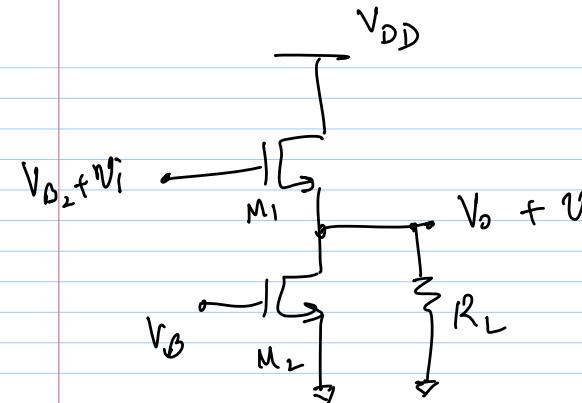
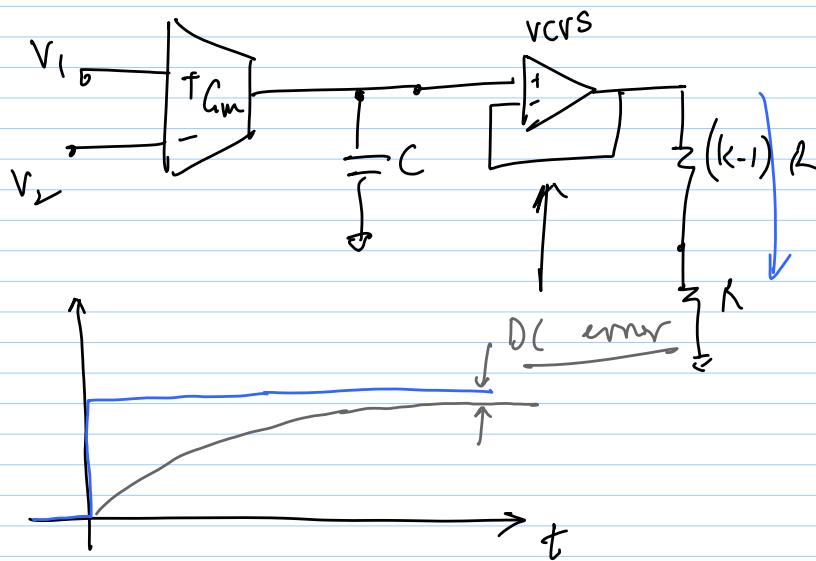


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Lec 18

$$V_{o\min} = V_{DSAT_2}$$

$$V_{o\max} = V_{DD} - V_{DSAT_1}$$

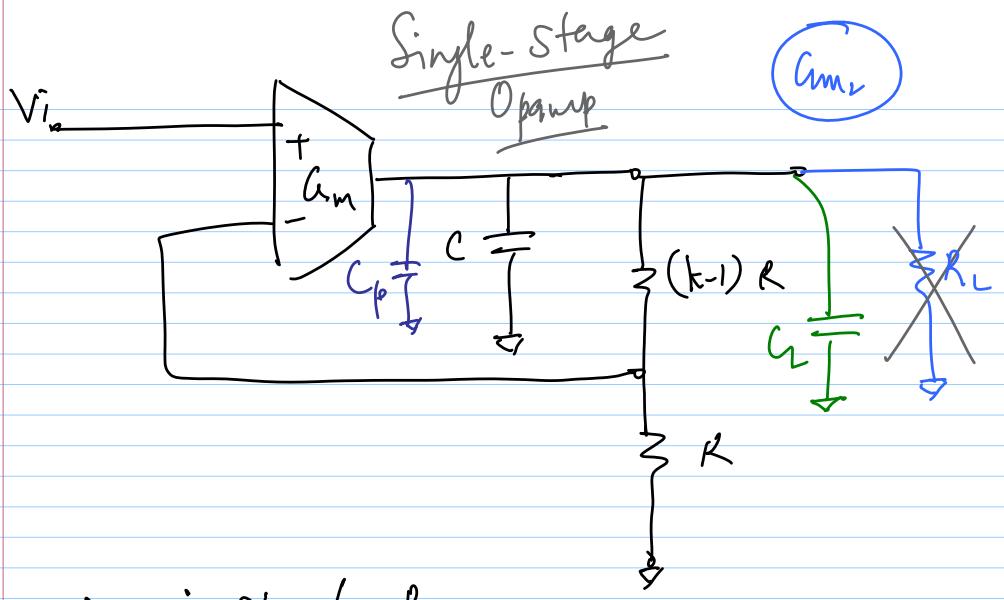
$$\Rightarrow r_{A_1} = V_{DD} + V_T_1$$

$$V_{o\max} = V_{DD} - V_{GS},$$

\* Increase  $R$  so that  $\text{hm}R$  is large enough

$$G_m \uparrow \Rightarrow I_{bias} \uparrow$$

$$\Rightarrow W/L \uparrow \left\{ \begin{array}{l} \text{par Cap.} \uparrow \\ \text{smaller devices} \end{array} \right.$$



$$\text{DC gain} \approx \text{hm}R_L$$

Single stage Opamp

- 1) DC gain  $\Rightarrow$  do not drive resistive loads ( $G_m R_L$ )
- 2)  $G_m = g_{m1}$ ;  $\omega_n = g_{m1}/C_L$
- 3)

