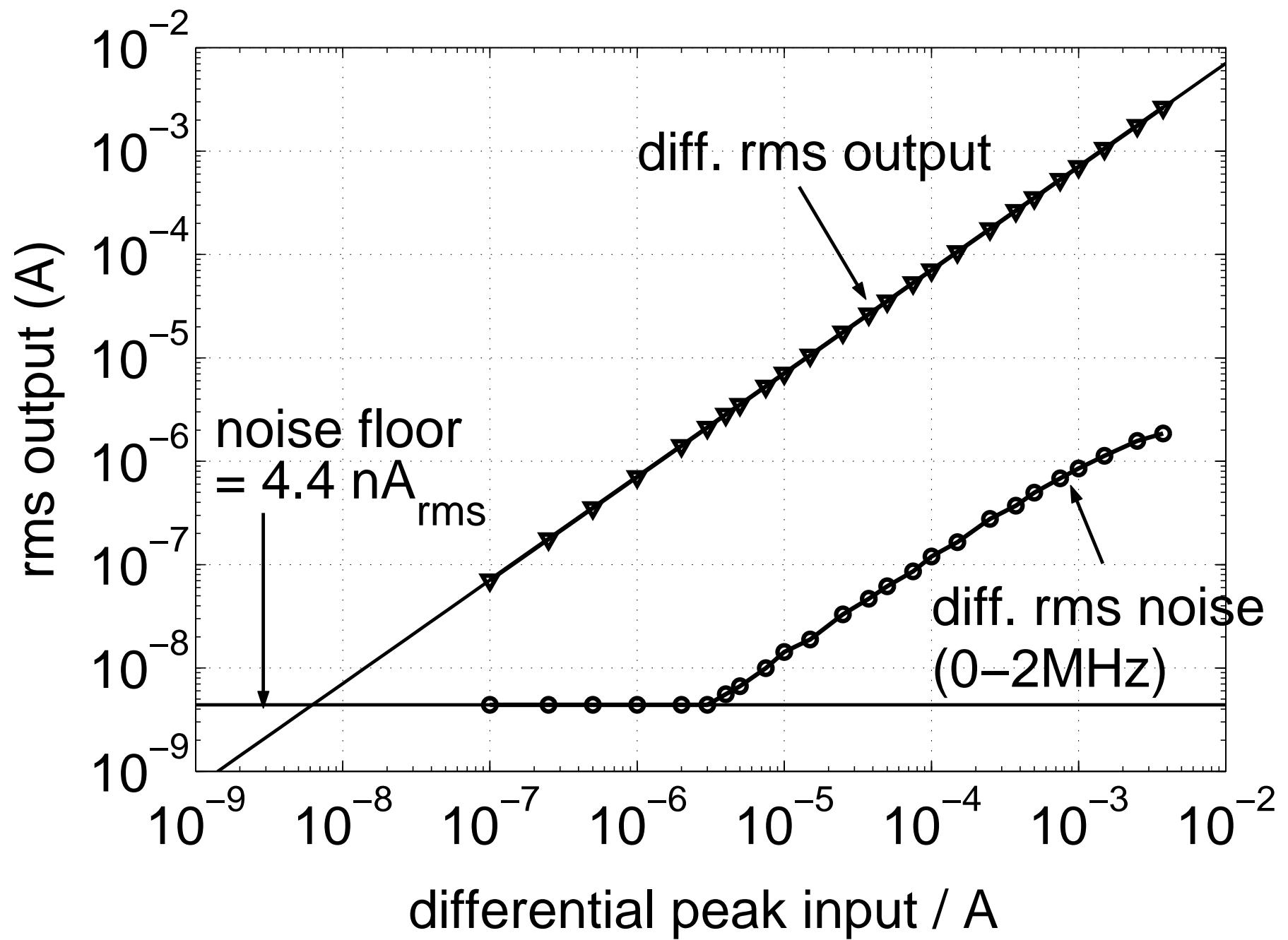
Measurement conditions

- $V_{dd} = 2.5 \text{ V}$.
- $I_{tune} = 5 \mu\text{A}$.
- I_{bias} : $3 \mu\text{A}$ to 2.5 mA .
- “Dynamic” bias: $I_{bias} = 2 \times$ the single-ended input peak, subject to a minimum of $3 \mu\text{A}$.
- 400 kHz input for THD measurements.
- $1 \text{ MHz} \pm 20 \text{ kHz}$ inputs for IM_3 measurements.

