Monopoles and Dipoles

- Monopoles and dipoles are widely used antennas in wireless communications systems.
- Monopoles are particularly popular for portable units and on automobiles and other vehicles.
- In practice, wide use is made of the *quarter-wavelength monopole*.

Monopoles and Dipoles

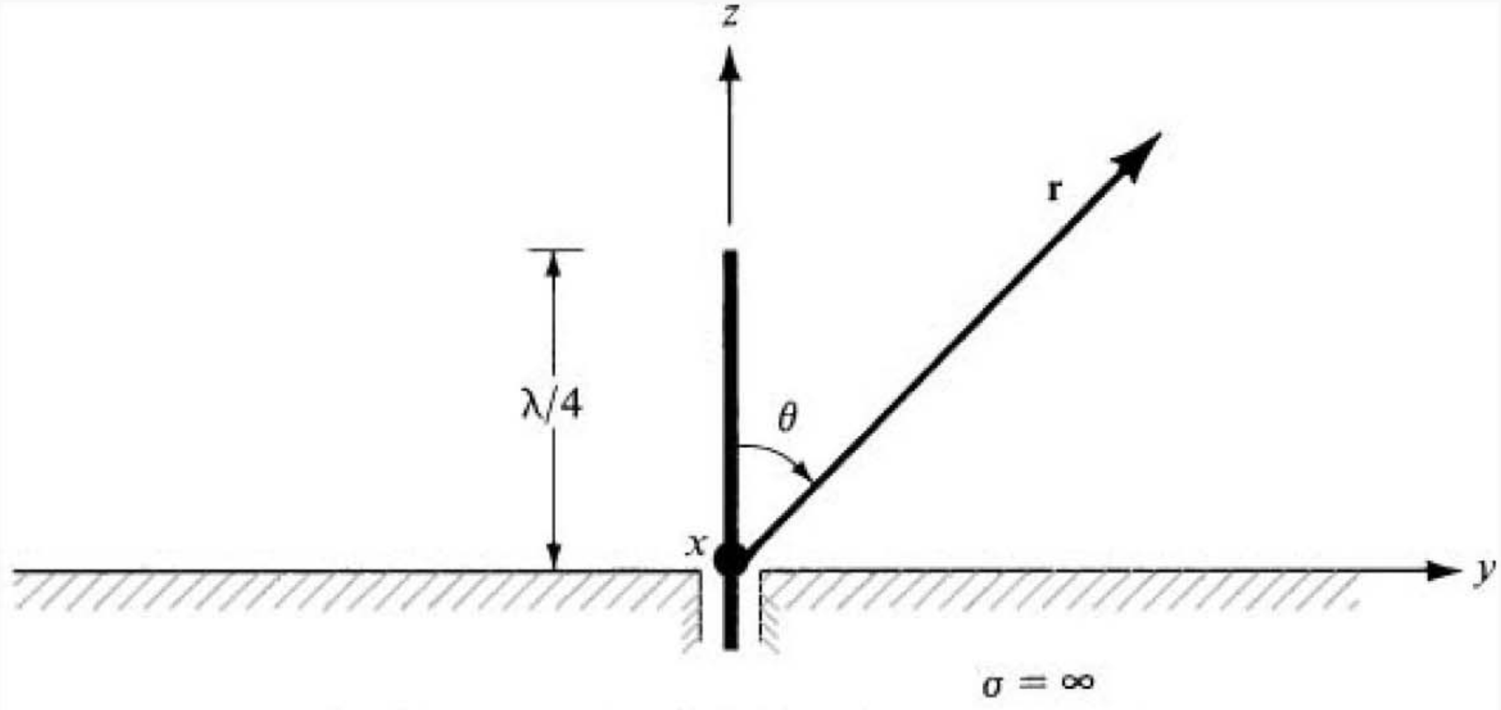


Fig. 4.19(a)

