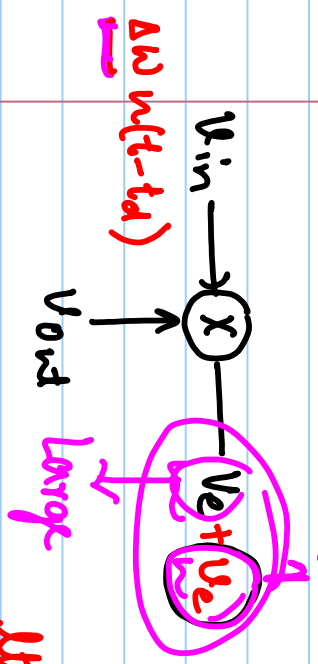


Lecture # 11

Note Title

small

09-06-2020

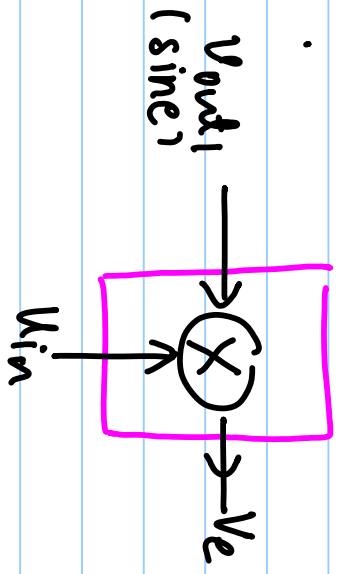


at $t=0$, $\Delta\omega(t) \rightarrow \Phi_e \neq 0, V_e \neq 0$

$$V_e = \frac{1}{2} \left[\sin(\Phi_e) + \sin(2\omega t + \Phi_e) \right]$$

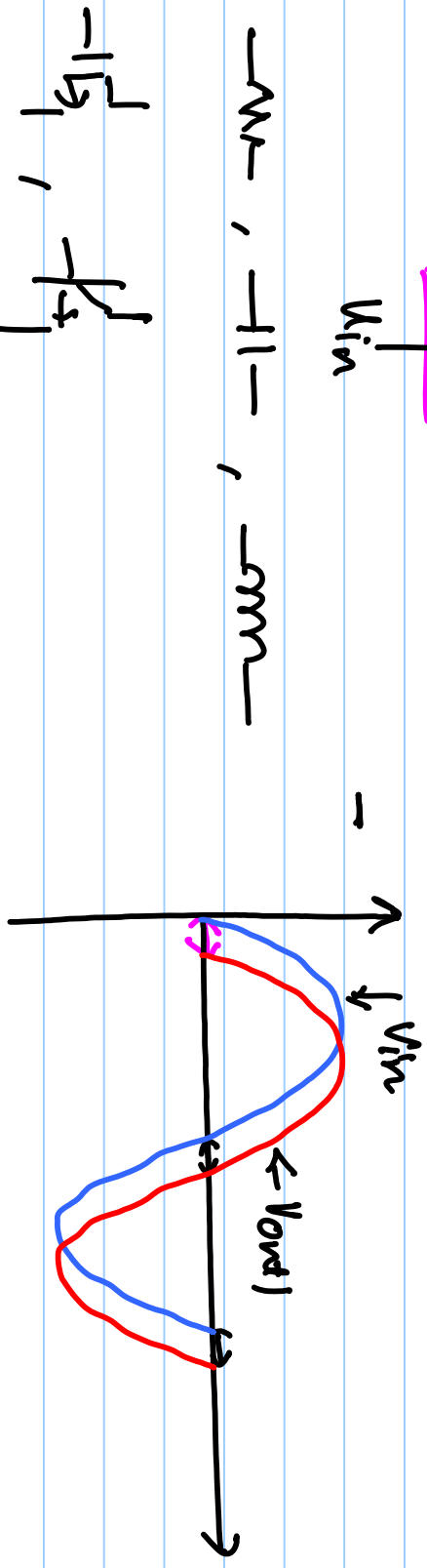
$$\text{At } V_e = \frac{1}{2} \sin(\Phi_e) \quad \checkmark$$

$$|V_e| \leq \frac{1}{2}$$



In PLL, we will have two periodic signals.

