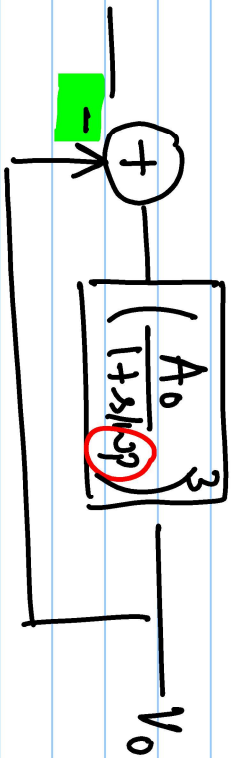
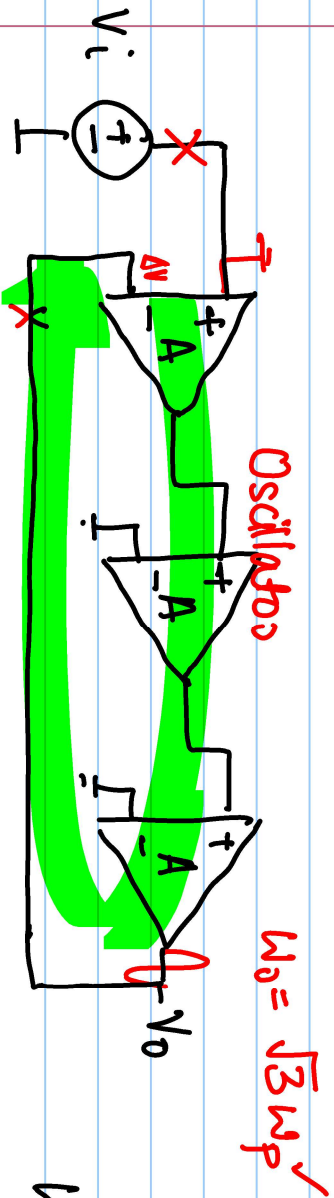


lecture # 42

Ring Oscillator.

1 Amplifier based ring oscillator



$$A(s) = \frac{A_0}{1 + s/\omega_p}$$

$$L_u = \left( \frac{A_0}{1 + s/\omega_p} \right)^3$$

$$\frac{V_0}{V_i} = \frac{L_u}{1 + L_u} = \frac{1}{1 + \frac{1}{L_u}}$$

$$\frac{V_0}{V_i} = \frac{1}{1 + \frac{1}{\cancel{A_0^3} B}}$$

Poles for closed loop tf:

$$1 + \frac{(1+s/w_p)^3}{A_0^3} = 0$$

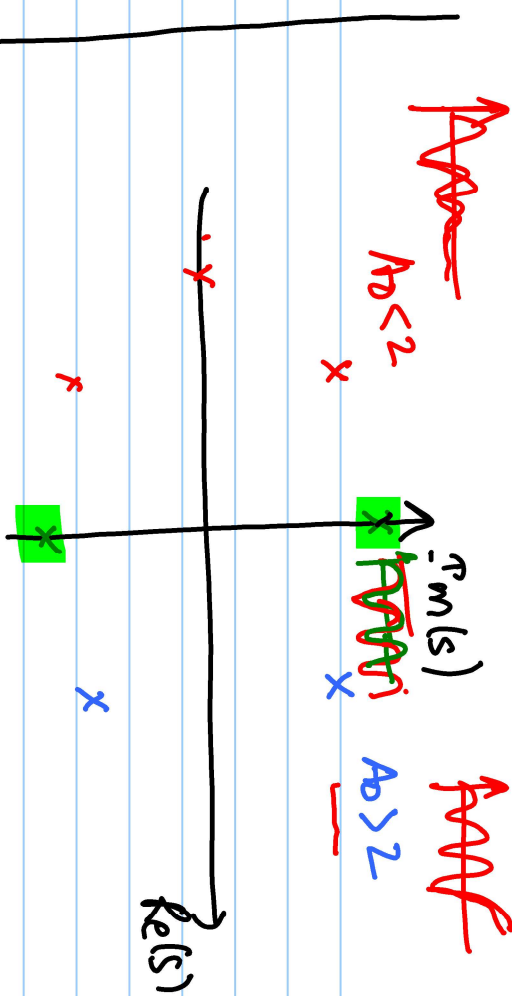
$$\begin{aligned} (1+s/w_p)^3 &= -A_0^3 = (-1) A_0^3 \\ &= A_0^3 e^{j(2k\pi + \pi)} \end{aligned}$$