Equivalent of a λ/4 Monopole on a PEC

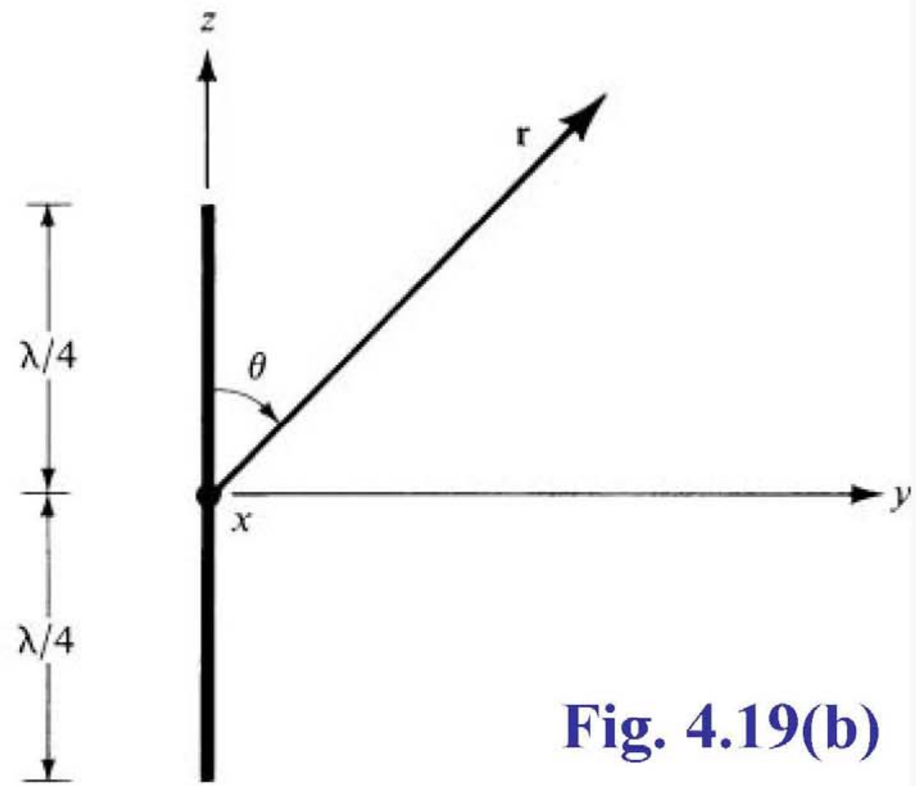


Fig. 4.19(b)

Half-Wavelength Dipole ($l = \lambda/2$)

$$E_\theta \simeq j\eta \frac{I_0 e^{-jkr}}{2\pi r}\left[\underbrace{\frac{\cos\left(\frac{\pi}{2}\cos\theta\right)}{\sin\theta}}_{Field\ Pattern}\right] \quad (4\text{-}84)$$

$$H_\phi \simeq j \frac{I_0 e^{-jkr}}{2\pi r}\left[\underbrace{\frac{\cos\left(\frac{\pi}{2}\cos\theta\right)}{\sin\theta}}_{Field\ Pattern}\right] = \frac{E_\theta}{\eta} \quad (4\text{-}85)$$

Directivity?

- Dipole in free space, and a monopole of half the length above a perfect ground plane will have the same radiation intensity in the top-half plane (since fields are identical).
- If the dipole emits power P, then the monopole emits power P/2, since the fields are zero in bottom hemisphere.
- Thus, $D_d = \frac{4\pi U_m}{P}$, whereas for the monopole, $D_m = \frac{4\pi U_m}{\frac{P}{2}} = 2D_d$

Radiation resistance?

- If the dipole emits power P, then the monopole emits power P/2, since the fields are zero in bottom hemisphere.

- Thus, $R_{rd} = 2P/I_0^2$, whereas for the monopole, $R_{rm} = 2(\frac{P}{2})/I_0^2 = R_{rd}/2$

Monopoles and Dipoles

$$\boxed{\begin{aligned} Z_{in}\ \text{(monopole)} &= \frac{1}{2} Z_{in}\ \text{(dipole)} \\ D_0\ \text{(monopole)} &= 2 D_0\ \text{(dipole)} \end{aligned}} \qquad (4\text{-}106)$$

Use Of Linear Monopole

Linear monopoles, especially $\lambda/4$, are used as transmitting/receiving elements for wireless mobile/cellular telephones.

Examples:

A. Civilian Band (CB) Radio
B. Cellular Telephone
C. Rooftop Automobile Antenna
D. Amateur Radio Antenna

Examples of Antennas on Cellular and Cordless Telephones, Walkie-Talkies, and CB Radios

Fig. 4.22

Triangular Array Of Linear Dipoles For Wireless Mobile Communication Base Stations