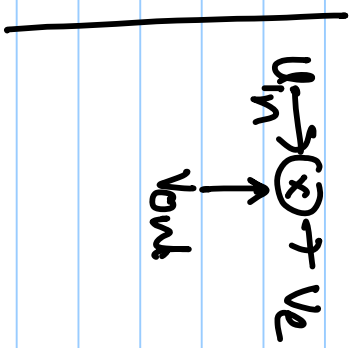
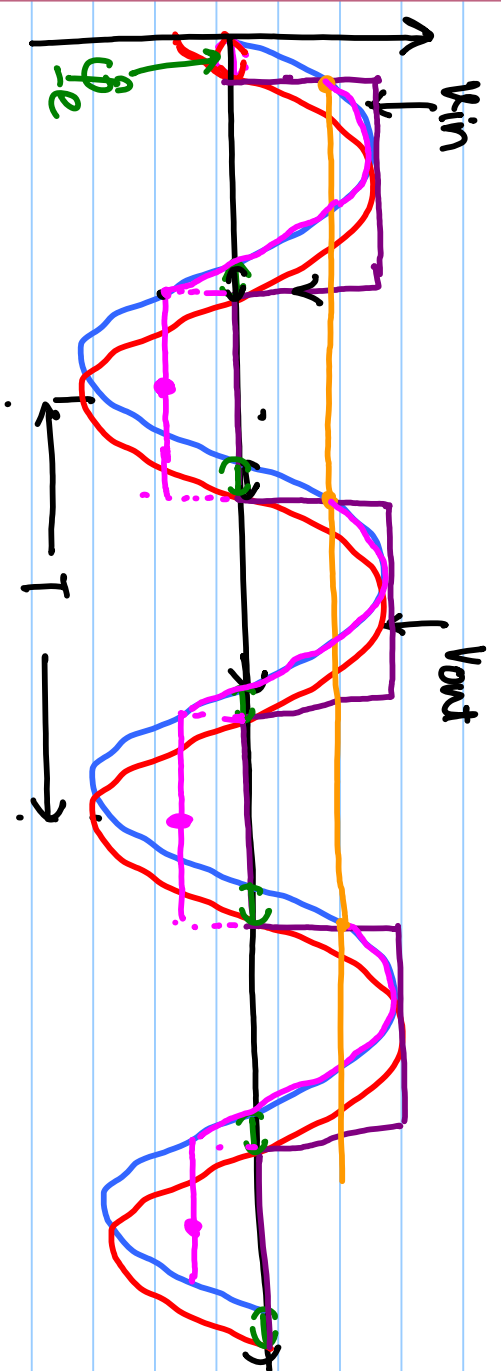
input (Ref). Output (out)



freq. U_{in} , V_{out} @ 100 MHz

@ 1 GHz.

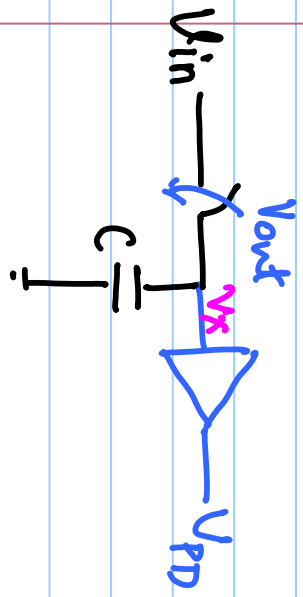
Sample & Hold PD



$$V_e = \frac{1}{2} \sin(\phi_e)$$

$$\phi_e = 2\pi \cdot \frac{\Delta t}{T}$$

$$V_e = \frac{1}{2} \sin\left(2\pi \cdot \frac{\Delta t}{T}\right)$$



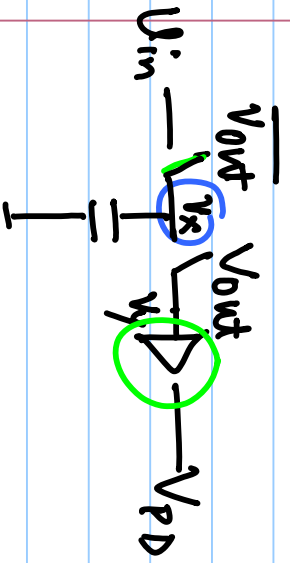
$$V_{PD} = A_{in} \sin(\omega_{in} \cdot (nT)) \quad \checkmark$$

