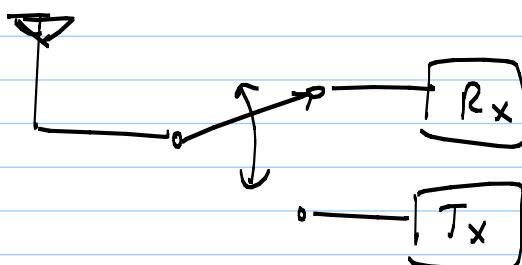


Lecture #2

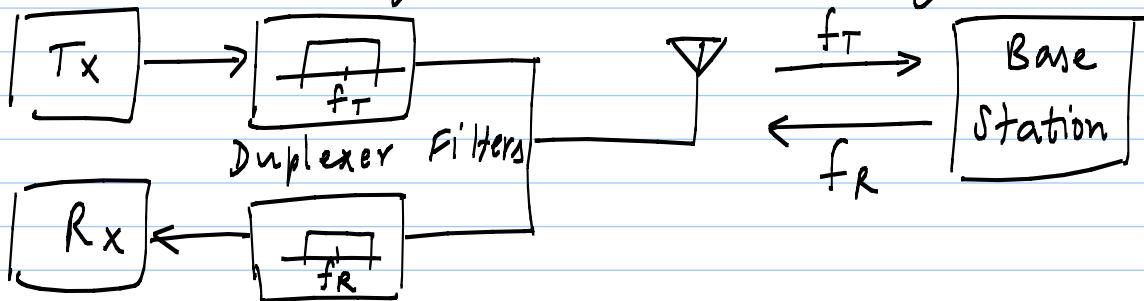
Multiple Access Techniques

1. TDD - time division duplexing



- * Direct comm with peers
- * R_x/T_x not on at same time
- * RF switch loss is critical
- * Nearby T_x can desensitise R_x

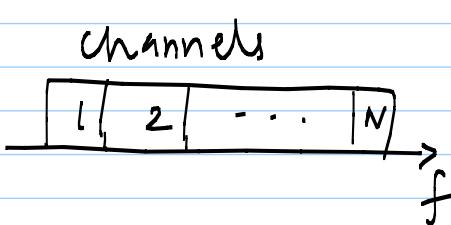
2. FDD - Frequency division multiplexing



- * BPFs isolate $R_x - T_x$
- * T_x/R_x duplex (ie on at same time)
- * Base station performs $f_T \leftrightarrow f_R$ conversion
- * Loss of Duplexer/BPFs is critical
(usually $>$ RF switch loss)

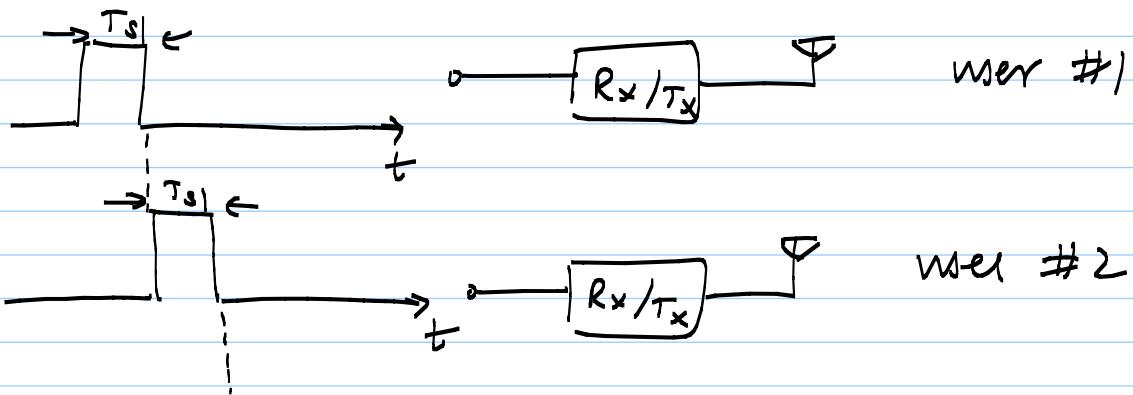
3. FDMA - Frequency Division Multiple Access

- * Each user is assigned a channel
(fixed till end of call)



- * Separate channels for R_x & T_x
- * Max # of simultaneous users = $\frac{\text{freq. Band Width}}{\text{channel width}}$
⇒ insufficient capacity in crowded areas!