$$\frac{s}{w_p} = -1 + A_0 e^{j\left(\frac{2k\pi}{3} + \frac{\pi}{3}\right)} \quad ; k=0, 1, 2$$

$$\begin{aligned} \frac{s}{w_p} &= -1 + A_0 e^{j\pi/3}, \quad -1 + A_0 e^{j\left(\frac{2\pi}{3} + \frac{\pi}{3}\right)}, \quad -1 + A_0 e^{j\left(\frac{4\pi}{3} + \frac{\pi}{3}\right)} \\ &= -1 + A_0 \left(\frac{1}{2} + j\frac{\sqrt{3}}{2}\right), \quad -1 + A_0 (-1), \quad -1 + A_0 \left(\frac{1}{2} - j\frac{\sqrt{3}}{2}\right) \end{aligned}$$

Roots in R.N.P :  $-1 + \frac{A_0}{2} > 0 \Rightarrow A_0 > 2$

Roots in L.N.P :  $-1 + \frac{A_0}{2} < 0 \Rightarrow A_0 < 2$



Barkhausen Criterion.

$$L_h(s) = - \left( \frac{A_0}{1 + s/\omega_p} \right)^3$$

$$|L_h(\omega_0)| = 1$$

$$\cancel{\Delta} L_h(\omega_0) = \underline{2K\pi}$$

$$|L_h(\omega_0)| = \left| \left( \frac{A_0}{1 + j\frac{\omega_0}{\omega_p}} \right)^3 \right| = 1$$

$$\cancel{\Delta} L_h(\omega_0) = +180^\circ - 3 \tan^{-1} \left( \frac{\omega_0}{\omega_p} \right) = \underline{+2K\pi}$$

$$\left| \left( \frac{A_0}{1 + j\sqrt{3}\frac{\omega_p}{\omega_p}} \right)^3 \right| = 1$$