$$= A_{in} \sin\left(2\pi f_{in} \cdot \left(\frac{n}{f_{out}}\right)\right)$$

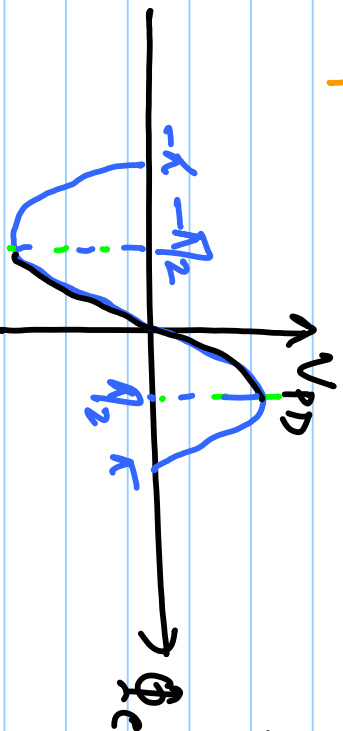
When $V_{out} = 1$, switch is closed

$V_{out} = 0$ switch is open.

$$\begin{aligned}
 V_{PD} &= A_{in} \sin \left(\omega_{in} \left(nT + \frac{T}{2} + \Delta t \right) \right) \\
 &= A_{in} \sin \left(2\pi \left(f_{in} \cdot T \right) n + \underbrace{2\pi f_{in} \cdot T}_{= 2\pi\kappa} + 2\pi \cdot \underbrace{f_{in} \cdot \Delta t}_{\frac{2\pi}{T}} \right) \\
 &= A_{in} \sin \left(\kappa + 2\pi \cdot \frac{\Delta t}{T} \right) \\
 &= A_{in} \sin \left(\kappa + \phi_e \right) \\
 \therefore A_{in} \sin \left(\phi_e \right)
 \end{aligned}
 \left| \begin{array}{l} f_{in} \cdot T = 1 \end{array} \right.$$

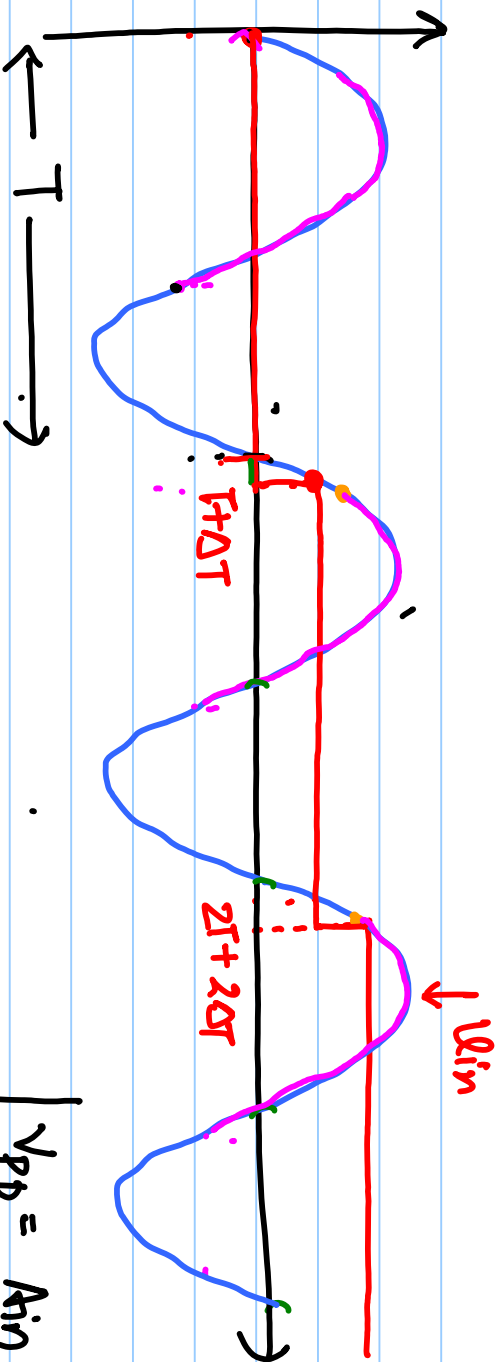
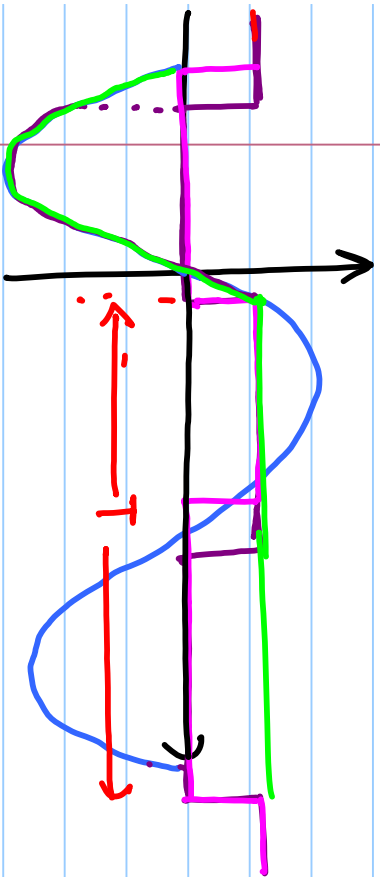


