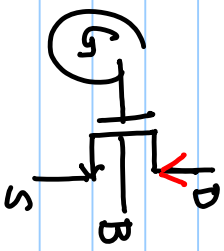


Lecture #7

NMOS



Cut off: $I_{DS} = 0, V_{GS} - V_{th} < 0$

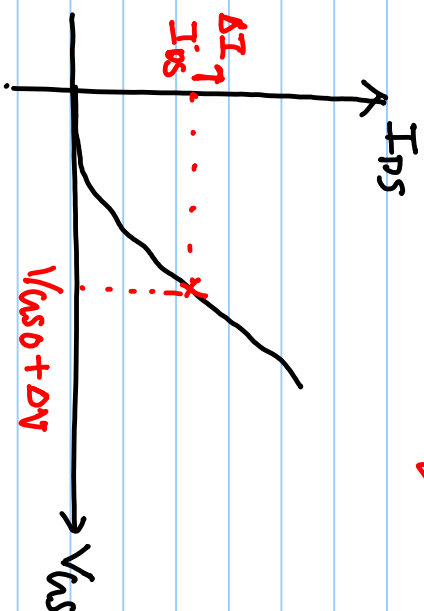
Linear: $I_{DS} = \mu_n C_{ox} \frac{W}{L} \left((V_{GS} - V_{th}) V_{DS} - \frac{V_{DS}^2}{2} \right), V_{DS} < V_{GS} - V_{th}$

Saturation: $I_{DS} = \mu_n C_{ox} \frac{W}{2L} \left[V_{GS} - V_{th} \right]^2 (1 + \lambda V_{DS}), V_{DS} \gg V_{GS} - V_{th}$

1. Resistor

2. Current source :

3. $\frac{\Delta I_{DS}}{\Delta V_{GS}} = \text{constant}$



$$\Delta I_{DS} = \left(\frac{\partial I_{DS}}{\partial V_{GS}} \right) \cdot \Delta V_{GS}$$

$$= \left(\frac{2 I_{DS}}{V_{GS} - V_{th}} \right) \cdot \Delta V_{GS}$$

$$I_{DS}' = \mu_n C_{ox} \frac{W}{2L} \left[V_{GS} + \Delta V - V_{thn} \right]^2$$

$$= \mu_n C_{ox} \frac{W}{2L} \left[\underbrace{(V_{GS} - V_{thn})^2}_{\substack{\left(\frac{2I_{DS}}{V_{GS} - V_{thn}} \right)^2}} + \Delta V^2 + 2(\Delta V) \underbrace{(V_{GS} - V_{thn})}_{\substack{\left(\frac{\Delta V}{V_{GS} - V_{thn}} \right)^2}} \right]$$

$$= I_{DSn} + \frac{2I_{DS}}{V_{GS} - V_{thn}} \cdot \Delta V + I_{DS} \cdot \left(\frac{\Delta V}{V_{GS} - V_{thn}} \right)^2$$

≈ 0

$$\Delta V \ll V_{GS} - V_{thn}$$

$$I_{DS} + \Delta I_{DS} = I_{DS}' = I_{DS} + \frac{2I_{DS}}{V_{GS} - V_{thn}} \cdot \Delta V$$

$$\Delta I_{DS} = \underbrace{\frac{2I_{DS}}{V_{GS} - V_{thn}}}_{g_m} \cdot \Delta V$$

Sat:

$$g_m = \frac{\partial I_{DS}}{\partial V_{GS}} = \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{th}) \quad (1)$$

$$\approx \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{th}) \quad \checkmark$$

