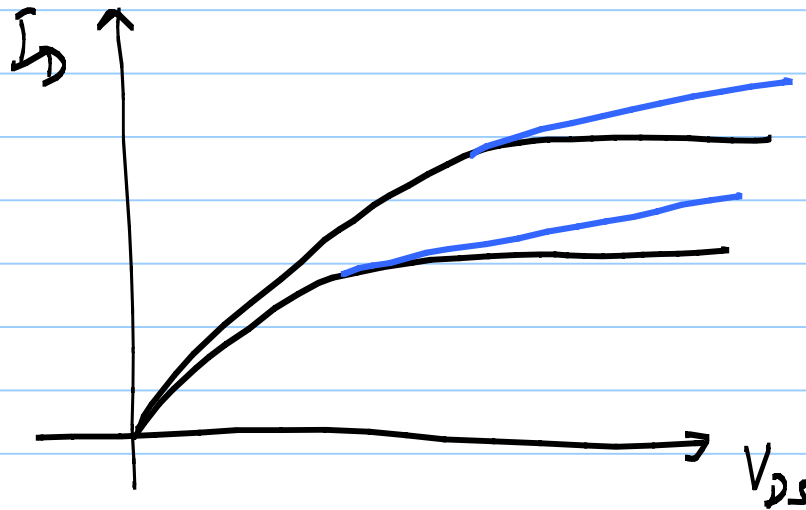


25/8/20

Lecture 12

Real MOSFET Characteristics



We assumed that in sat.,

$$I_D = f(V_{GS}) \text{ only}$$

In reality,

$$I_D = f(V_{GS}, V_{DS})$$

$$I_D = \frac{1}{2} \mu_n C_{ox} \left(\frac{W}{L}\right) (V_{GS} - V_T)^2 (1 + \lambda V_{DS})$$

for a good MOSFET, $\lambda \ll 1$

"Channel Length Modulation"

for op. pt. calculations, use: $I_D = \frac{1}{2} \mu_n C_{ox} \left(\frac{W}{L}\right) (V_{GS} - V_T)^2$

2-port parameters:

$$y_{11} = y_{12} = 0$$

$$y_{21} = g_m = \frac{\partial I_D}{\partial V_{gs}} = \mu_n C_{ox} \left(\frac{W}{L} \right) (V_{gs} - V_T) (1 + \lambda V_{ds})$$

$$\approx \mu_n C_{ox} \left(\frac{W}{L} \right) (V_{gs} - V_T) \quad \left\{ \begin{array}{l} \text{assume same} \\ \text{as before} \end{array} \right\}$$

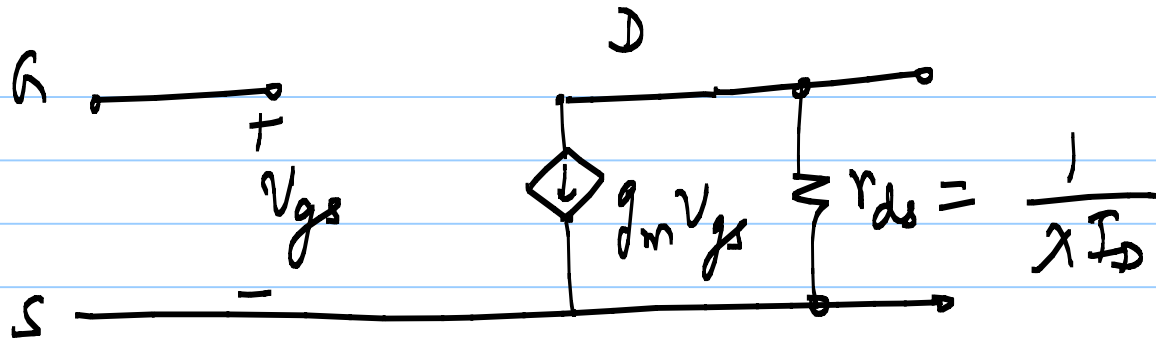
$$g_{ds} = y_{22} = \frac{\partial I_D}{\partial V_{ds}} = \underbrace{\frac{1}{2} \mu_n C_{ox} \left(\frac{W}{L} \right) (V_{gs} - V_T)^2}_{\approx I_D} \cdot \lambda$$