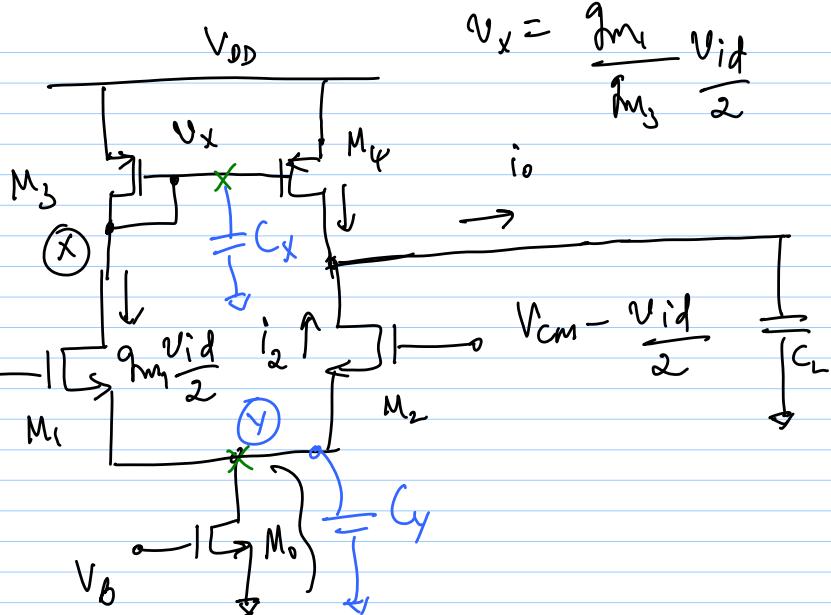
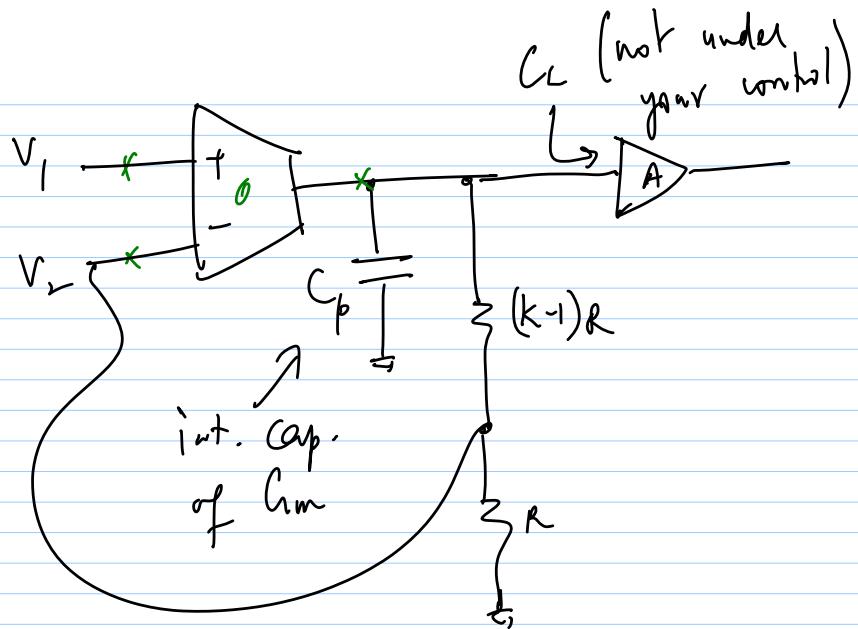
load cap  $C_L$

↳ acts as integrating cap

If you have  $C + C_L$

$$w_u = \frac{g_m'}{C_L + C_L} \quad \left( \text{if } g_m' \gg g_m \right)$$

$$w_u = \frac{g_m}{C_L}$$



$$i_4 = g_{m_4} V_x$$

$$= g_{m_1} \frac{V_{id}}{2} \quad (g_{m_4} = g_{m_1})$$

$$i_2 = g_{m_2} \frac{V_{id}}{2}$$

$$i_o = i_2 + i_4 = g_{m_1} V_{id}$$

$$\underline{\text{No } C_L} \Rightarrow V_o = i_o \cdot r_o = g_{m_1} V_{id} (r_{ds_1} || r_{ds_2})$$

$$\text{No } r_o \Rightarrow V_o = i_o \cdot \frac{1}{sC_L}$$

$$= \frac{g_{m1}}{sC_L} V_{id}$$

Actual dkt

$$\frac{V_o}{V_{id}} = \frac{g_{m1}}{g_o + sC_L} = \frac{g_{m1}r_o}{1 + sC_Lr_o}$$

\* finite dc gain =  $g_{m1}r_o = g_{m1}/g_o$

$$i_4 = \frac{g_{m4} \cdot g_{m1} V_{id}/2}{g_{m3} + sC_X}$$

$$i_2 = g_{m1} V_{id}/2$$

$$i_o = i_4 + i_2 \\ = g_{m1} \frac{V_{id}}{2} \left[ 1 + \frac{g_{m4}}{g_{m3} + sC_X} \right]$$

\* Swing limits

\* noise

\* slew rate

\* Non-dominant poles & zeros

$\frac{C_Y}{C_X}$  - No pole due to  $V_L$

$$\frac{C_Y}{C_X} - V_X = i_1 \left( \frac{1}{g_{m3}} \parallel \frac{1}{sC_X} \right)$$