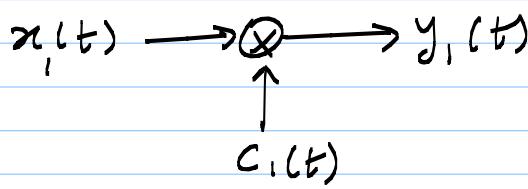
4. TDMA — Time division multiple Access



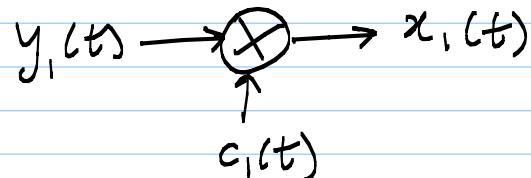
- * Data transmitted/received as a "burst"
- * Same freq. band shared in time
- * Each Rx/Tx uses a time slot
- * All slots put together \equiv a Frame
- * PA can be turned on/off to save power
- * More complex - time synchronisation

5. CDMA — Code Division Multiple Access

at Tx_1 :



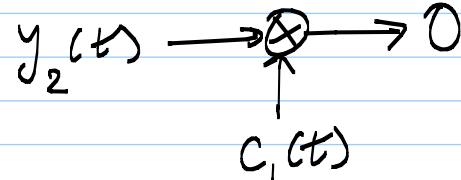
at Rx_1 :



at Tx_2 :



at Rx_1 :



Basic idea: use orthogonal codes to separate users in signal space:

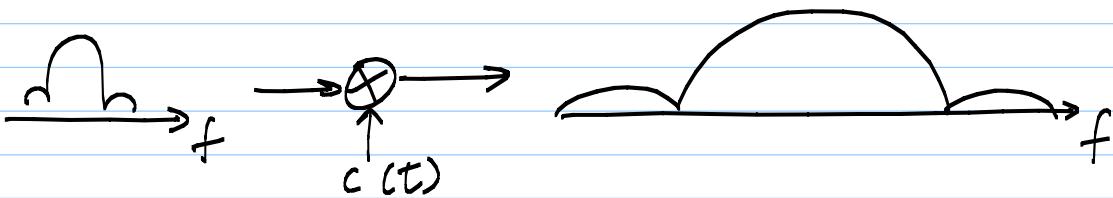
$$\begin{aligned} c_i \cdot c_j &= [0] \text{ for } i \neq j \\ &= [1] \text{ for } i = j \end{aligned}$$

examples of orthogonal codes:

$$c_1 = (1, 1, 1, 1) ; c_2 = (1, -1, 1, -1) ;$$

$$c_3 = (1, 1, -1, -1) ; c_4 = (1, -1, -1, 1) ;$$

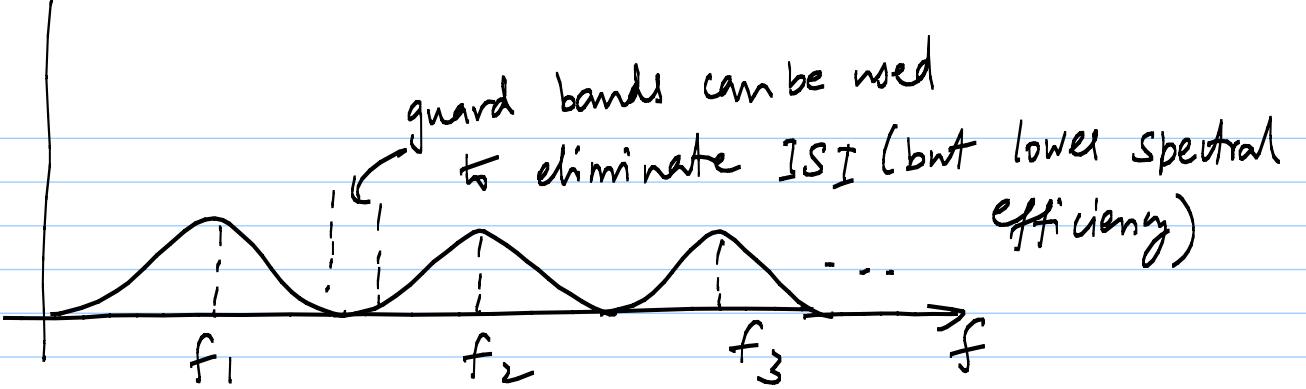
Codes can be used to spread the baseband signal (spread-spectrum comm.)



- * BW usage (apparently) increases, but multiple users can occupy overlapping frequencies
- * Other users' signals appear as random noise (ideally)
- * CDMA has a soft capacity limit - increasing the number of users gradually raises the noise floor.
- * Power control is critical - one single large Tx signal can raise the noise floor for all users.
- * Note : TDMA - orthogonality in time
FDMA - orthogonality in frequency

6) OFDM - Orthogonal Frequency division multiplexing

- * Tx BW divided into many sub-channels
- * Sub-channels orthogonal to each other

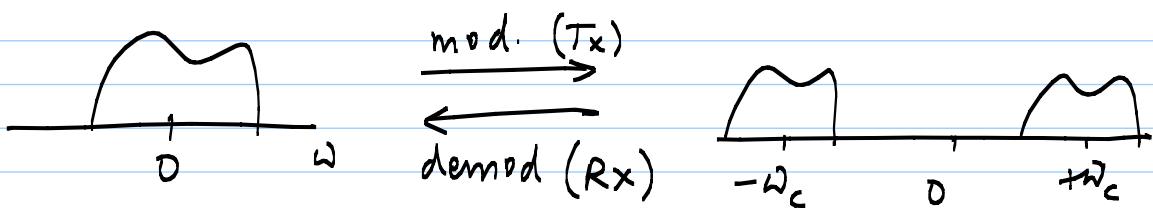


* Can be designed such that sinc response nulls occur at adjacent sub-carrier frequencies

Now, for some examples:

- 1) GSM - TDMA/FDMA and FDD
- 2) Qualcomm CDMA - CDMA and FDD
- 3) DECT (cordless telephone) - TDMA/FDMA and TDD
- 4) 802.11g (WiFi) - OFDM and TDD

Analog & Digital Comm.



Baseband
signal

Passband signal
(with carrier)

Why modulation?

wired systems: shielding is better at high freq.

wireless systems: a) antenna size can be smaller ($\propto \lambda$)
b) share spectrum through regulation (to avoid overlap)

Passband Signal

$$x(t) = a(t) \cdot \cos [\omega_c t + \theta(t)]$$

↑
 amplitude ↑
 phase

$$\text{frequency} = \omega_c + \frac{d\theta(t)}{dt}$$

Analog Modulation : AM, FM (PM)