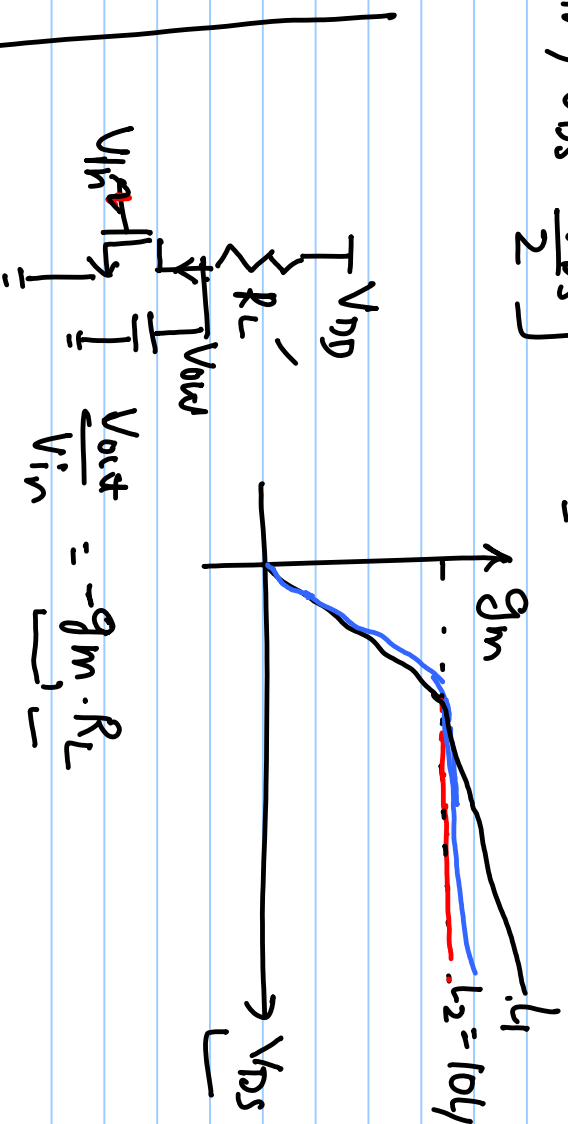
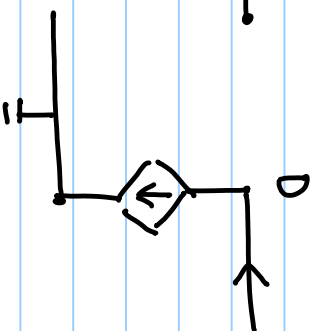
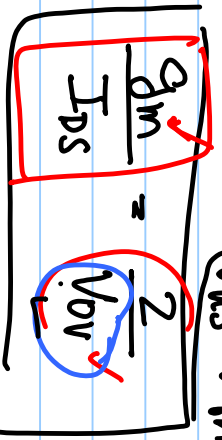
linear

$$g_m = \frac{\partial I_{DS}}{\partial V_{GS}} = \mu_n C_{ox} \frac{W}{L} V_{DS} \quad (2)$$

$$I_{DS} = \mu_n C_{ox} \frac{W}{L} \left[(V_{GS} - V_{th}) V_{DS} - \frac{V_{DS}^2}{2} \right]$$

$$\Delta I_{DS} = g_m \cdot \Delta V_{GS}$$

Sat:
$$g_m = \frac{2 I_{DS}}{V_{GS} - V_{th}} = \frac{2 I_{DS}}{V_{OV}}$$



$$\frac{V_{out}}{V_{in}} = -g_m \cdot R_L$$

$$\frac{g_m}{I_{DS}} = X \rightarrow I_{mA} \Rightarrow g_m = X \cdot I_{mA}$$

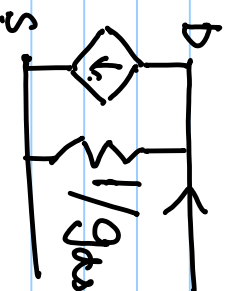
$$\frac{g_m}{I_{DS}} = 2X \Rightarrow g_m = 2X \cdot I_{mA}$$

$$V_{ov} > 2V_T$$

Sat:

$$g_{ds} = \frac{\partial I_{DS}}{\partial V_{DS}} = \mu_n C_{ox} \frac{W}{2L} (V_{GS} - V_{th})^2 \lambda = \frac{I_{DS}}{V_{DS}} \frac{\lambda}{(1 + \lambda V_{DS})} \approx \lambda I_{DS}$$

$$\text{lin. } \frac{\partial I_{DS}}{\partial V_{DS}} = \mu_n C_{ox} \frac{W}{L} [(V_{GS} - V_{th}) - V_{DS}]$$