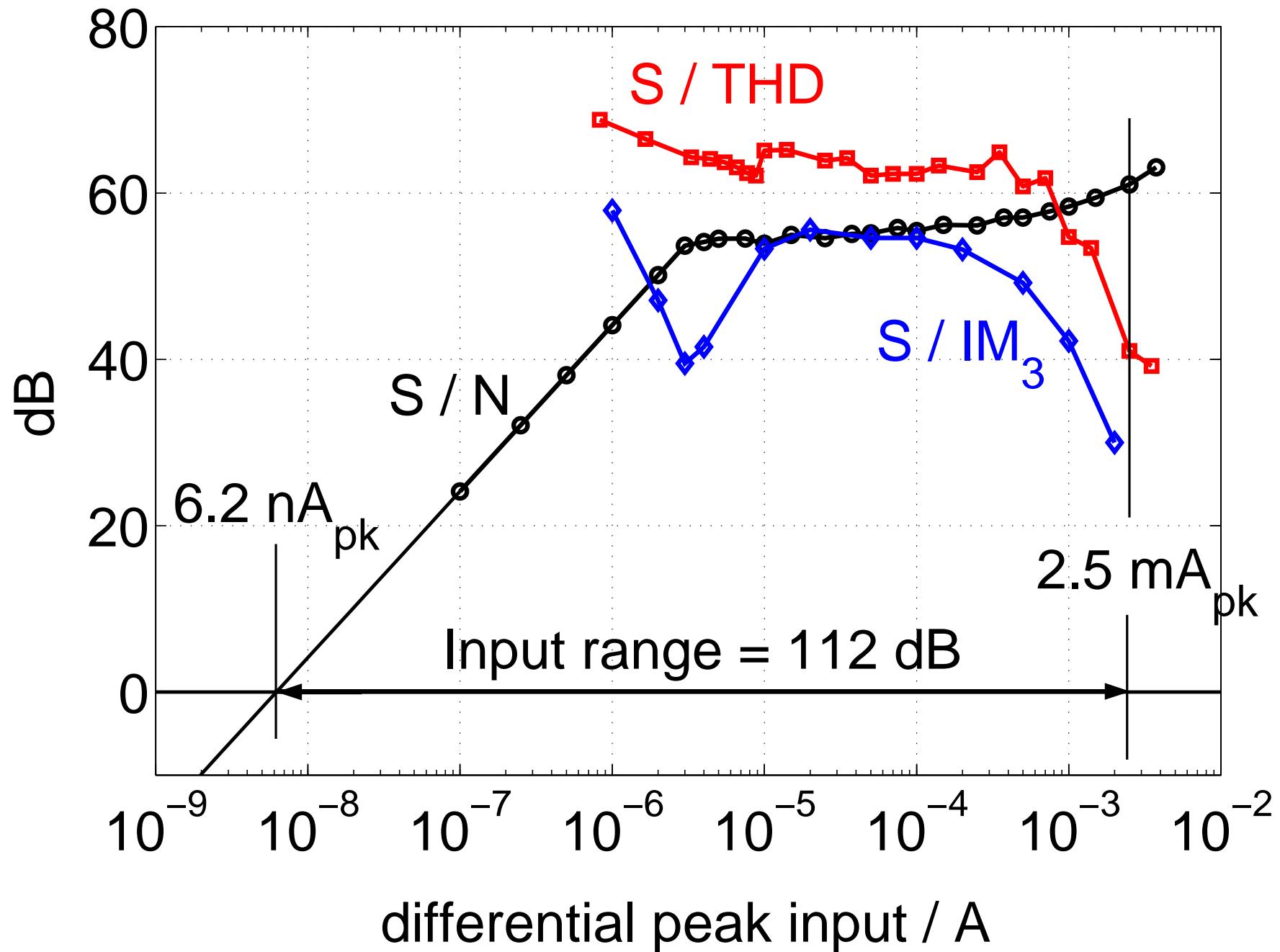
Measured results: Frequency response



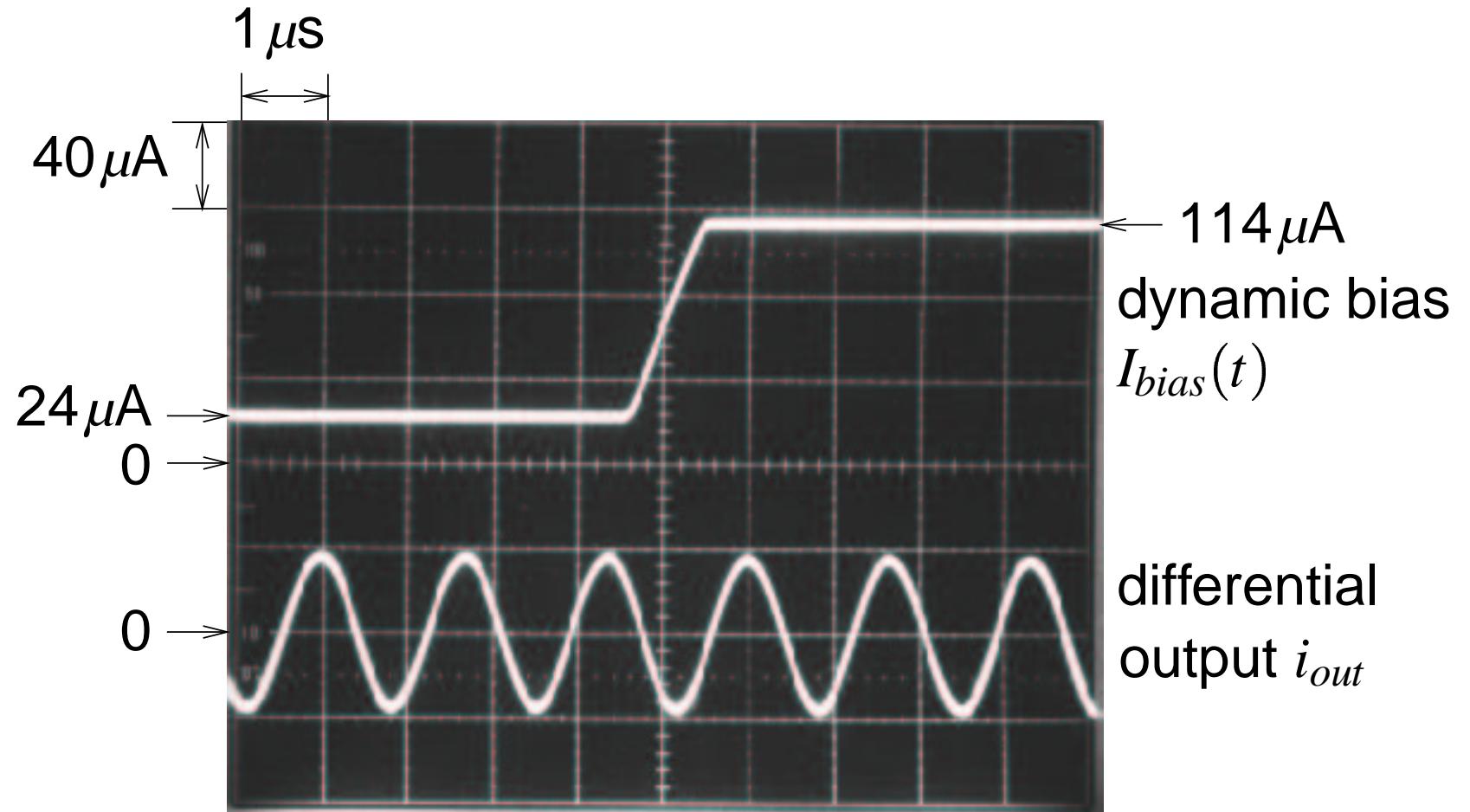