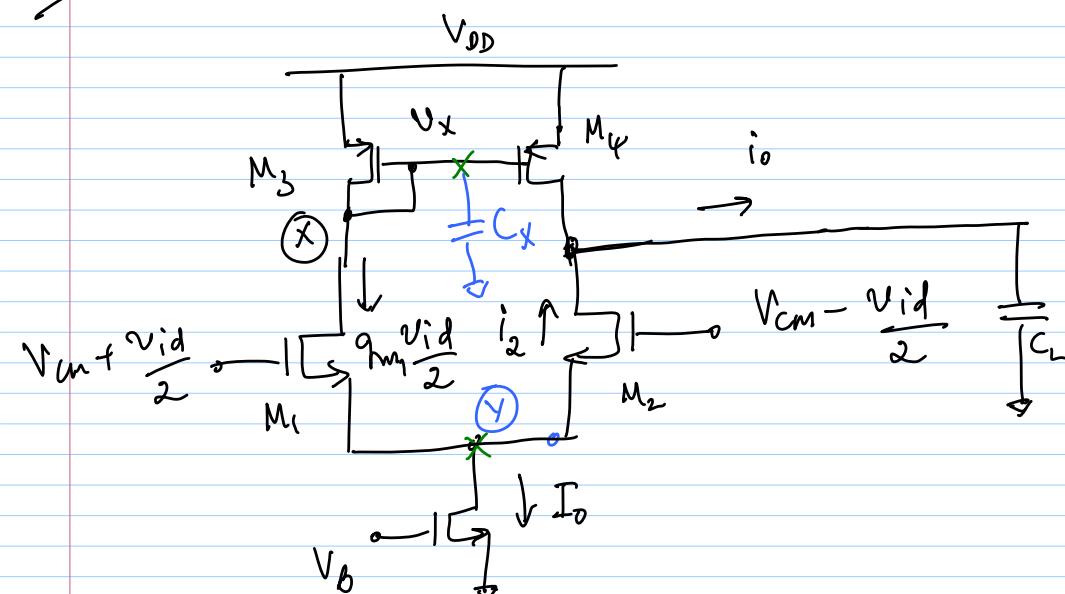
$$= \frac{i_1}{g_{m3} + sC_X}$$



$$i_4 = g_{m4} \cdot V_X$$

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Lec 19



$$i_o = i_2 + i_4$$

$$= g_{m_1} \frac{V_{id}}{2} + \frac{g_{m_4} g_{m_1} V_{id}/2}{g_{m_3} + sC_x}$$

$$= g_{m_1} \frac{V_{id}}{2} \left[ 1 + \frac{g_{m_3}}{g_{m_3} + sC_x} \right]$$

$$V_o = \frac{i_o}{g_o + sC_L}$$

$$|w_n| = \frac{g_{m_1}}{C_L}; |w_p| = \frac{g_o}{C_L}$$

$$\beta_w = \frac{\omega_n}{k};$$

$$\omega_p = -\frac{g_{m_3}}{C_x}; \omega_z = -\frac{2g_{m_3}}{C_x} = 2\omega_p$$

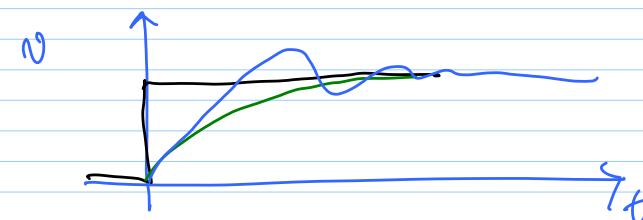
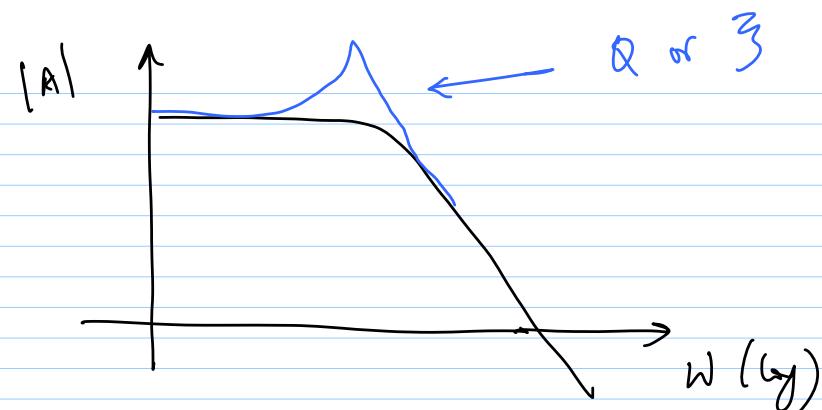
Unity gain  $\Rightarrow \omega_p \rightarrow \omega_n$