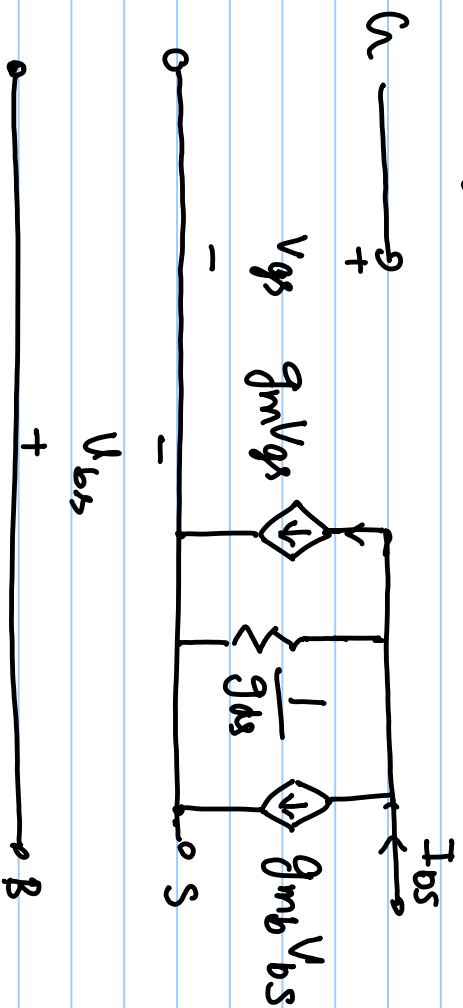


Sat:

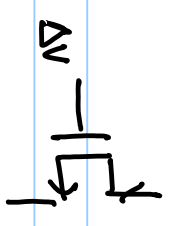
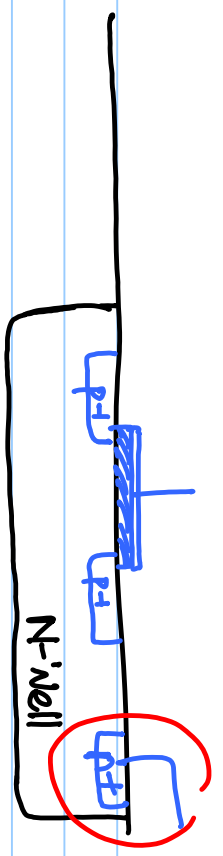
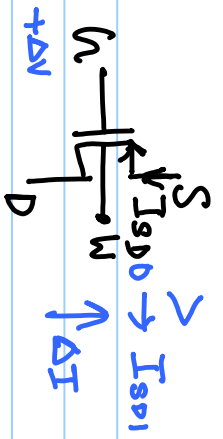
$$g_{mb} = \frac{\partial I_{DS}}{\partial V_{BS}} = \underbrace{\mu_n C_{ox} \frac{W}{L}}_{g_m} (V_{GS} - V_{th}) (1 + \lambda V_{DS}) \left(-\frac{\partial V_{th}}{\partial V_{BS}} \right)$$

$$= g_m \left(-\frac{\partial V_{th}}{\partial V_{BS}} \right)$$

$$= \eta g_m$$



DC Small signal Model.

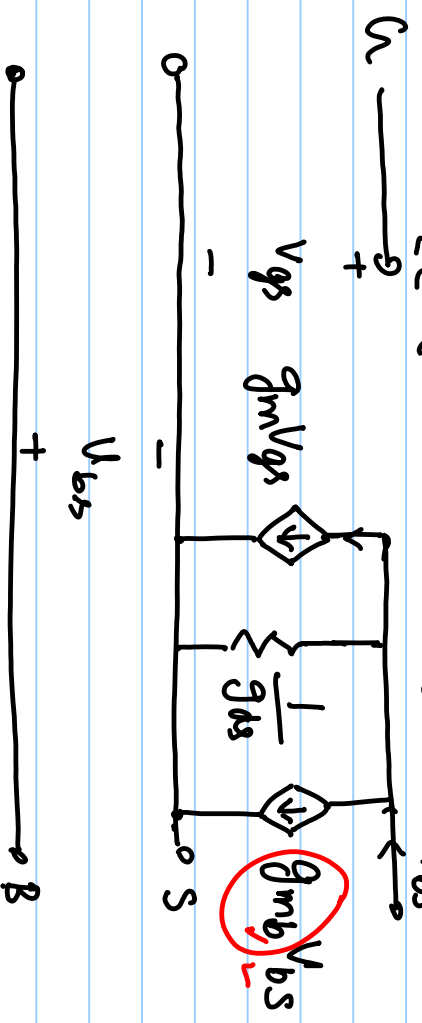


p-sub

Cut-off: $V_{SG} - |V_{thp}| < 0$, $I_{SD} = 0$

linear: $I_{SD} = \mu_p C_{ox} \frac{W}{L} \left[(V_{SG} - |V_{thp}|) V_{SD} - \frac{V_{SD}^2}{2} \right]$, $V_{SD} < (V_{SG} - |V_{thp}|)$

Sat: $I_{SD} = \mu_p C_{ox} \frac{W}{2L} \left[V_{SG} - |V_{thp}| \right]^2 (1 + \lambda V_{SD})$, $V_{SD} \geq V_{SG} - |V_{thp}|$



DC Small signal Model.