Fig. 4.23

Electric Conductor

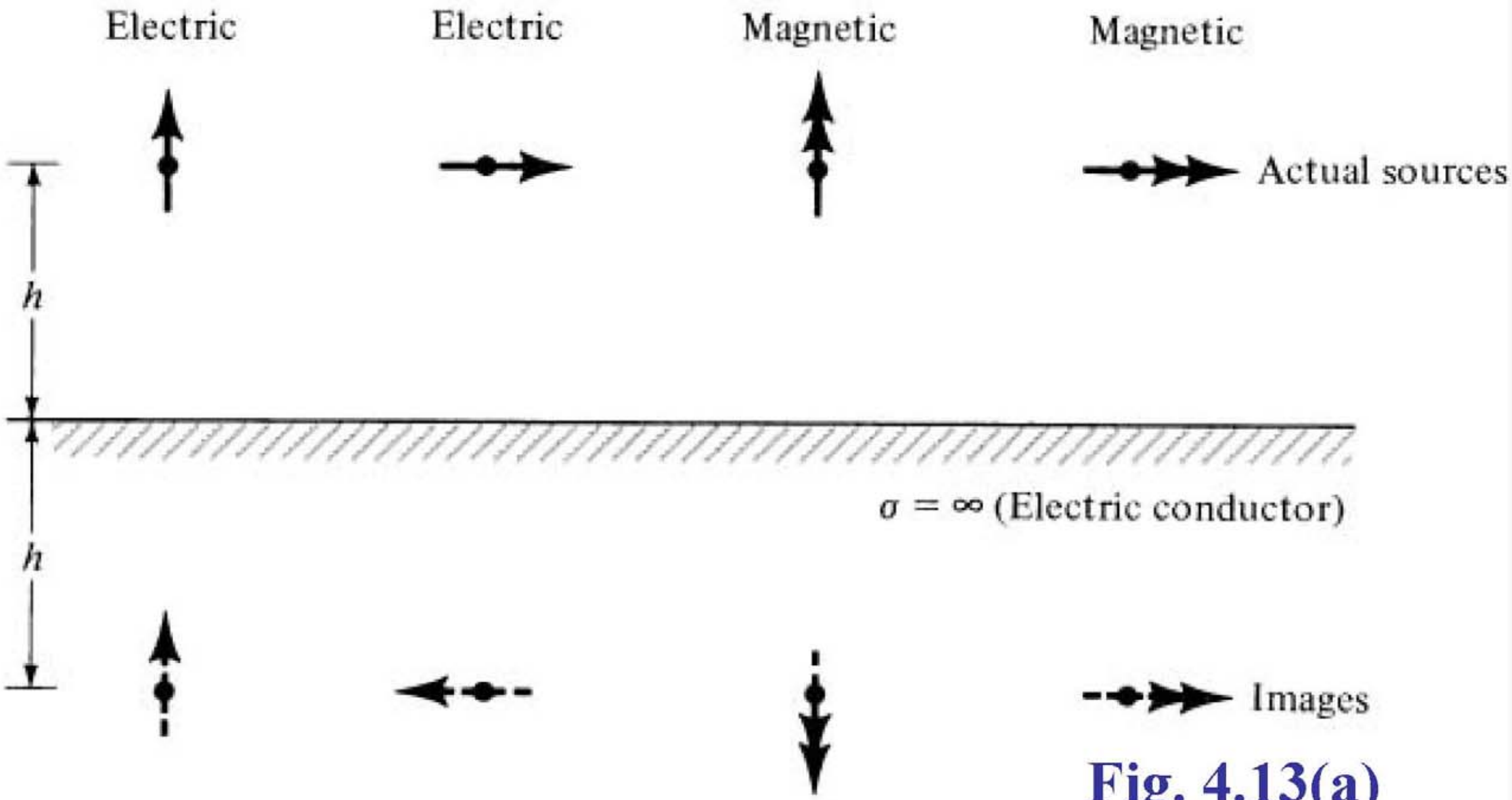


Fig. 4.13(a)

Horizontal Electric Dipole Above an Infinite Perfect Electric Conductor

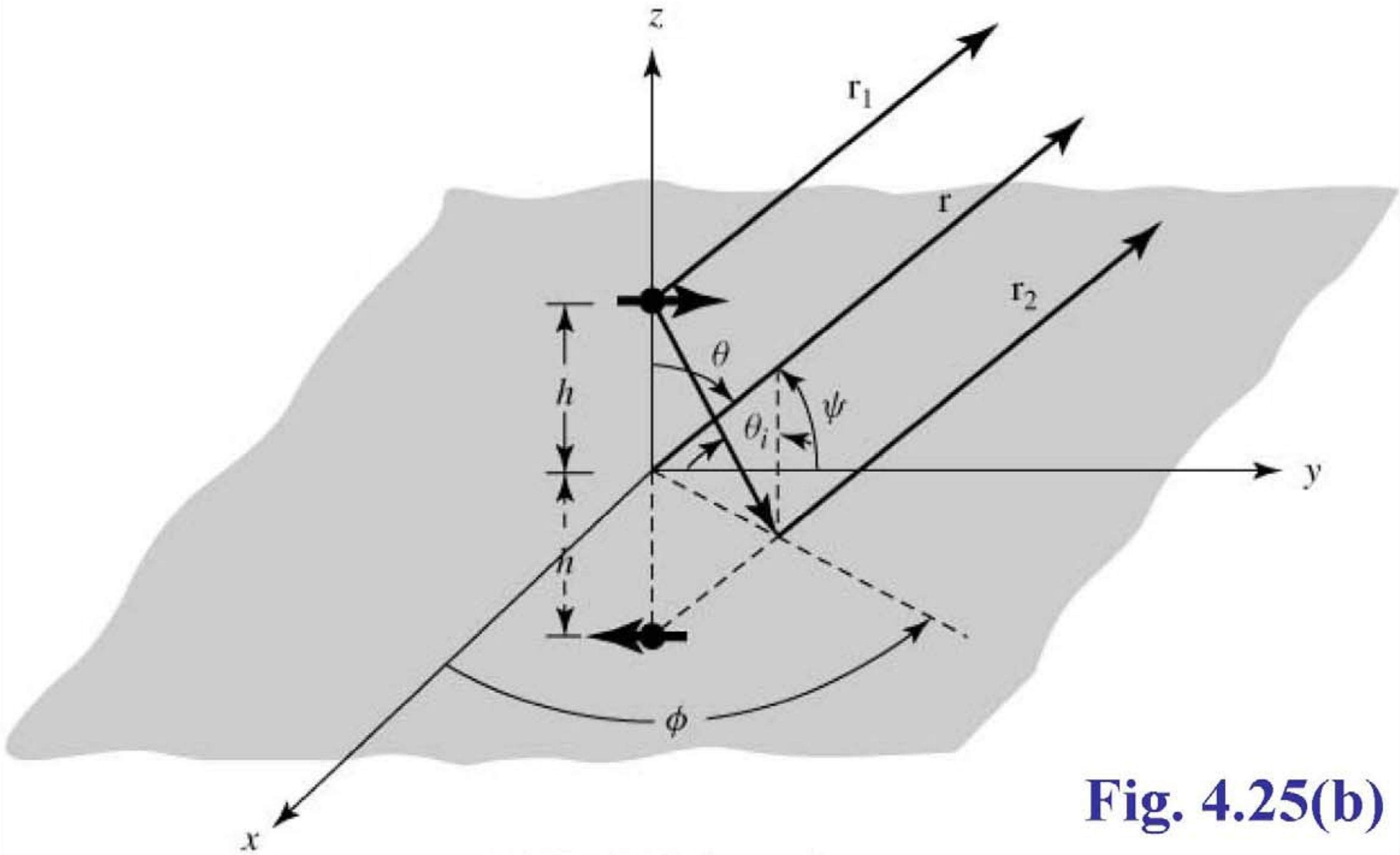


Fig. 4.25(b)