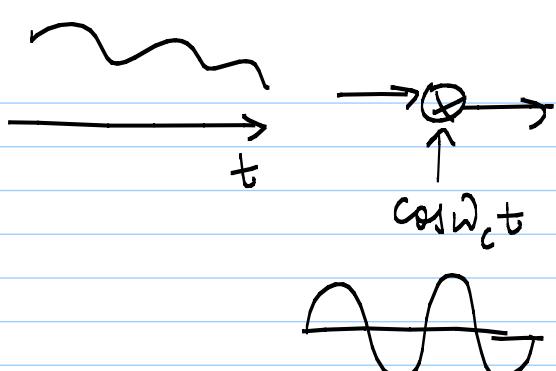
1. AM - amplitude modulation

$$x_{AM}(t) = A_c [1 + m \cdot x_{BB}(t)] \cos \omega_c t$$

\uparrow
 modulation
 index

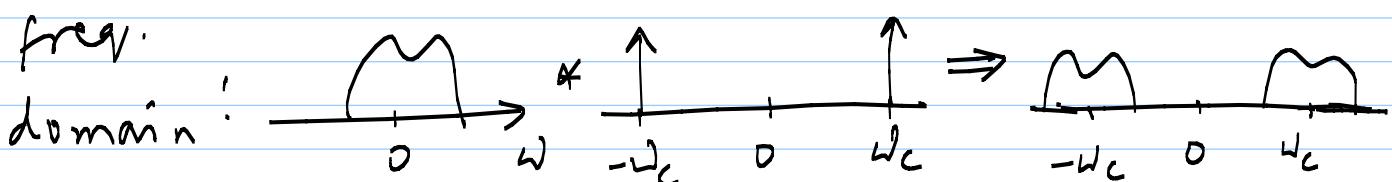
DSB-FC \Rightarrow Double-sideband full-carrier

time domain

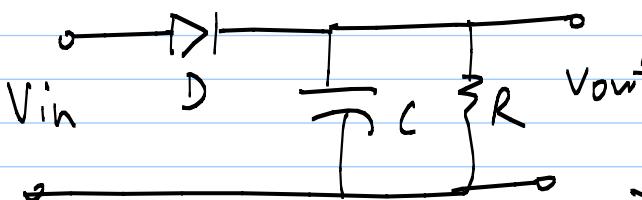


A hand-drawn diagram on lined paper showing an envelope detector circuit. A vertical line labeled "actual signal" has a wavy, periodic waveform. Above it, a horizontal line labeled "envelope" has a smooth, rounded rectangular waveform that follows the peaks of the actual signal. An arrow points from the label "envelope" to the top line, and another arrow points from the label "actual signal" to the bottom line.

free.



AM detector :



can be used as long
as $1 + m \sigma_{BB}^L(t) > 0$
for all t

Issues: 1) susceptible to noise 2) requires very linear PA

2. FM - frequency modulation

* mod. & demod. easier than PM

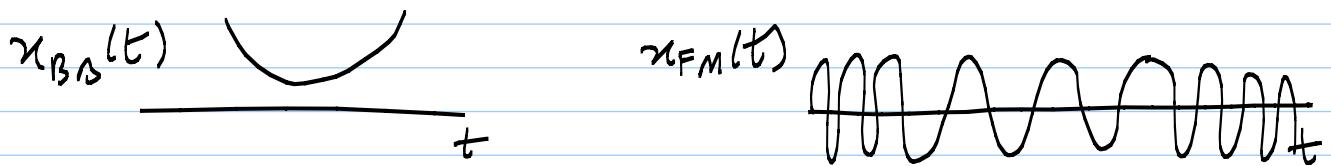
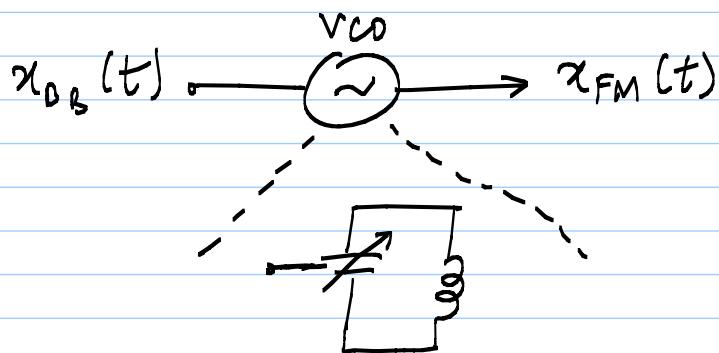
$$x_{FM}(t) = A_c \cos \left[\omega_c t + m \int_{-\infty}^t x_{BB}(t') dt' \right]$$