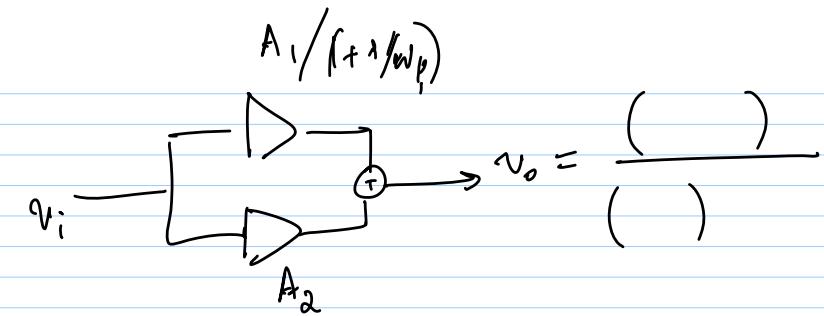
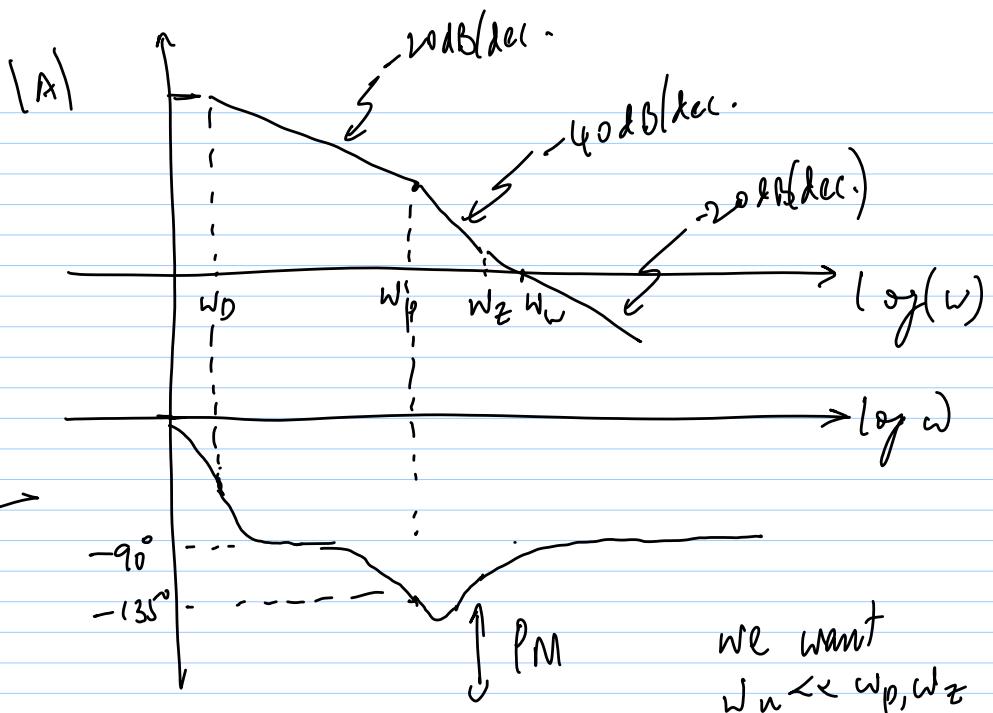
$$\frac{V_o}{V_{id}} = \frac{g_{m_1}}{2} \left[ 1 + \frac{g_{m_3}}{g_{m_3} + sC_x} \right] \cdot \frac{1}{g_{m_3} + sC_x}$$

$$= \frac{g_{m_1}}{2} \cdot \frac{1}{g_o + sC_L} \cdot \frac{2g_{m_3} + sC_x}{g_{m_3} + sC_x}$$

Dominant pole ( $w_D$ )  $w_z, w_p$

$$w_p = -\frac{g_{m_3}}{C_x}; w_z = -\frac{2g_{m_3}}{C_x} = 2\omega_p$$





\* multiple paths w/ different freq. response w/ phase shifts

$$\omega_D = \frac{g_m}{C_L}, \quad \omega_u = \frac{g_{m1}}{C_L}$$

$$\omega_p = \frac{g_{m3}}{C_X}; \quad \omega_z = \frac{2g_{m3}}{C_X}$$

$$\boxed{\frac{g_{m3}}{C_X} \gg \frac{g_{m1}}{C_L}}$$

- 1)  $\uparrow C_L \Rightarrow \omega_u \downarrow \text{X}$
- 2)  $\uparrow g_{m3} \Rightarrow \uparrow \left(\frac{w}{L}\right)_3 \Rightarrow C_X \uparrow \text{X}$
- 3)  $\downarrow C_X \Rightarrow \left(\frac{w}{L}\right)_3 \downarrow \text{@ same } I_o$

$$V_{DSAT_{3,4}} \uparrow$$

hit swing limits

- 4)  $\downarrow g_{m1} \Rightarrow \downarrow (w/L)_1 \Rightarrow \omega_u \downarrow$

↑  $I_o$  is the only way to improve available BW of the opamp.

### $C_M$ limits

#### 1) Input CM:

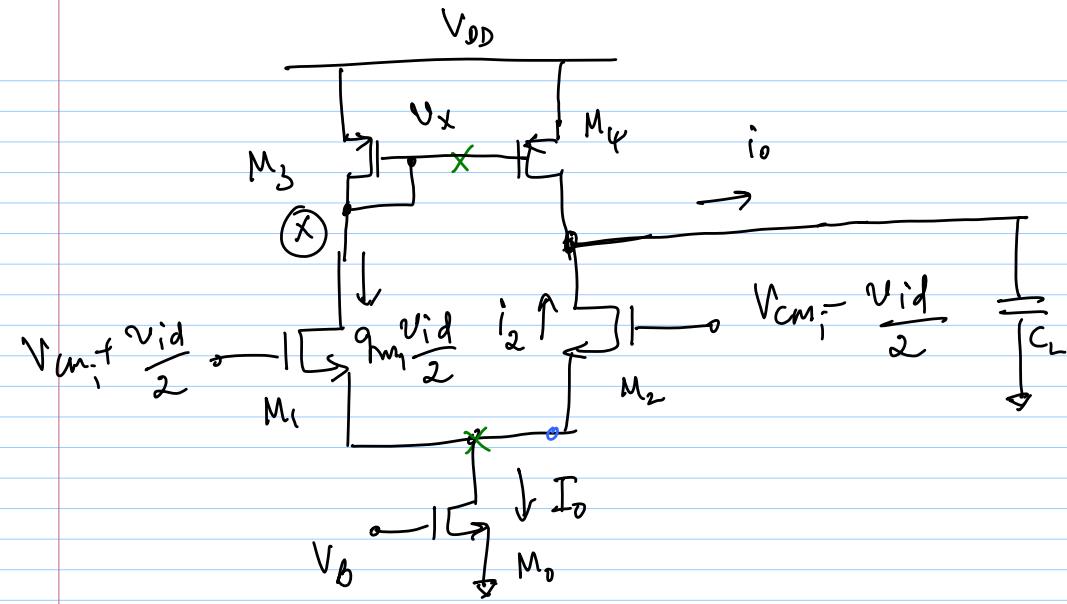
$$V_{CM,i}(\min) = V_{DSAT_0} + V_{AS_1}$$