$$\begin{aligned}
 V_{PD} &= A_{in} \sin \left(\omega_{in} \cdot (\Delta t) \right) \\
 &= A_{in} \sin \left(2\pi \cdot \frac{\Delta t}{T} \right) \\
 &= A_{in} \sin \left(\phi_e \right)
 \end{aligned}$$



- Monotonic Range
 $\phi_e \in \left[\frac{-\pi}{2} \text{ to } \frac{\pi}{2} \right]$

- $K_{PD} = \frac{dV_{PD}}{d\phi_e} = A_{in} \cos(\phi_e)$



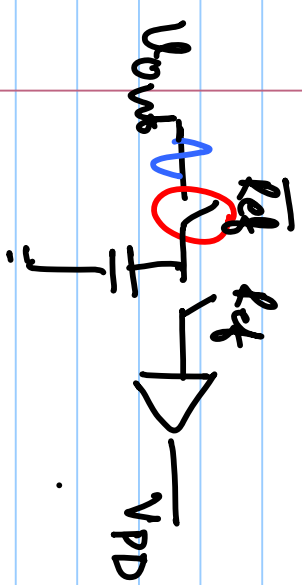
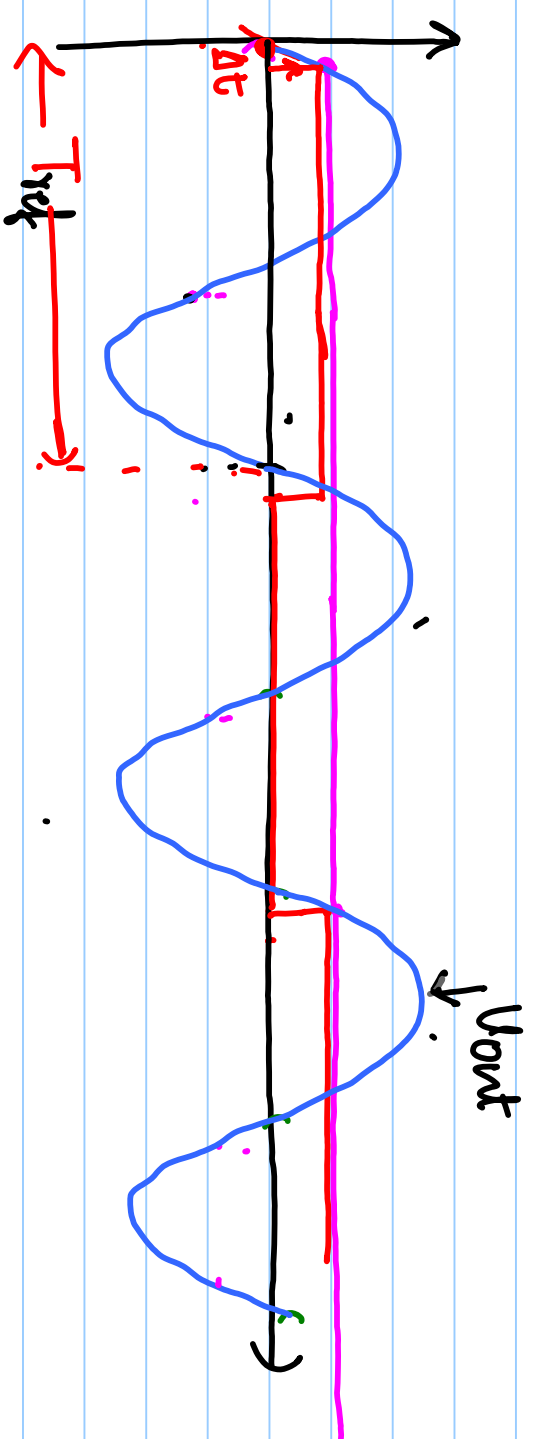
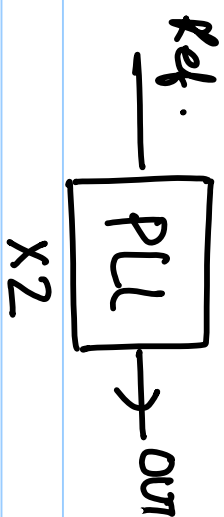
for N_{out} : time period be $T + \Delta T$