$$\Rightarrow \frac{A_0^3}{2^3} = 1$$

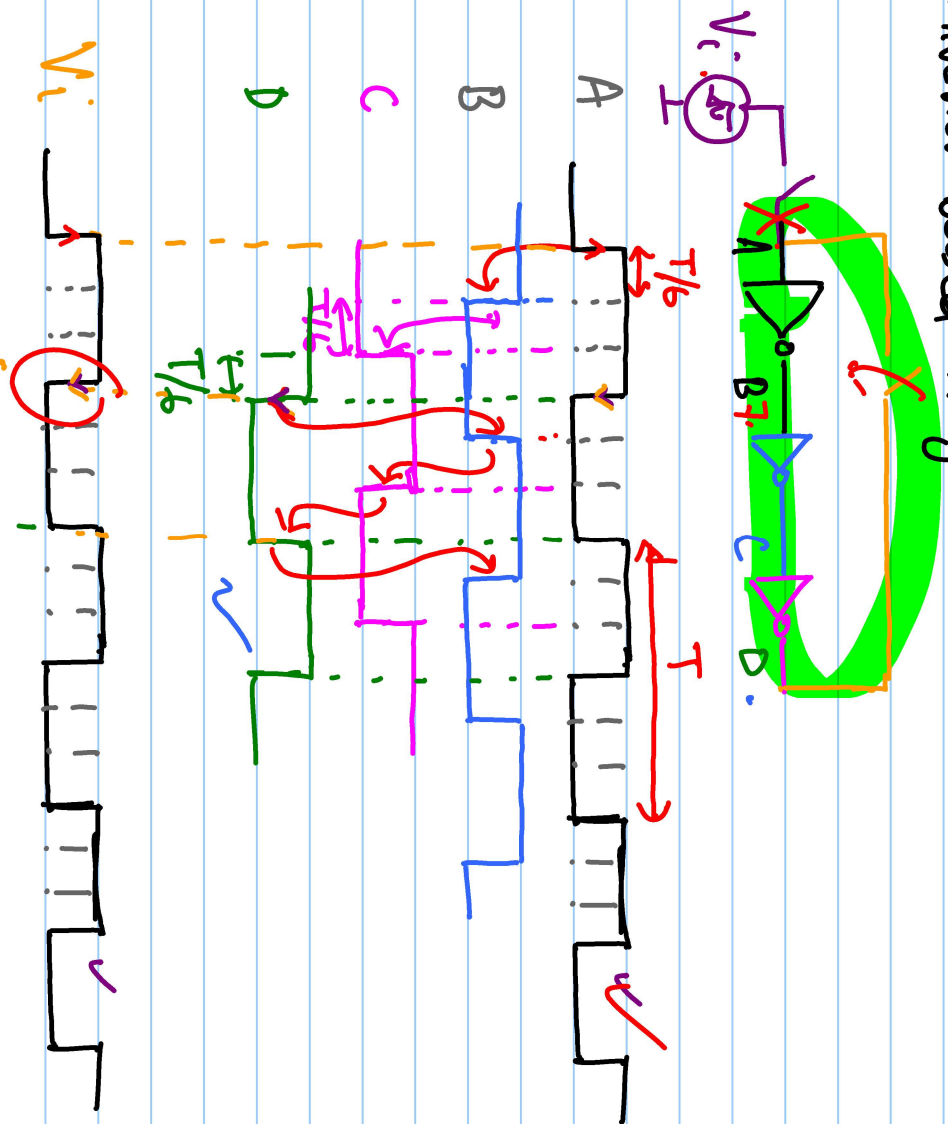
$$\boxed{A_0 = 2}$$

$$-3 \tan^{-1} \left( \frac{\omega_0}{\omega_p} \right) = -180^\circ; K=0$$

$$\frac{\omega_0}{\omega_p} = \tan(60^\circ)$$

$$\boxed{\omega_0 = \sqrt{3} \omega_p}$$

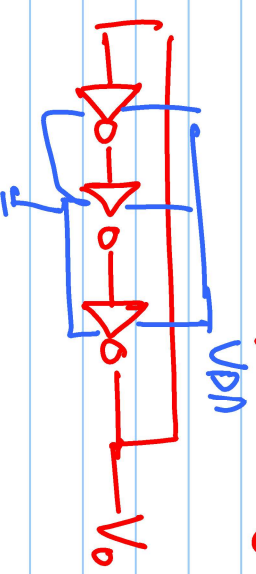
# Inverter-based Ring Oscillator.