$$E_{\psi}^{d} = j\eta \frac{kI_o \ell e^{-jkr_1}}{4\pi r_1} \sin\psi \qquad \text{(4-111)}$$

$$E_{\psi}^{r} = jR_h \eta \frac{kI_o \ell e^{-jkr_2}}{4\pi r_2} \sin\psi \qquad \text{(4-112)}$$

$$= -j\eta \frac{kI_o \ell e^{-jkr_2}}{4\pi r_2} \sin\psi \qquad \text{(4-112a)}$$

Elevation Plane ($\phi = 90°$) Amplitude Patterns

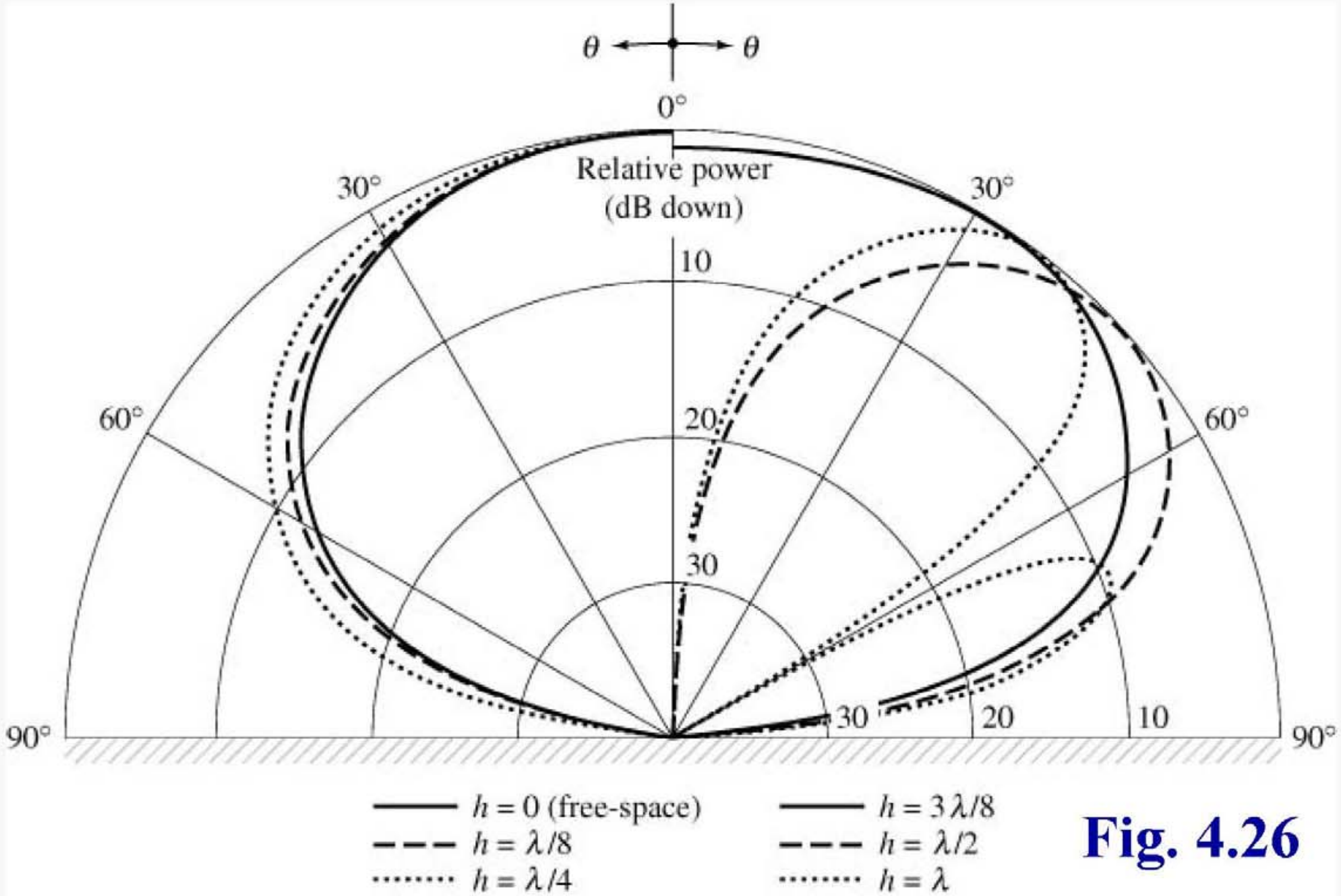


Fig. 4.26

3-D Amplitude Pattern of a Horizontal Dipole