$$y_{pb} = A_{in} \sin(\omega_{in} \tau(T + \Delta T))$$

$$= A_{in} \sin\left(\frac{2\pi\tau}{T}(T + \Delta T)\right)$$

$$= A_{in} \sin\left(2\pi\left(\frac{\tau\Delta T}{T}\right)\right)$$

Sub-sampled PLLs.



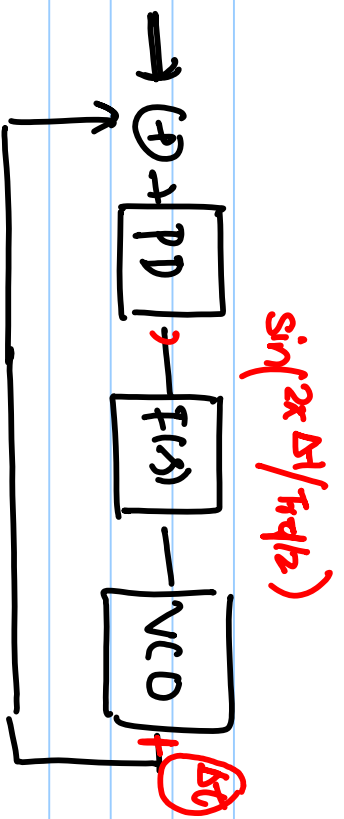
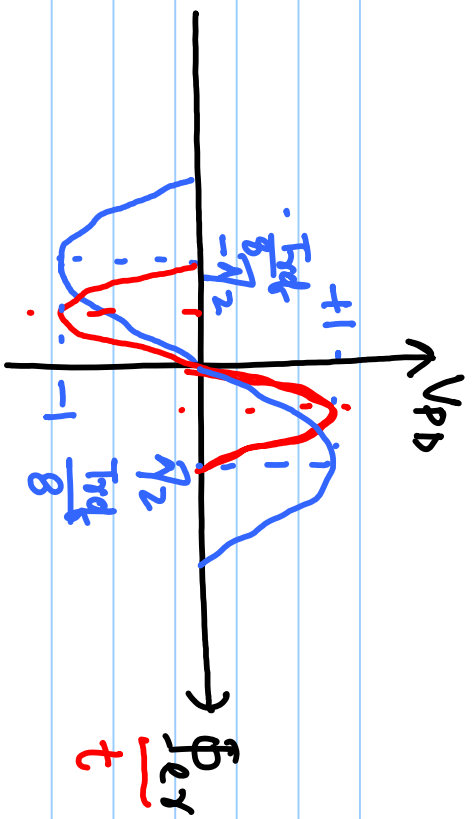
$$V_{PD} = A_{out} \cdot \sin(2\pi f_{out} \cdot (nT_{ref} + \Delta b))$$

$$= A_{out} \sin\left(2\pi \left(\underbrace{n \cdot f_{out}}_{f_{ref}}\right) + 2\pi \underbrace{\Delta b}_{T_{out}}\right)$$

$$= A_{out} \sin\left(2\pi \cdot \frac{\Delta b}{T_{out}}\right)$$

$$\left[V_{PD} \stackrel{=}{=} A_{out} \sin\left(\underbrace{\Phi_{er}}_{f_{ref}}\right) \right]$$