output
of

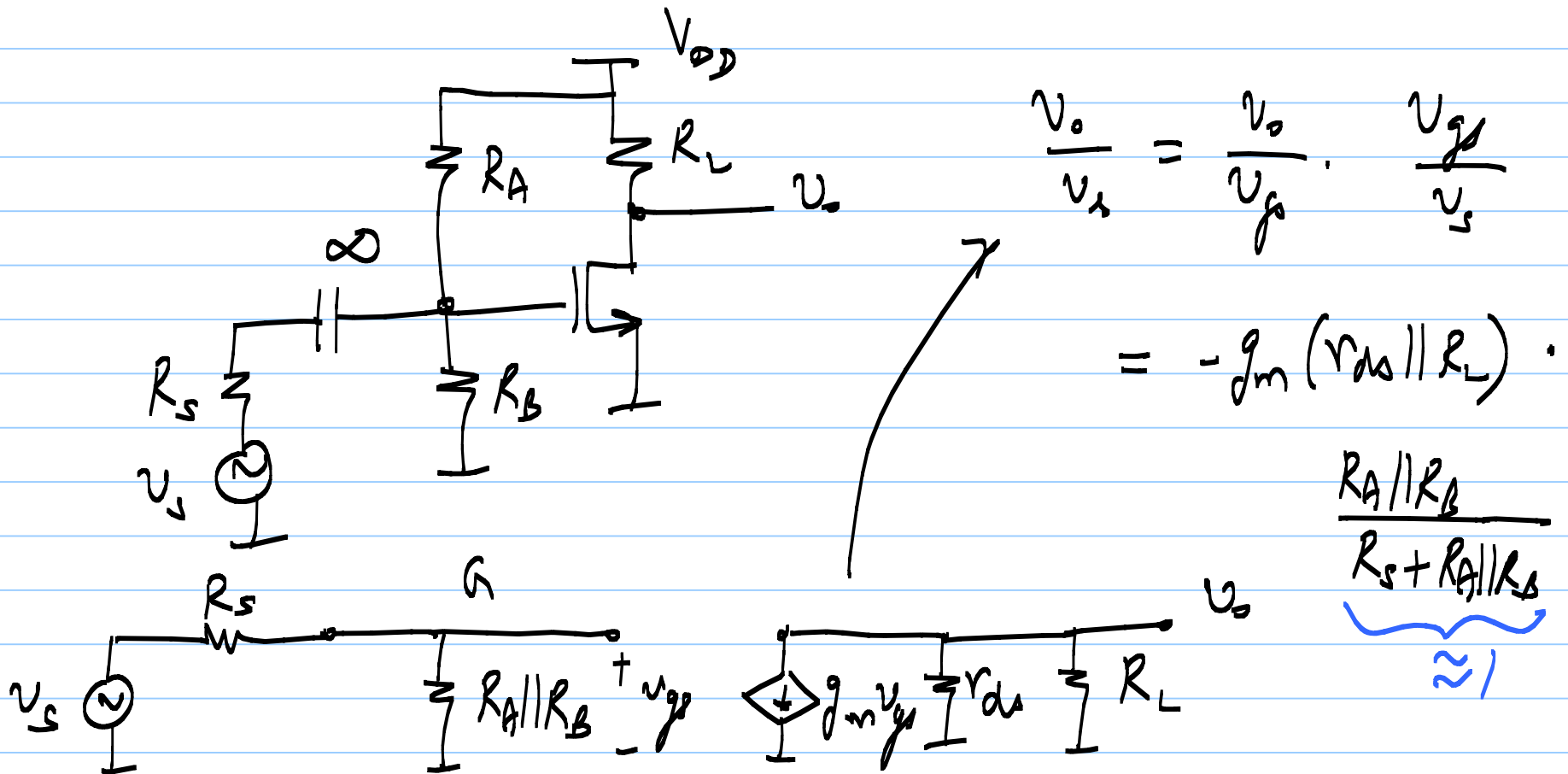
conductance
MOSFET

$$g_{ds} \approx \lambda \cdot I_D$$

$$r_{ds} = \frac{1}{g_{ds}} = \frac{1}{\lambda I_D}$$



Common-Source amp.