$h = \lambda$

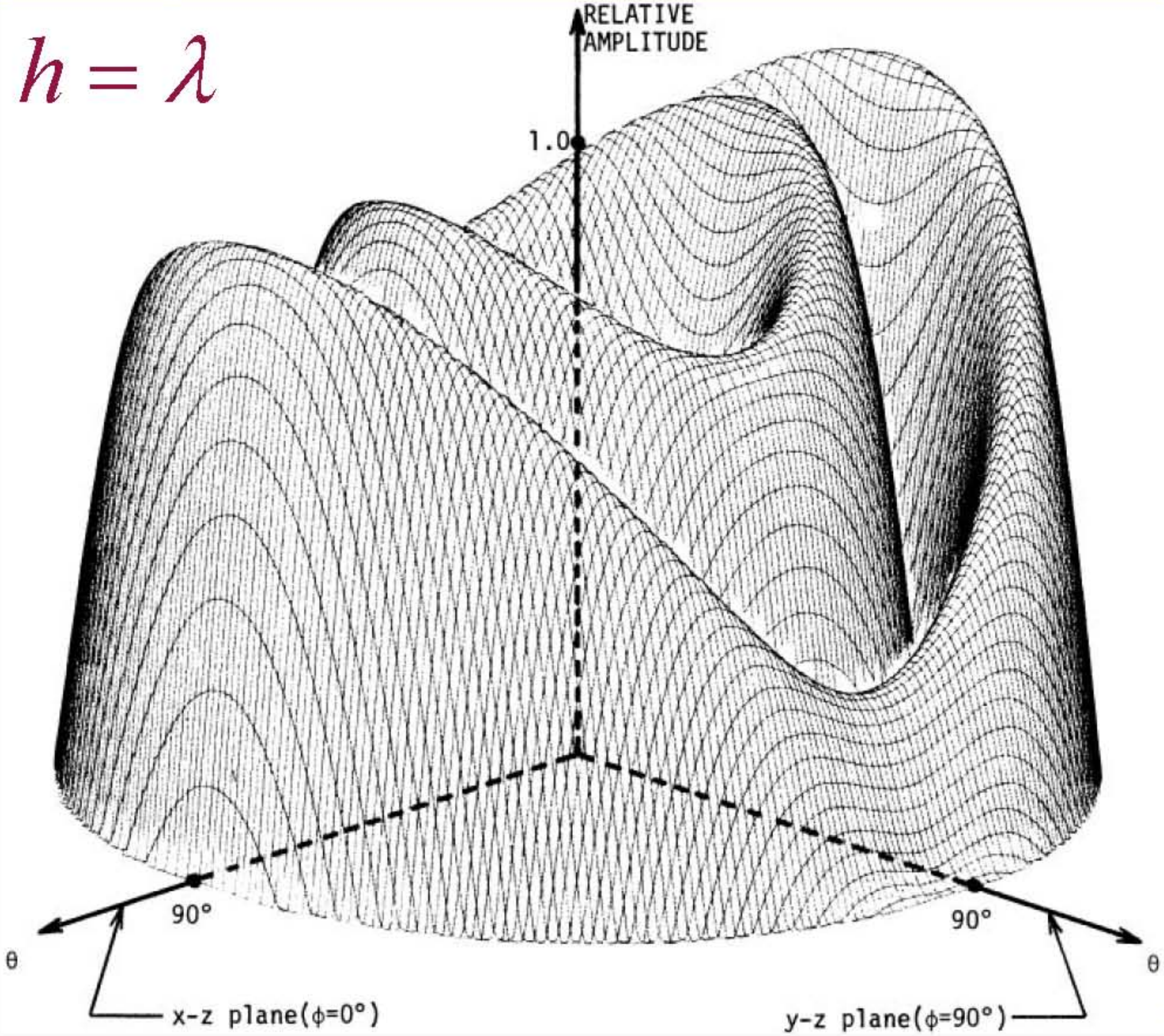


Fig. 4.27

Scalloping of Amplitude Pattern of Horizontal Dipole

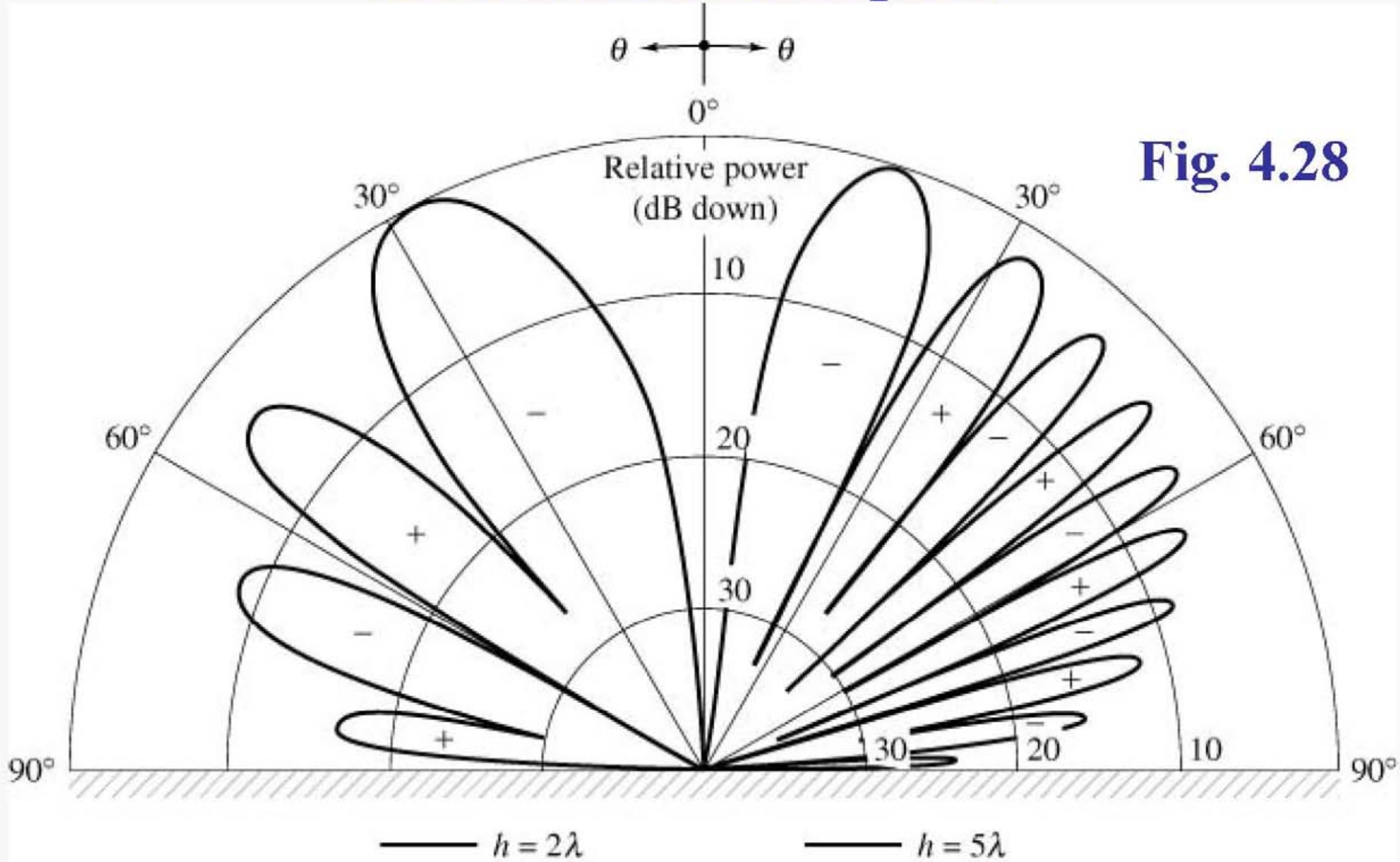


Fig. 4.28

Horizontal Dipole

$$\text{Number of Lobes} \cong 2\left(\frac{h}{\lambda}\right) \qquad (4\text{-}117)$$

$$h >> \lambda$$

Loss-y Earth

- For lower frequencies, (<3 MHz), PEC is not a good approximation.