$$V_{CM,i}(\max) = V_{DD} - V_{SA_3} + V_{T_1}$$

#### 2) Output CM:

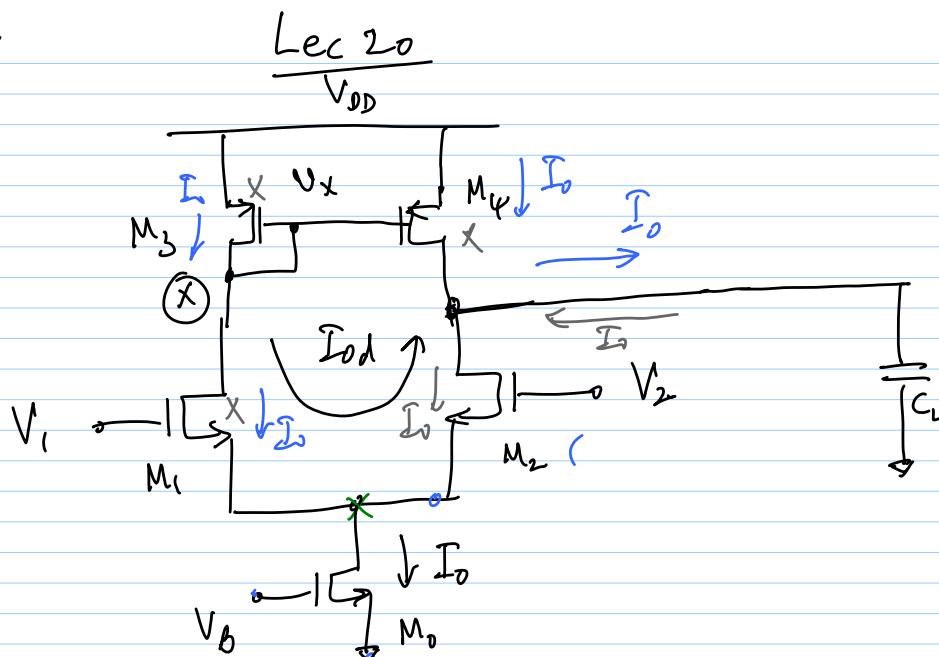
$$V_{CM_o}(\min) = V_{CM_i} - V_{T_1}$$

$$V_{CM_o}(\max) = V_{DD} - V_{OSAT_3}$$



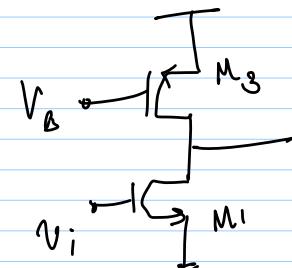
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Lec 20



# Noise

$$\bar{e_n^2} = \frac{16kT}{3} \left[ \frac{1}{g_{m1}} + \frac{g_{m3}}{g_{m1}^2} \right]$$