Φ_{er} : w.r.t output freq.



if $f_{out} = 2f_{ref} \Rightarrow \frac{\pi}{2} \Rightarrow \frac{T_{out}}{4}$

$$\phi_{err} = \pi/2 = 2\pi \cdot \frac{\Delta t}{T_{out}}$$

$$\Delta t = \frac{T_{out}}{4} = \frac{T_{ref}/2}{4} = \frac{T_{ref}}{8}$$

$$f_{out} = 4f_{ref} \Rightarrow \frac{\pi}{2} \rightarrow \frac{T_{out}}{4} = \frac{T_{ref}/4}{4} = \frac{T_{ref}}{16}$$

Case # 1: $f_{out} = 2f_{ref}$
 $V_{pd} = \sin\left(2\pi \cdot \frac{\Delta t}{T_{ref}/2}\right)$

Case # 2: $f_{out} = 4f_{ref}$

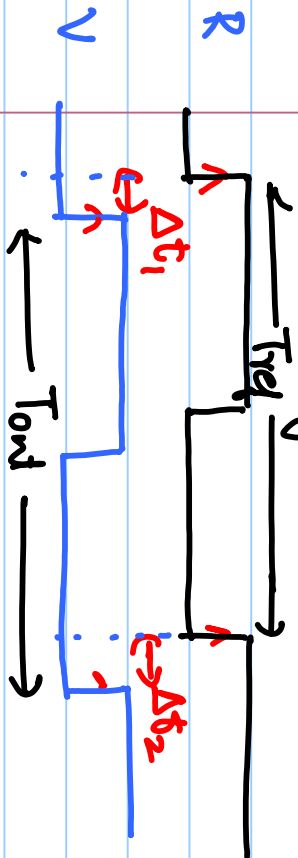
$$V_{pd} = \sin\left(2\pi \cdot \frac{\Delta t}{T_{ref}/4}\right)$$

1. Mixer / Multiplier based PD
 2. Sample & hold based PD
- } Analog PD

Digital PDs.



1. EXOR gate PD



$$R = \sin(\omega_{in}t + \Phi_{in}(0))$$

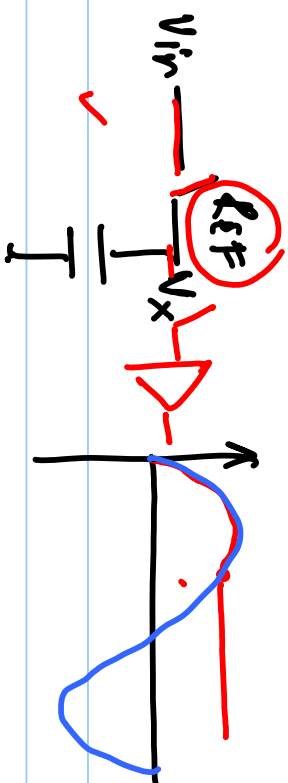
$$V = \sin(\omega_{out}t + \Phi_{out}(0))$$

$$\Phi_{er} = (\omega_{in} - \omega_{out})t + \Phi_{in}(0) - \Phi_{out}(0)$$

Phase error is defined w.r.t Reference

freq.