- As engineering design, to improve radiation efficiency often metal plates are placed below antennas to simulate PEC planes.

Constitutive Parameters of Ground

$$\varepsilon_0, \mu_0, \sigma = 0$$

$$\varepsilon_1, \mu_1, \sigma_1$$

Constitutive Parameters of Ground

$$\varepsilon_1, \mu_1, \sigma_1$$

or

$$\dot{\varepsilon}_1, \mu_1$$

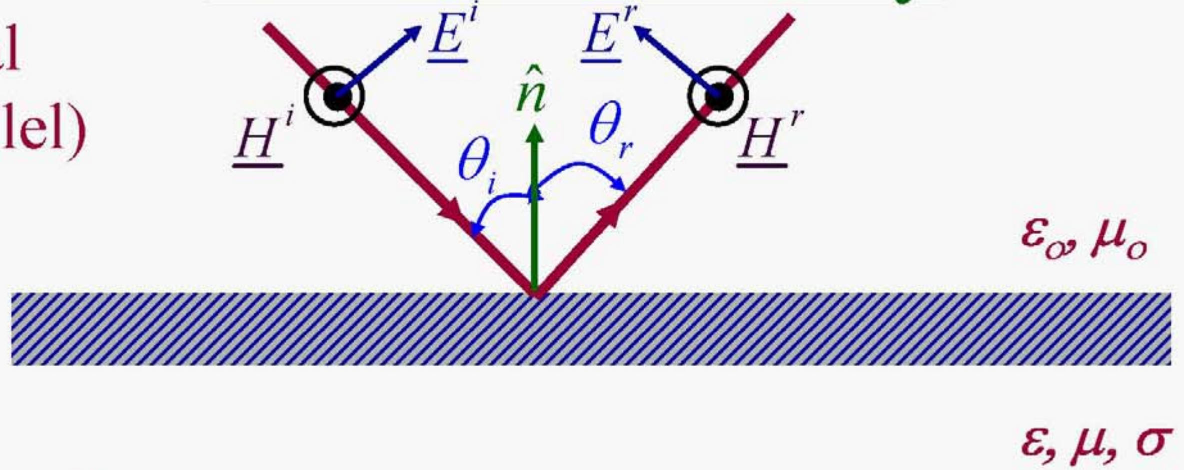
Could also include magnetic conductivity, if necessary.

