## EE611 Problem Set 4

- 1. Suppose that the transmitter signal pulse g(t) has duration T and unit energy and the received signal pulse is h(t) = g(t) + ag(t T). Determine the equivalent discrete-time filter model for the sampled matched filter output.
- 2. The equivalent discrete-time model for an ISI channel is given by

$$v_k = I_k + 0.5I_{k-1} + \eta_k$$

where  $\{I_k\}$  is the transmitted information sequence and  $\{\eta_k\}$  is AWGN. In the absence of noise, sketch a constellation for the received signal if the transmitted signal constellation is:

- (a) a binary constellation: +1 and -1.
- (b) a 4-point constellation: (1, 1), (1, -1), (-1, 1), and -1, -1).

Repeat the above problem for the discrete-time channel given by

$$v_k = I_k + I_{k-1} + \eta_k$$

3. The equivalent discrete-time model for an ISI channel is given by

$$v_k = I_k + 0.5I_{k-1} + 0.25I_{k-2} + \eta_k,$$

where  $\{I_k\}$  is the transmitted information sequence and  $\{\eta_k\}$  is AWGN. Binary signalling  $(\pm 1)$  is used.

- (a) Sketch the trellis diagram representing the channel. Label each state transition (branch) with the corresponding input and output.
- (b) Consider two paths in the trellis corresponding to the state sequences  $\{A_k\}_{k=1}^D$  and  $\{B_k\}_{k=1}^D$ . If  $A_l = B_l$  and  $A_{l+1} \neq B_{l+1}$  for some l, then show that  $A_{l+2} \neq B_{l+2}$ .
- (c) Based on the answer to part (b), determine the minimum possible length for an error event for the above trellis.
- (d) Consider the error event described by the following 2 paths (sequence of states):

Path A: 
$$\{(1,1), (1,1), (1,1), (1,1)\}$$

and

Path B: 
$$\{(1,1), (-1,1), (1,-1), (1,1)\}$$

Given that the all 1's sequence is transmitted, determine the probability that the likelihood metric for path B is larger than the likelihood metric for path A.

(e) What is the minimum possible length for an error event for the following channel:

$$v_k = \sum_{n=0}^{L} f_n I_{k-n} + \eta_k,$$

4. Binary PAM is used to transmit information over a linear filter channel. The discrete-time model for the matched filter output is

$$y_k = 0.5I_{k-1} + 1.25I_k + 0.5I_{k+1} + \nu_k.$$

(a) Design a three-tap linear equalizer so that the output of the equalizer in the absence of noise is

$$\hat{I}_{k} = \sum_{n=-2}^{2} q_{n} I_{k-n},$$
$$q_{m} = \begin{cases} 1 & (m=0) \\ 0 & (m=\pm 1) \end{cases}$$

- (b) Determine  $q_m$  for  $m = \pm 2$ .
- (c) Design a 5-tap equalizer so that

$$\hat{I}_k = \sum_{n=-3}^3 q_n I_{k-n},$$

where

where

$$q_m = \begin{cases} 1 & (m=0) \\ 0 & (m=\pm 1, \ \pm 2) \end{cases}$$

- 5. (a) Assuming that the variance of noise at the output of the matched filter is 0.125, design 3-tap and 5-tap linear MMSE equalizers for the channel in the previous problem.
  - (b) Determine  $q_m$ , by convolving the impulse response of the equalizer with the channel response.