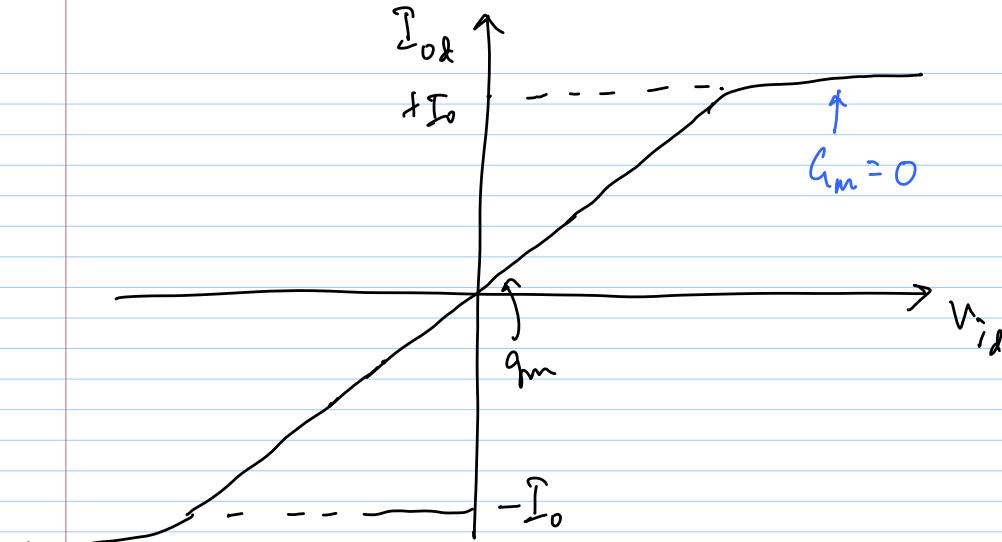
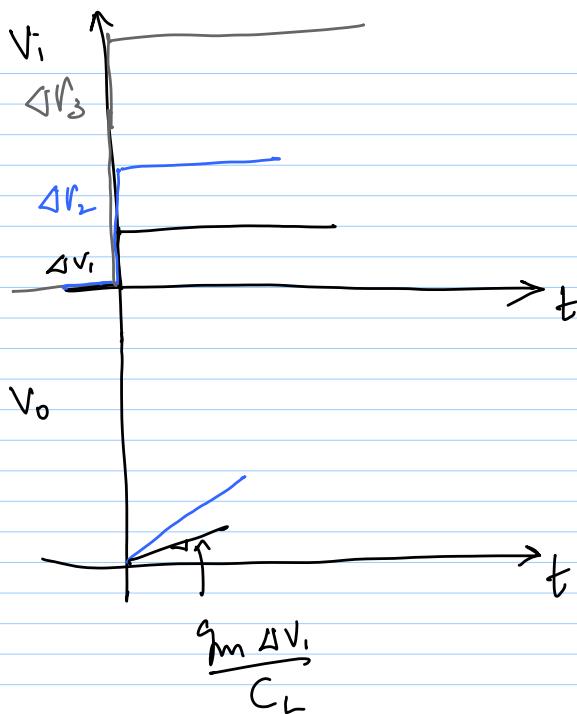
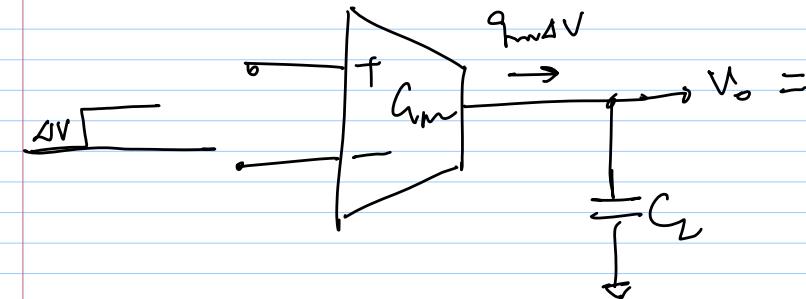
$$\frac{8kT}{3} \left[ \frac{1}{g_{m1}} + \frac{g_{m3}}{g_{m1}^2} \right]$$

\* Offset

$$V_{os} = \Delta V_{T_{1,2}} + \frac{g_{m_3}}{g_{m_1}} \Delta V_{T_{3,4}}$$

$$\sigma_{os}^2 = \sigma_{V_{T_{1,2}}}^2 + \left( \frac{g_{m_3}}{g_{m_1}} \right)^2 \sigma_{V_{T_{3,4}}}^2$$