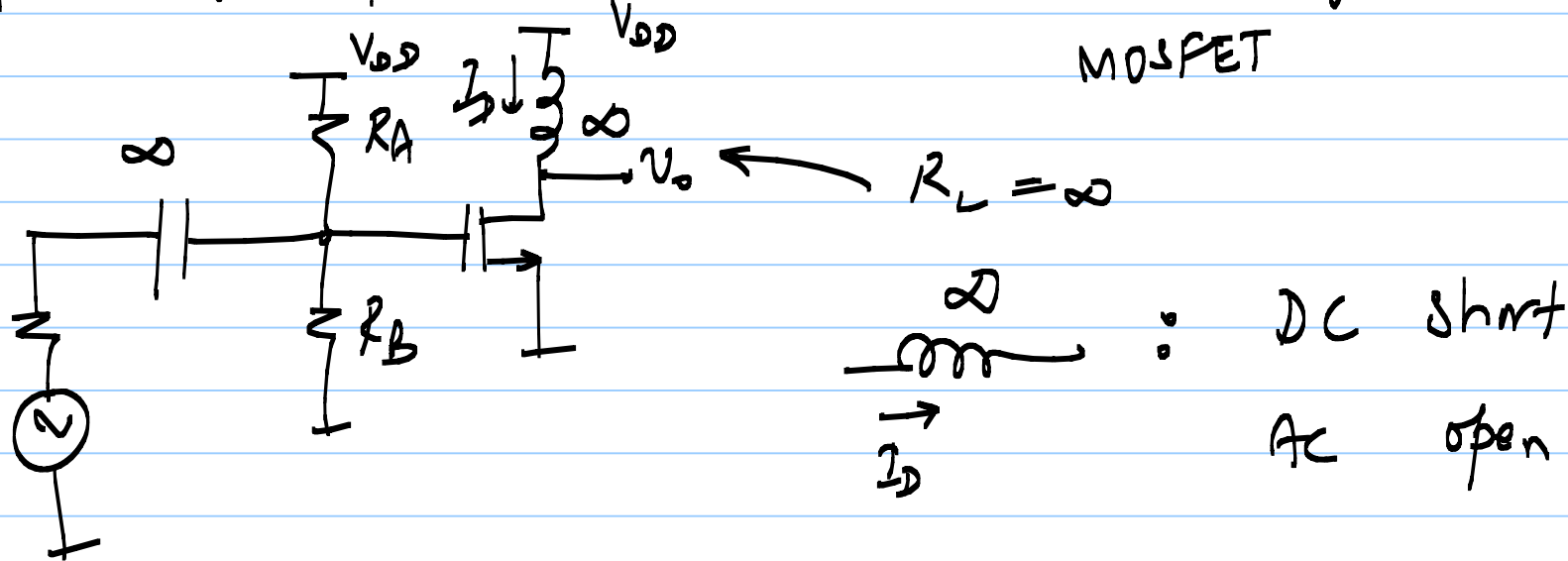


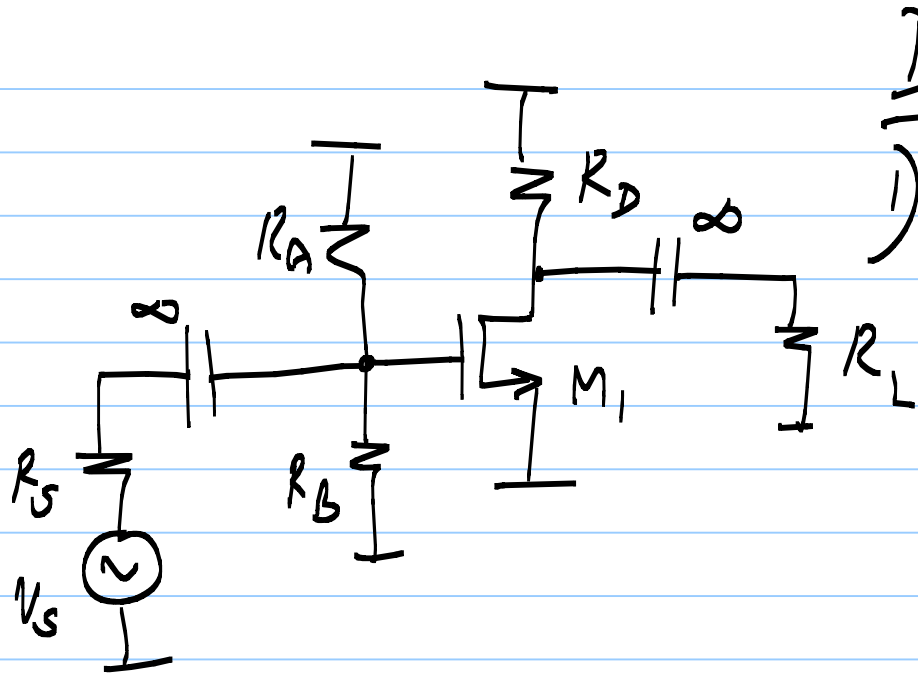
$$\frac{v_o}{v_s} \approx -g_m (r_{ds} \parallel R_L) \quad \text{gain is smaller}$$

Max possible gain : $R_L = \infty$

$$|\text{max. gain}| = g_m r_{ds} \equiv \text{"intrinsic gain" of}$$

MOSFET





Issues:

- 1) M_1 parameters vary with
 - a) ambient temp.
 - b) time
 - c) device to device

i.e. random variations in device

e.g. if V_T changes, $\xrightarrow{\text{properties}}$ I_D changes \rightarrow changes in $g_m, r_{ds}, V_{DS}, V_{ov}$ etc.

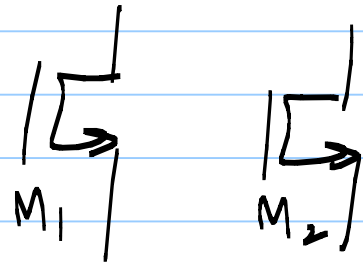
\Rightarrow Undesirable

2) Tolerance in R_A, R_B etc

On an IC :

R, C, MOSFET etc.

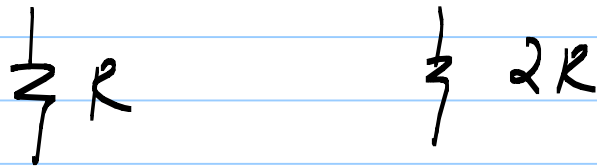
- 1) Multiple copies of the same device nominally have same properties, (on the same IC)



V_{T1}, V_{T2}
 μ_{n1}, μ_{n2}
etc. } vary together

- 2) Properties may vary across multiple ICs

- 3) Ratios of like components vary similarly



$\frac{R_B}{R_A + R_B}$ — used to generate V_{AS} from V_{DD}

Issues still exist:

1) V_{GS} is constant ($\propto V_{DD}$), but V_T etc. can vary $\Rightarrow I_D$ varies \Rightarrow gain, swing limits vary.

\Rightarrow Make V_{GS} vary with V_T to get desired I_D

Use "negative feedback" to generate V_{GS}

* Desired Value (desired bias drain current = I_D)

* Measure actual value (I_D)

* Compare actual with desired ($I_D \leftrightarrow I_0$)

* Use error to move actual value towards
desired value (make $I_D \rightarrow I_0$, using V_{as})