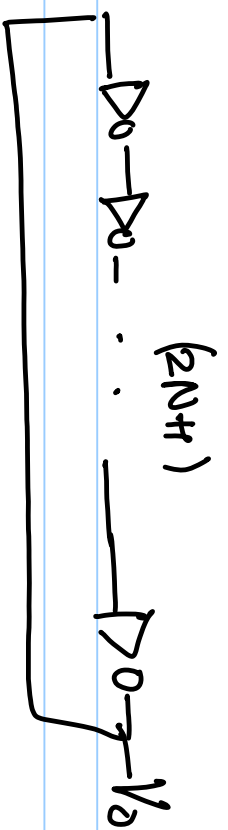
i/p  $\rightarrow$  o/p

For inverters i/p to o/p delay is  $T/6$

Inverter delay =  $\frac{T}{6}$



$$f_0 = \frac{1}{T}$$

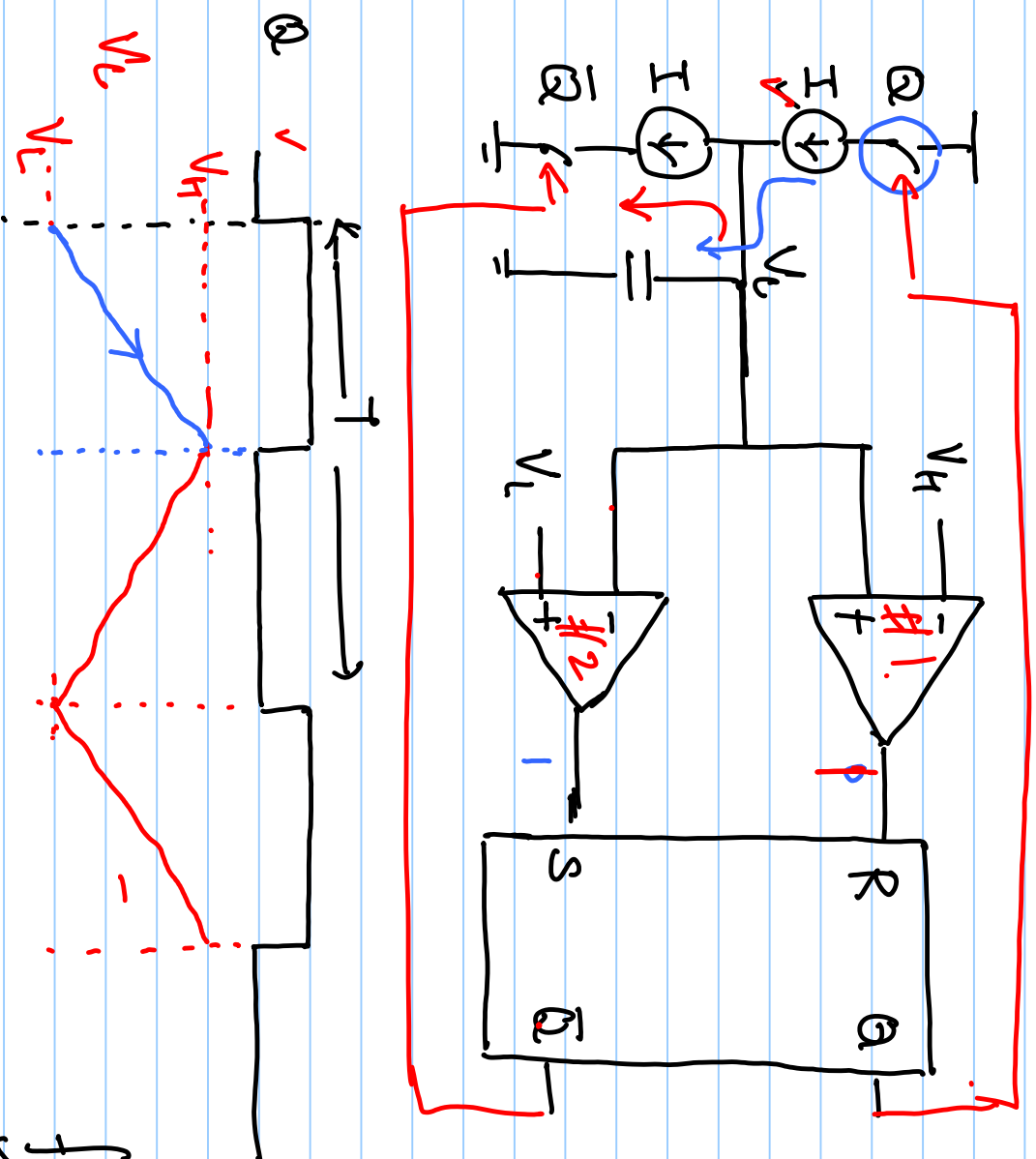


$$t_d = \frac{T}{2(2N+1)}$$

$$f_0 = \frac{1}{T}$$

- Frequency stability  $\Delta F$  (ppm) is poor
- Noise is more
- Large tunability
- Area is small.