FM index

$$x_{PM}(t) = A_c \cos \left[\omega_c t + m x_{BB}(t) \right]$$

PM index

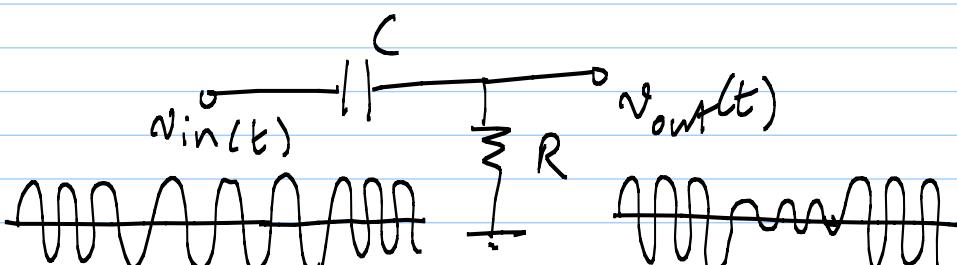
Simple FM modulator:



Simple FM demodulator:

$$\begin{aligned} v_{in(t)} &\xrightarrow{\text{HPF}} v_o(t) = A \sin \left[\omega_c t + m x_{BB}(t) \right] \\ &= x_{FM}(t) \end{aligned}$$

↑ differentiation



use envelope
detector on this!

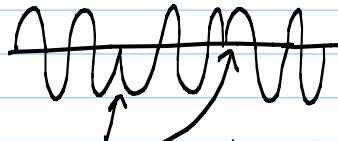
Digital Modulation: ASK, PSK, FSK etc.

↑
less sensitive to
amplitude noise

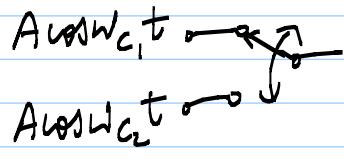
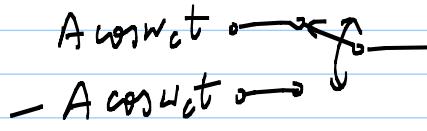
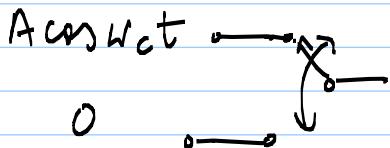
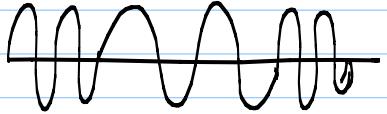
ASK

PSK

FSK



phase transitions



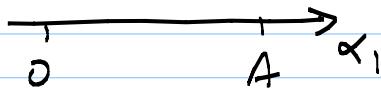
{ Binary PSK }
or BPSK

{ BFSK }

Signal constellation Diagram:

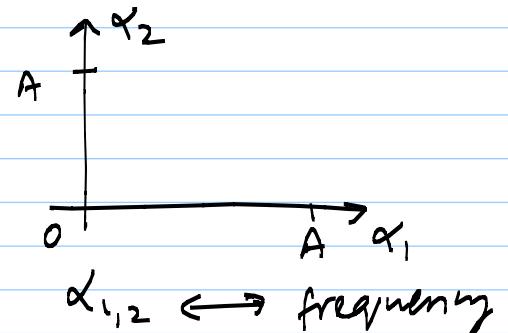
Represents possible symbols in complex plane