Measured results: Noise



Measured results: Distortion

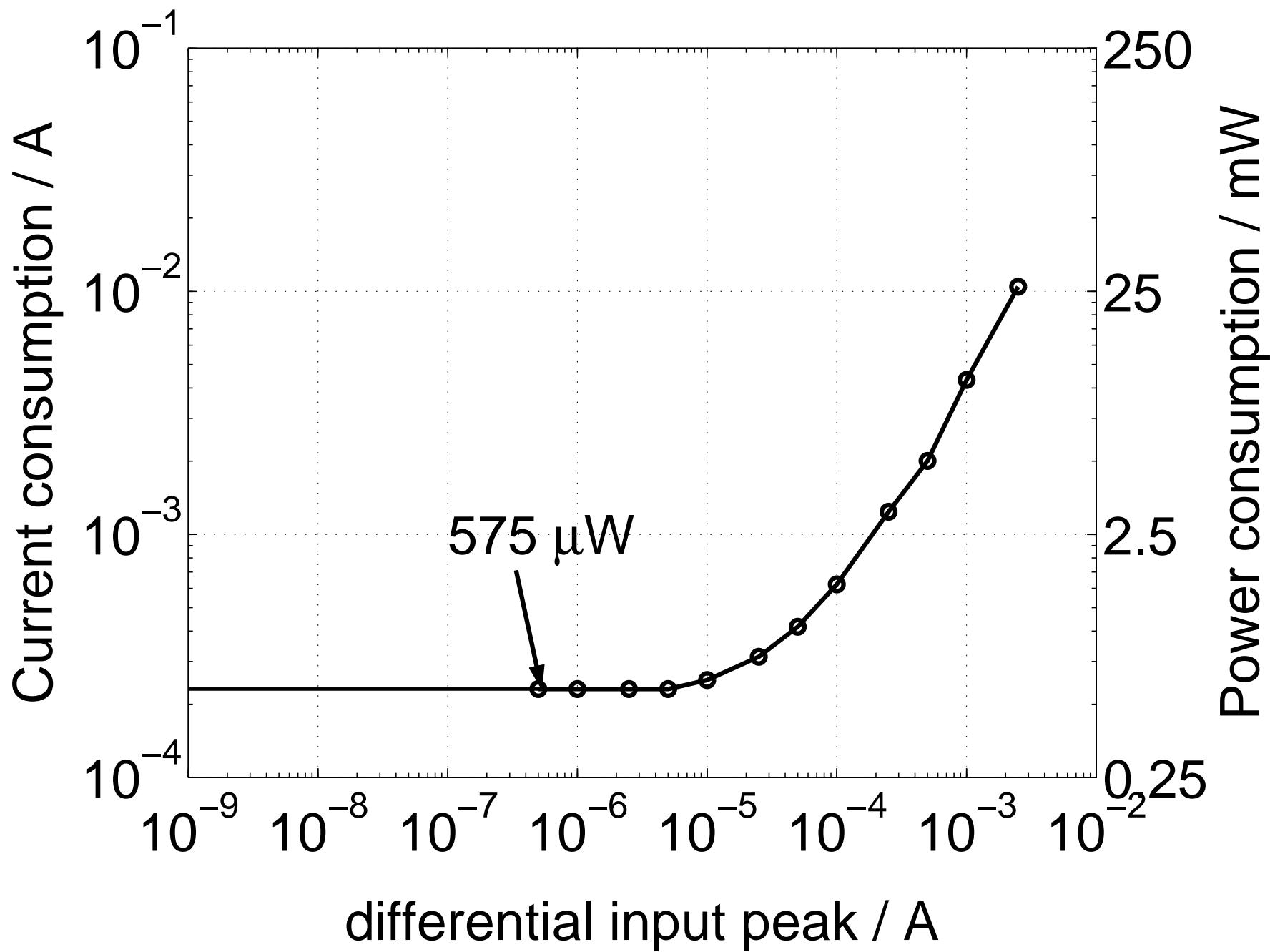


Output with dynamically changing bias



- 20 μA single-ended peak input @ 600 kHz.
- I_{bias} switched from 24 μA to 114 μA .
- Barely perceptible change in the output.