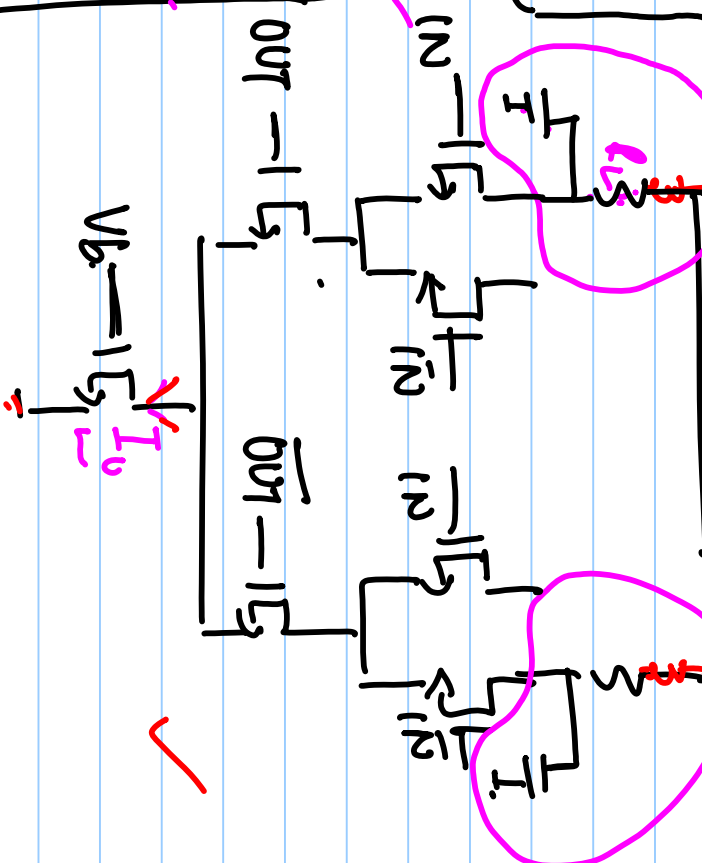
$$\Phi_{er} = 2\pi \cdot \frac{\Delta t}{T_{ref}}$$

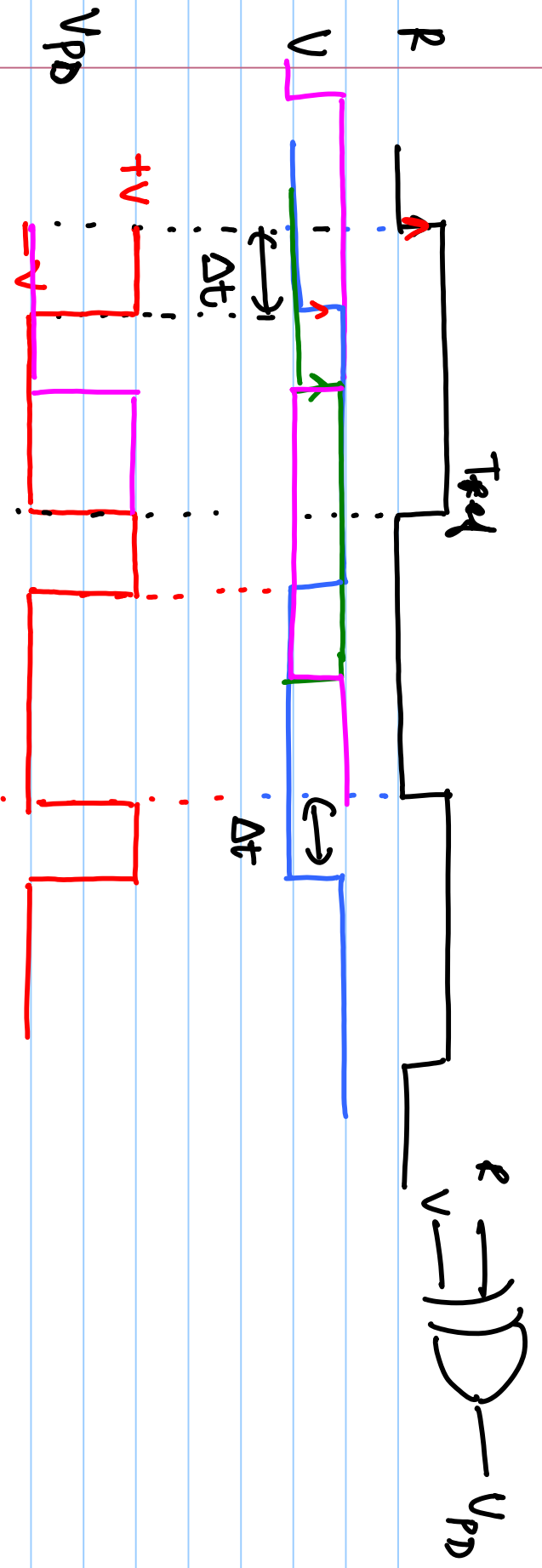


In a given technology.