ASK

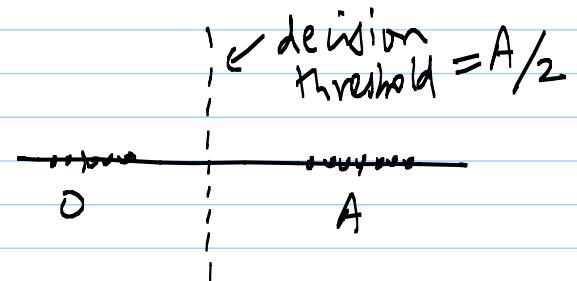


$\alpha_1 \leftrightarrow$ amplitude

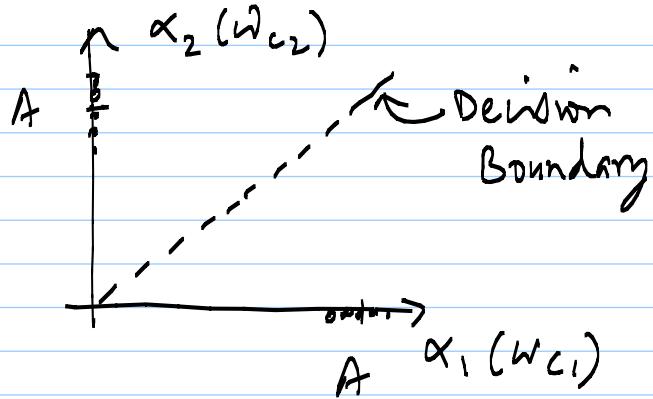
FSK



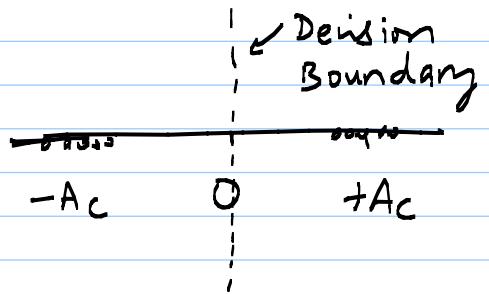
effect of noise in ASK:



Noise with FSK:



BPSK

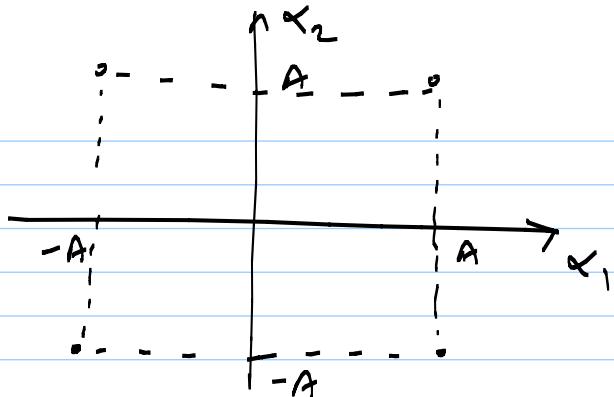


Quadrature Modulation:

$$x(t) = b_m A \cos \omega_c t - b_{m+1} A \sin \omega_c t$$

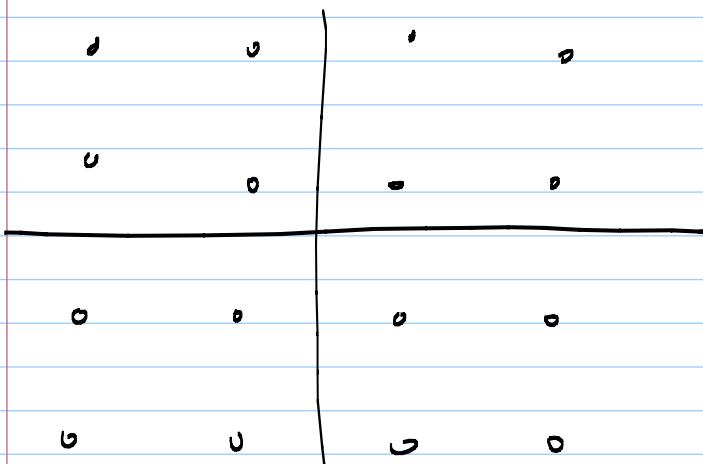
e.g. QPSK, MSK, 8-PSK
 (WCDMA) \MSK (EDGE)
 (GSM)

QPSK:



In general,
only certain
transitions
are allowed

QAM - Quadrature Amplitude Modulation



e.g. HSDPA
uses 16-QAM