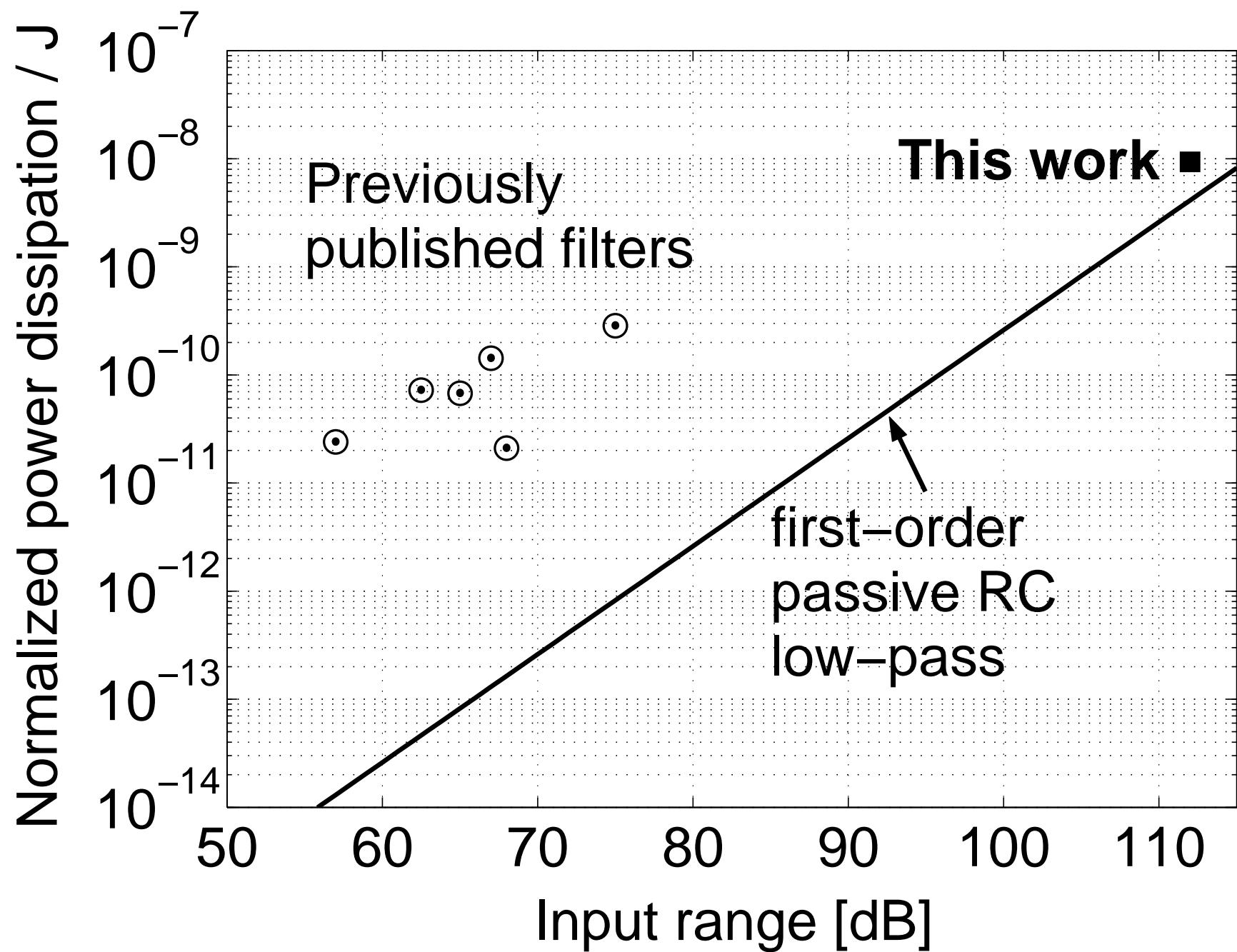
Measured results: Power consumption



Summary

Technology	0.25 μ m BiCMOS	
Area (excl. pads)	0.52 mm ²	
Supply voltage	2.5 V	
-3 dB BW	930 kHz	
I_{bias}	3 μ A	2.5 mA
Power diss. (P_d)	575 μ W	26.1 mW
Output noise	4.4 nA	1.5 μ A
THD (input peak = 0.5 I_{bias})	-64.3 dB	-41 dB
IM ₃ (input peak = 0.5 I_{bias})	-39.5 dB	-30 dB
Input range (THD \leq 40 dB)	112 dB	
Maximum P_d / Order · BW	9.35 nJ	

Comparison



Conclusions

- A technique for dynamic biasing without disturbing the output.
- Noise and power reduction for small input signals.
- A log-domain filter with 112 dB input range in a $0.25\mu\text{m}$ BiCMOS technology.
- Quiescent power: $40\times$ smaller than maximum.
- Over an order of magnitude improvement in power efficiency of analog filters.