where

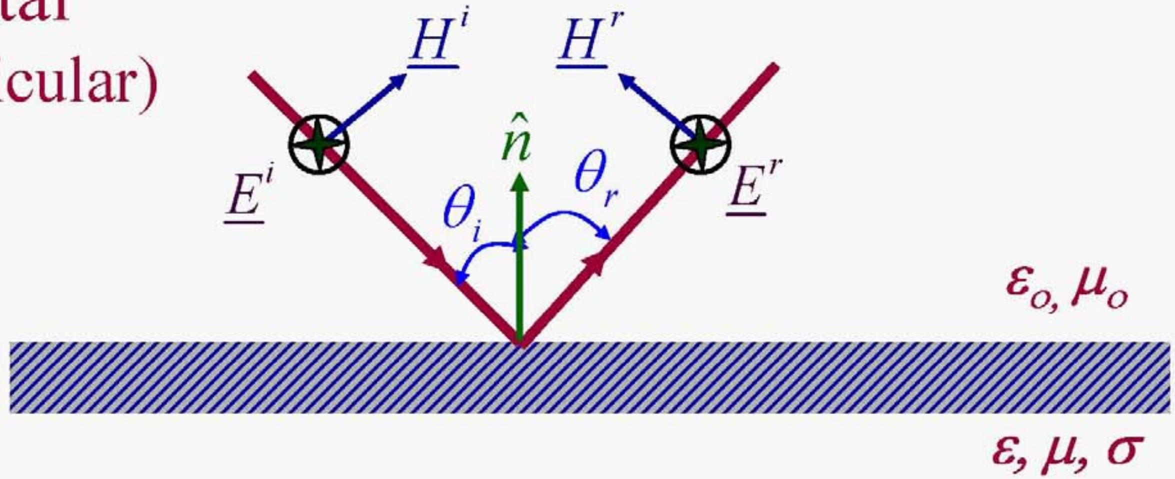
$$\dot{\varepsilon}_1 = \varepsilon_1' - j\varepsilon_1'' = \varepsilon_1 - j\frac{\sigma_1}{\omega}$$

Interface Geometry

Vertical (Parallel)



Horizontal (Perpendicular)



Vertical Polarization

$$E_\theta = j\eta \frac{kI_o \ell e^{-jkr}}{4\pi r} \sin\theta \left[e^{jkh\cos\theta} + R_v e^{-jkh\cos\theta} \right]$$

$$R_v = \frac{\eta_0 \cos\theta_i - \eta_1 \cos\theta_t}{\eta_0 \cos\theta_i + \eta_1 \cos\theta_t} = -R_\parallel \qquad (4\text{-}125)$$

$$\eta_1 = \sqrt{\frac{j\omega\mu_1}{\sigma_1 + j\omega\varepsilon_1}}, \qquad \eta_0 = \sqrt{\frac{\mu_o}{\varepsilon_o}}$$

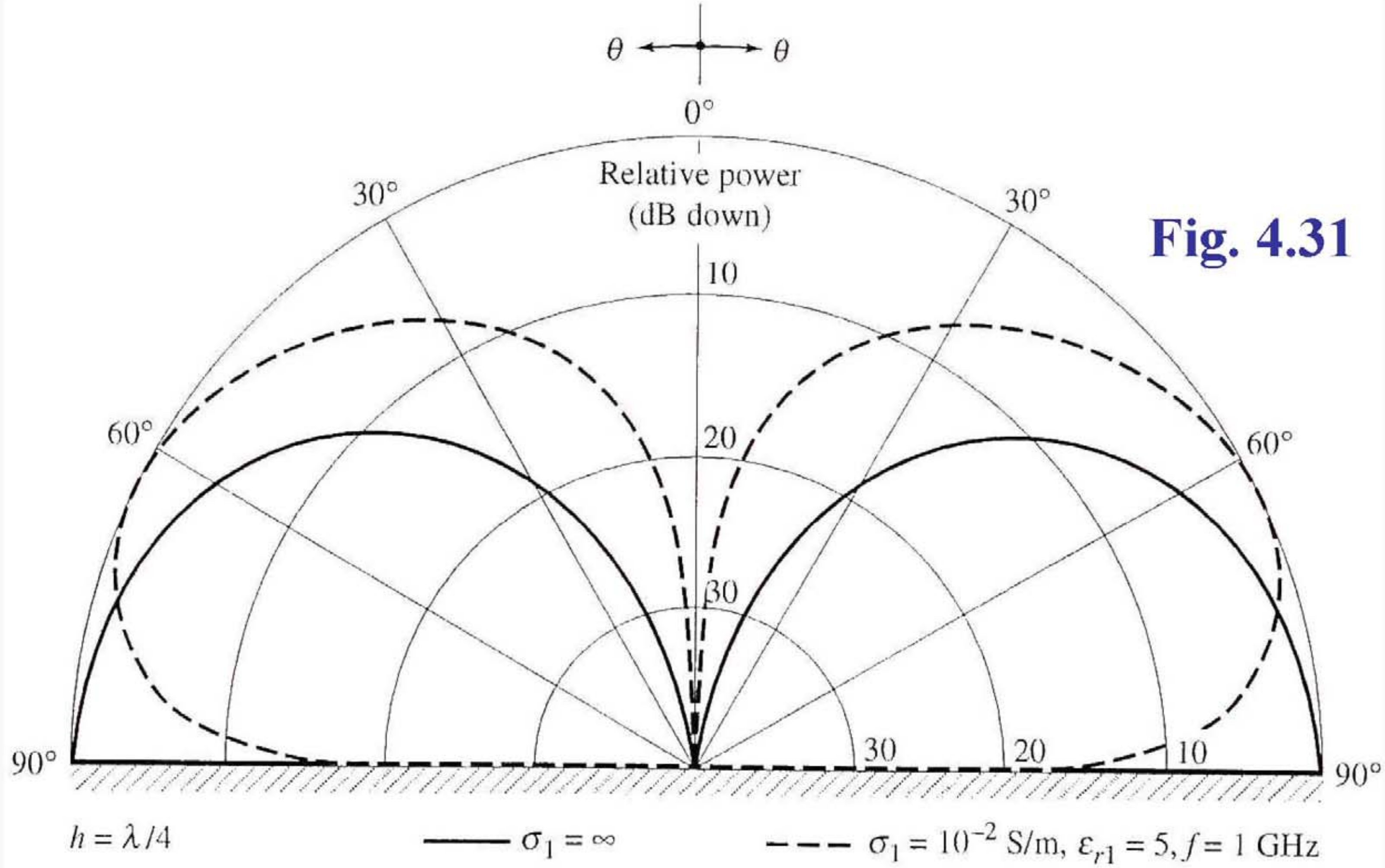
$$\gamma_0 \sin\theta_i = \gamma_1 \sin\theta_t \qquad (4\text{-}126)$$