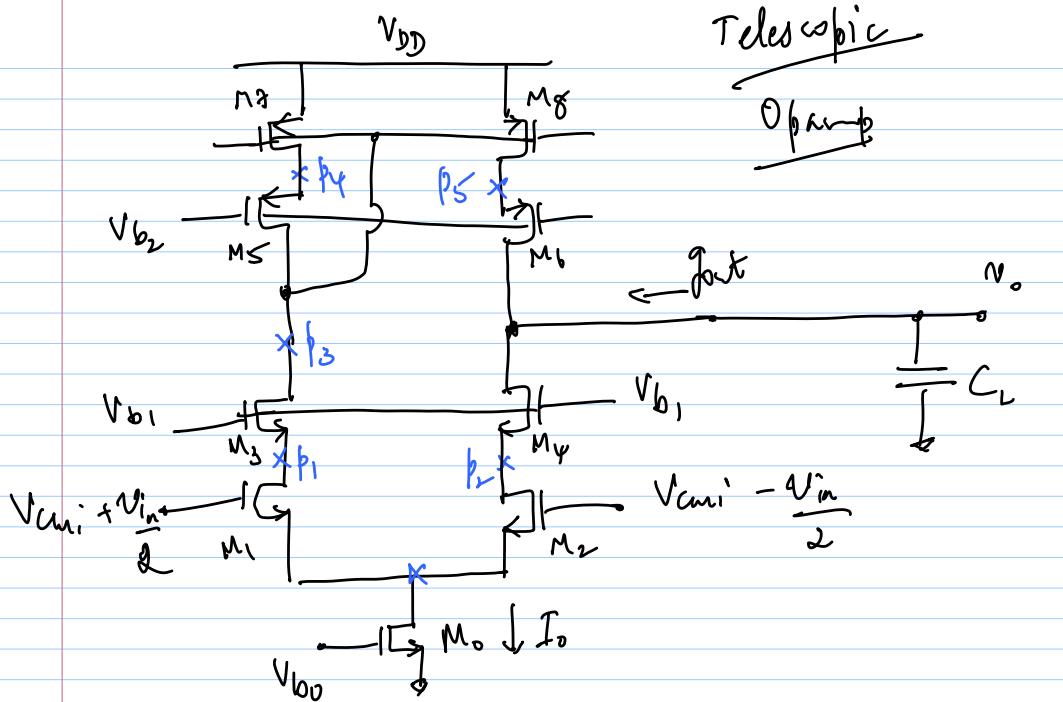
\* Slow Rate  $\frac{dV_o}{dt}$



$$I_{out \text{ (max)}} = I_o$$

$$\text{+ve SR} = \frac{I_o}{C_L} (V_{ds})$$

$$\text{-ve SR} = -\frac{I_o}{C_L} (V_{ds})$$



3/3/14

## Lec 21

\* DC performance - R load

\* DC gain - no DC error

$g_m$  ← affects  $\omega_n$

$g_o$  ← increased through cascade

$$1) r_{out} = (g_{m4} r_{ds4}) r_{ds2} \parallel (g_{m1} r_{ds1}) r_{ds8}$$

$$\text{DC gain} = g_{m1} r_{out} \Rightarrow \text{DC gain (1-stage op-amp)}$$

$$2) \omega_n = g_{m1}/C_L ; \omega_{pole} = \frac{1}{r_{out} C_L} = \frac{\omega_n}{C_L}$$

(same)

(much lower)

looks more like an ideal

$$3) V_{CMi} (\min) = V_{DSAT_0} + V_{as1}$$

$$V_{CMi} (\max) = V_{b1} - V_{as4} + V_{T_2}$$

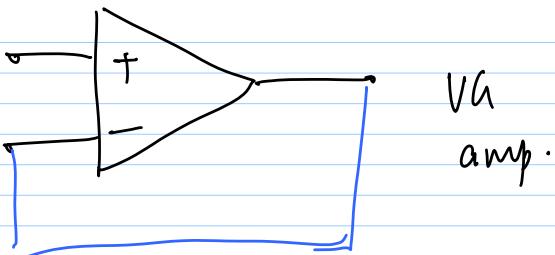
$$4) V_{CMo} (M:AX) = V_{OP} - V_{DSAT_7} - V_{DSAT_5}$$

$$V_{CMo} (MN) = V_{b1} - V_{T_4}$$

$$5) \text{Slow rate} = \pm \frac{I_o}{C_L}$$

6) Note: Almost the same @ low freq.