# Relaxation Oscillators.



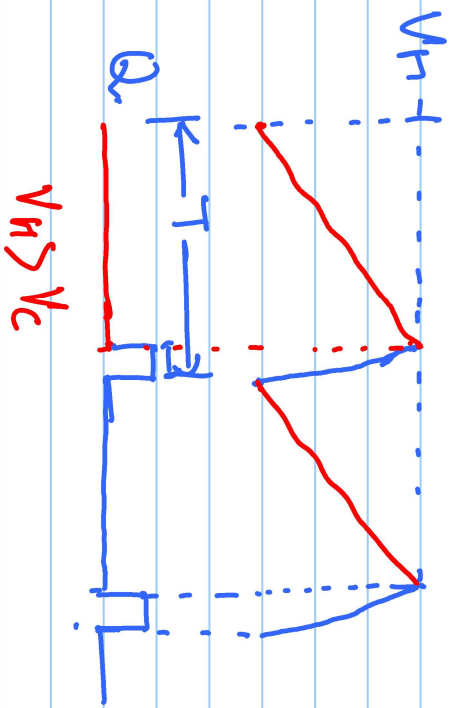
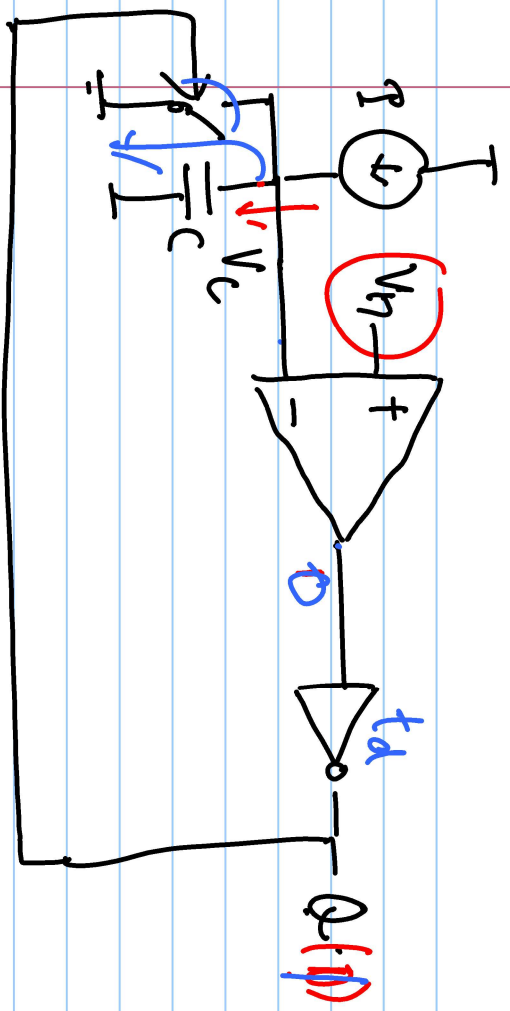
R	S	$\bar{Q}$	$\bar{\bar{Q}}$
0	0	Latch	Latch
1	0	0	1
0	1	1	0
1	1	0	0

- Switch is ON for signal value 1

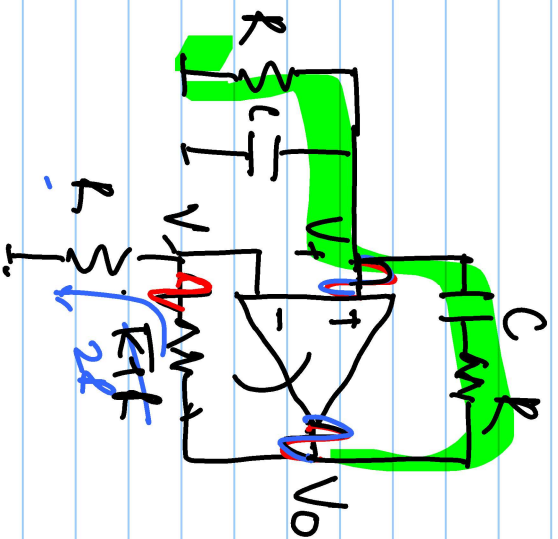
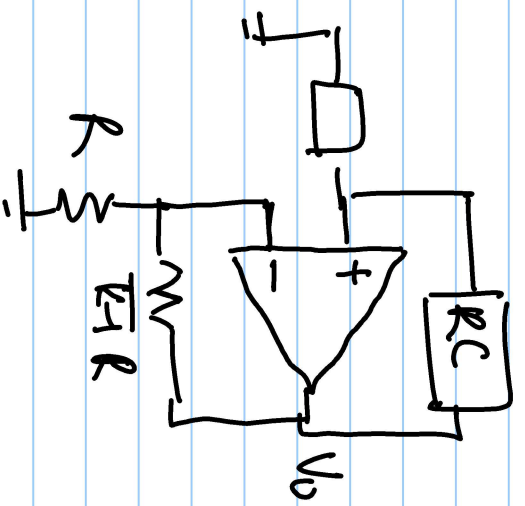
- Switch is OFF for signal value 0.

$$I \cdot \frac{I}{2} = C (V_H - V_L)$$

$$f = \frac{1}{T} = \frac{I}{2C (V_H - V_L)}$$



Wien Bridge Oscillator.



$$\frac{V_+}{V_0} = \frac{sRC}{1 + 3sRC + s^2R^2C^2}$$

$$\Delta \left( \frac{V_+}{V_0} \right) = 0 \quad \Rightarrow \quad \omega = \frac{1}{RC}$$

$$\left| \frac{V_+}{V_0} \right|_{\omega = 1/RC} = \frac{V_0}{3}$$

$$V_1 = \frac{V_0}{3} = \frac{1}{K} V_0$$

$$K=3$$

Gain of non-inverting amp. is  $> 3$ .