— Upper limit on freq to

correctly sample i/p freq.



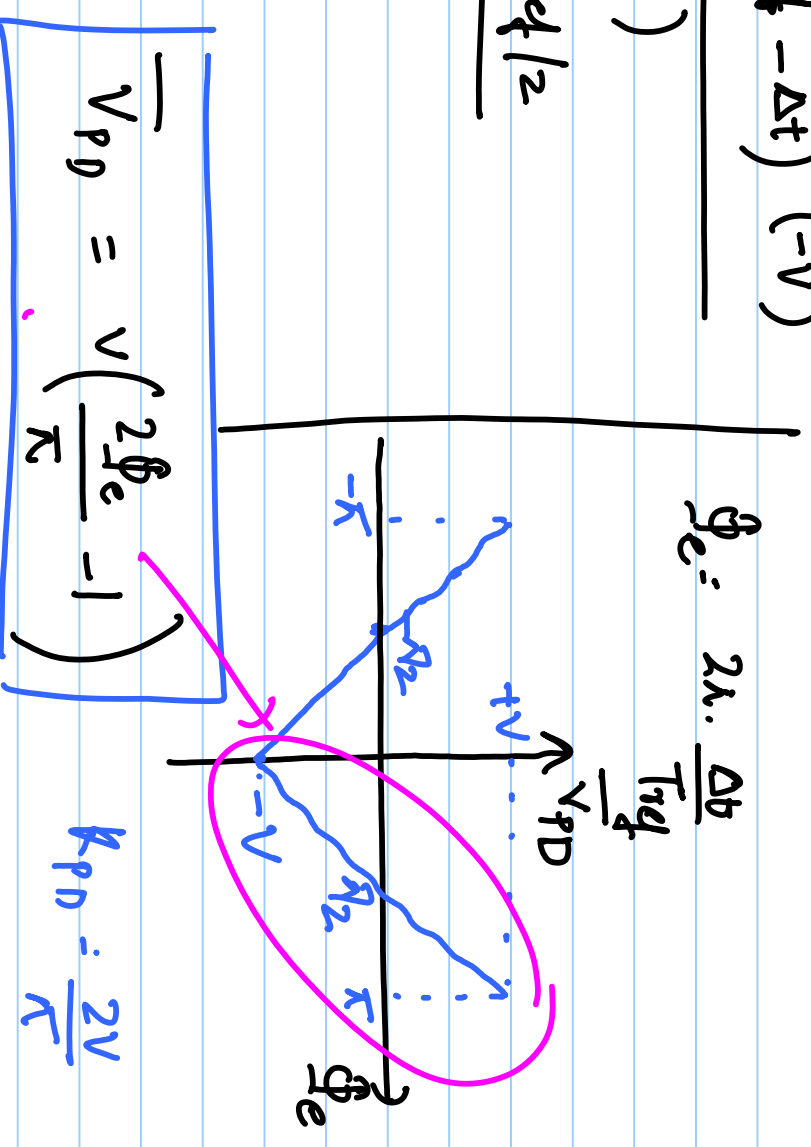


$$\overline{V_{PD}} = \frac{\Delta t \cdot +V + \left(\frac{T_{req} - \Delta t}{2}\right) (-V)}{(T_{req}/2)}$$

$$= \frac{2V \cdot \Delta t - V \cdot T_{req}/2}{T_{req}/2}$$

$$= 2V \cdot 2 \cdot \frac{\Delta t}{T_{req}} - V$$

$$= 2V \cdot \frac{\Phi_e}{\kappa} - V \Rightarrow \overline{V_{PD}} = V \left(\frac{2\Phi_e}{\kappa} - 1 \right)$$



$$k_{PD} = \frac{2V}{\kappa}$$