$$\frac{16kT}{3} \left[ \frac{1}{f_{MI}} + \frac{g_{m2}}{g_{m1}} \right]$$

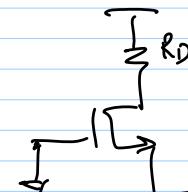


7) Offset

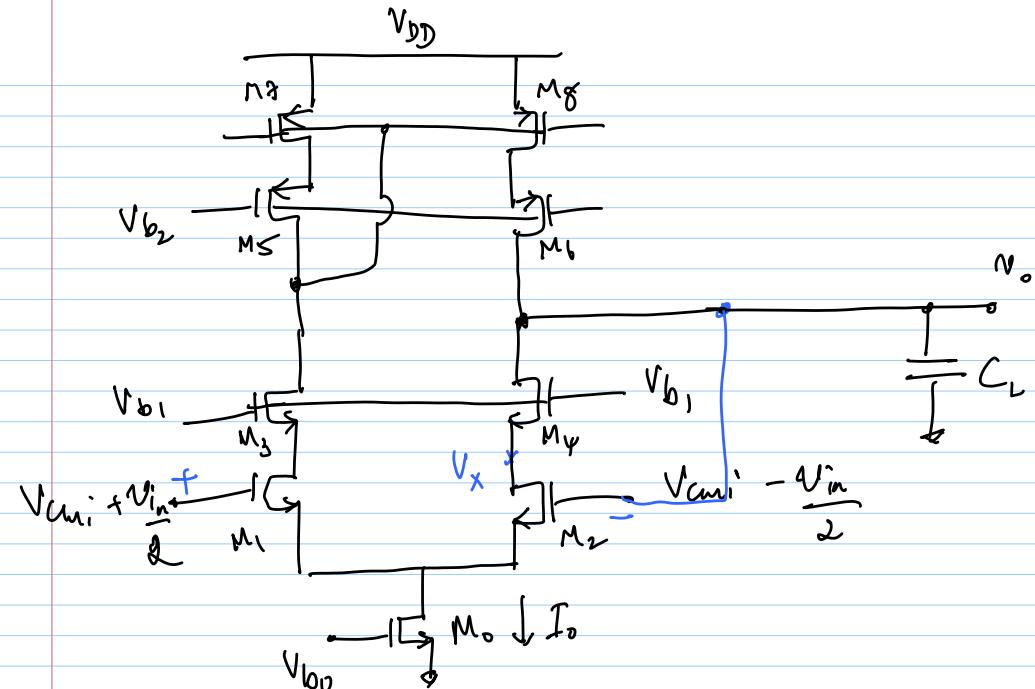
$r_{os}^2 = \text{same as before}$

$$r_{v_{T_{1,2}}}^2 + \left( \frac{g_{m2}}{g_{m1}} \right)^2 r_{v_{T_{2,3}}}^2$$

8) ND poles & zeros



$$R_{out} = \frac{1}{g_{m2}} \quad \text{only if } R_D \ll r_{ds}$$



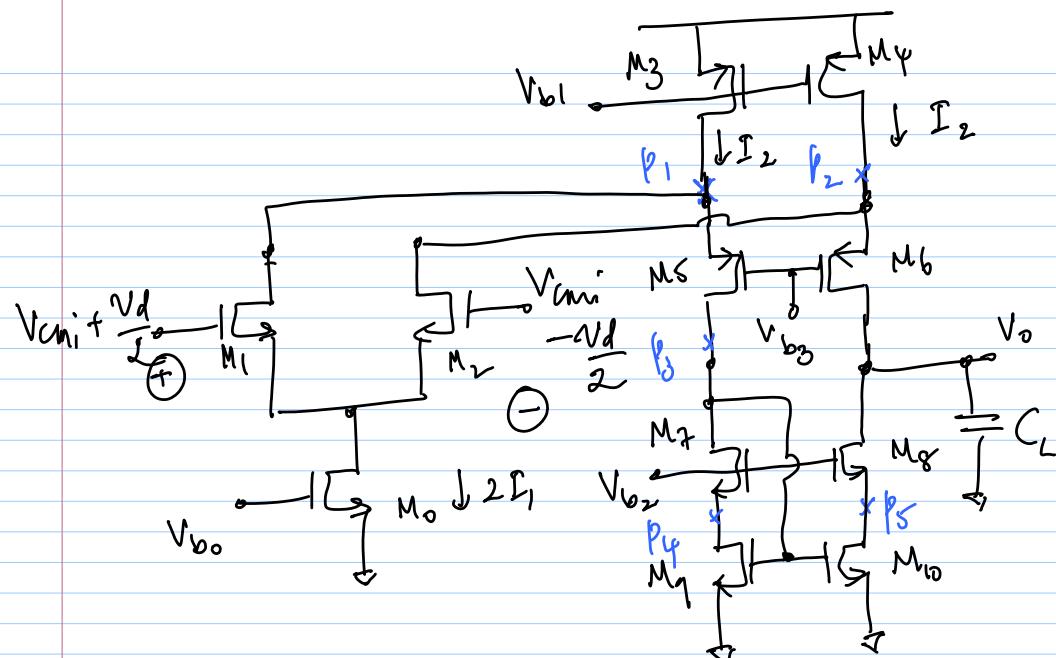
$$V_{o\min}(V_{cm\min}) = V_{b1} - V_{T4}$$

$$V_{o\max}(V_{cm\max}) = V_x + V_{T2} = V_{b1} - V_{asy} + V_{T2}$$

$$V_x = V_{b1} - V_{asy}$$

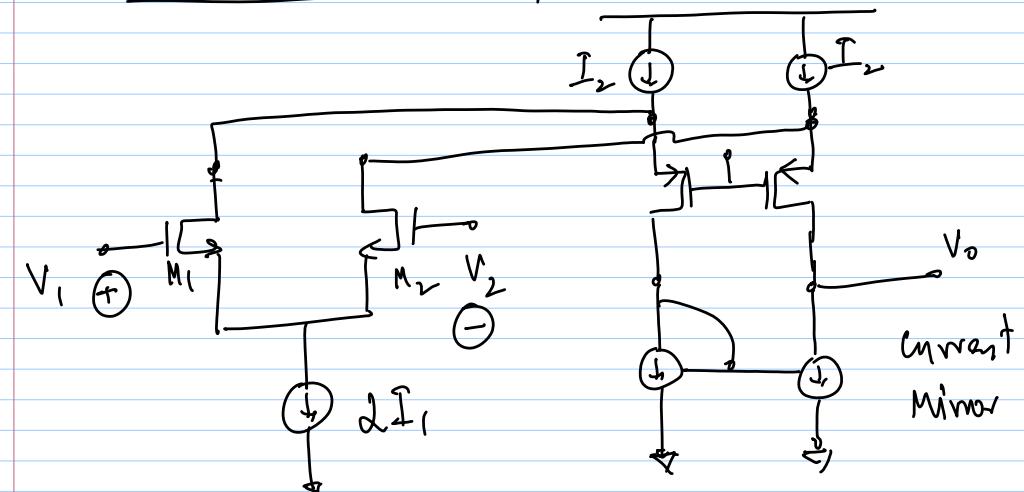
$$\Delta V_o = V_{T2} + V_{T4} - V_{asy}$$

$$\Delta V_o = V_{T2} - V_{DSAT_4} \sim \text{couple of } 100 \text{ mV}$$



4/3/14

## Lec 22 Folded Cascode Opamp



1) DC gain:

$$R_{out} = \left( g_m \cdot r_{ds} \right) r_{ds,10} \parallel \left( g_m \cdot r_{ds} \right) \left( r_{ds,2} \parallel r_{ds,4} \right)$$

slightly lesser than folded cascode opamp

$$\text{DC gain} = g_m \cdot R_{out}$$

$$2) \omega_n = \frac{g_m}{C_L}; \quad \omega_D = \frac{g_{out}}{C_L}$$

3) ND poles & zeroes  $\rightarrow p_1 - p_5$

$$4) V_{cmi}(\min) = V_{DSAT_2} + V_{as_1}$$

$$V_{cmi}(\max) = V_{b_3} + V_{shs} + V_{T_1}$$

$$5) V_{cmo}(\min) = V_{DSAT_8} + V_{DSAT_{10}}$$

$$V_{cmo}(\max) = V_{b_3} + V_{T_6}$$

6) Nout

$$\bar{e}_n^2 = \frac{16 k T}{\delta} \left[ \frac{1}{\hat{m}_1} + \frac{\hat{m}_3}{\hat{m}_1^2} + \frac{\hat{m}_1}{\hat{m}_1^2} \right]$$

7) Offset - Itw

8) Slew Rate