$$\gamma_0 = j\beta_0$$

$$\gamma_1 = \sqrt{j\omega\mu_1 \left(\sigma_1 + j\omega\varepsilon_1\right)} = \alpha_1 + j\beta_1$$

Vertical Polarization

Fig. 4.31



Horizontal Polarization

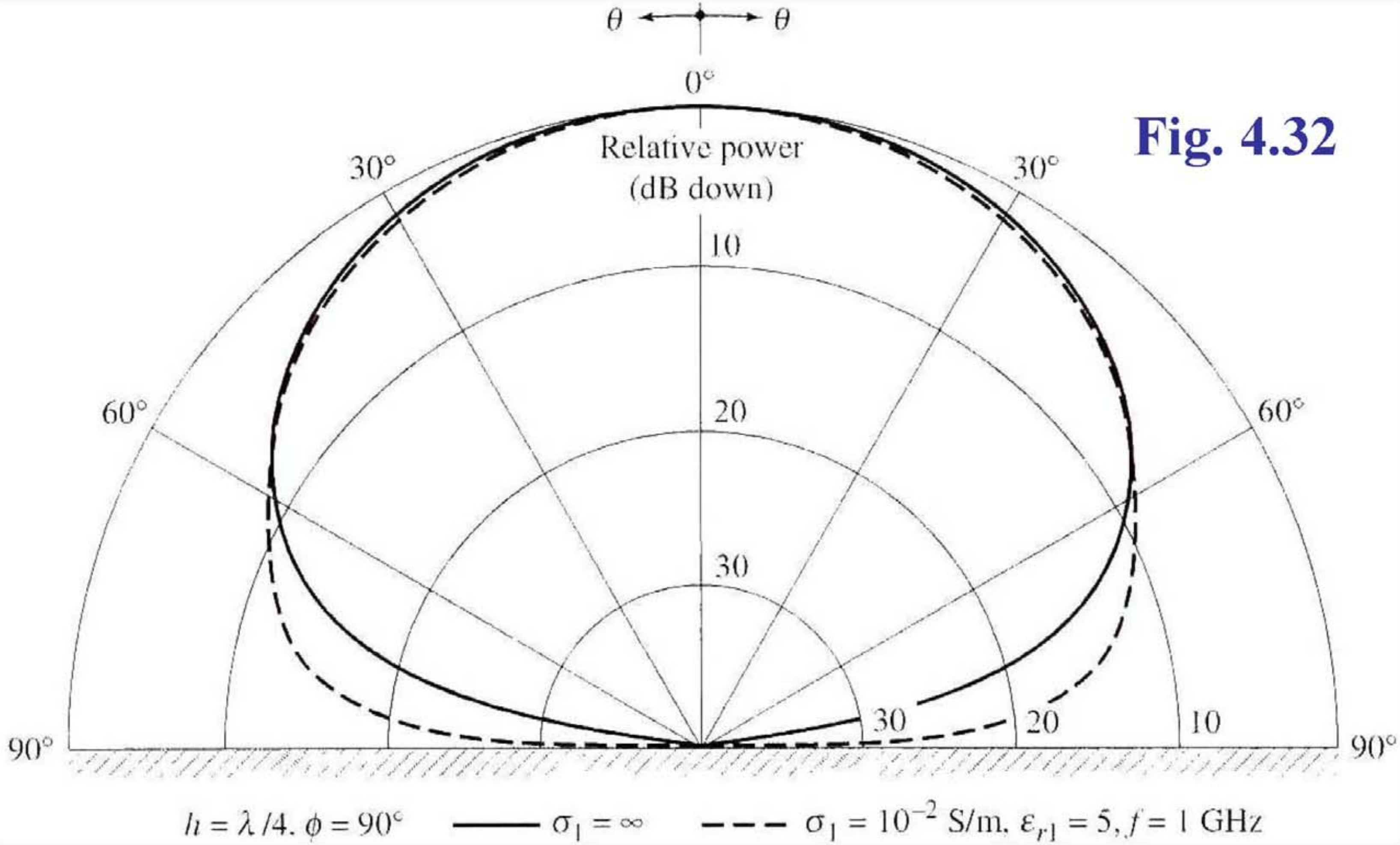


Fig. 4.32