

Stochastic Geometry and Wireless Networks

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What is stochastic geometry?

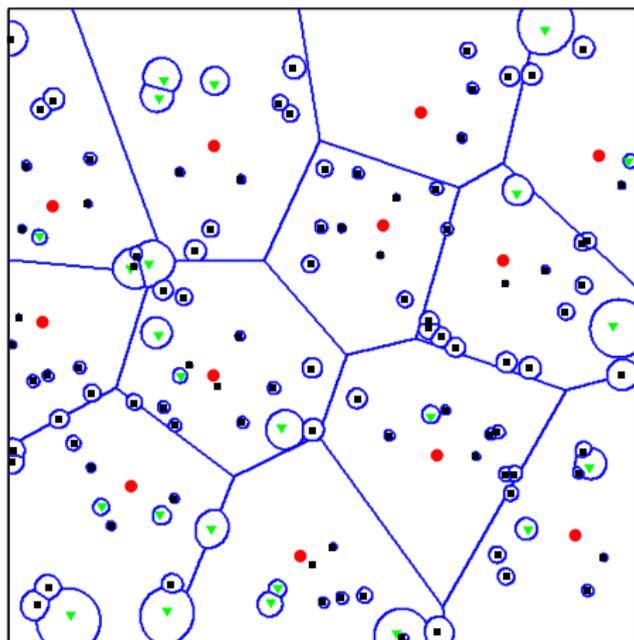
Stochastic geometry is the study of random spatial patterns

- ▶ Point processes
- ▶ Random tessellations
- ▶ Stereology

Applications

- ▶ Astronomy
- ▶ Communications
- ▶ Material science
- ▶ Image analysis and stereology
- ▶ Forestry
- ▶ Random matrix theory

Application to wireless networks



- ▶ Interference is a major limitation
- ▶ Networks are getting heterogeneous and decentralized

Outline

- * Primer on Point Processes
- * Ad hoc Networks
- * Cellular Networks
- * Heterogeneous Networks

Primer on Point Processes

* Primer on Point Processes

- Poisson point process
- Transformations of PPP
- Reduced Palm probability

* Ad hoc Networks

* Cellular Networks

* Heterogeneous Networks

What is a spatial point process?

Let \mathbb{N} be the set of all sequences $\phi \subset \mathbb{R}^2$ satisfying

1. (Finite) Any bounded set $A \subset \mathbb{R}^2$ contains finite number of points.
2. (Simple) $x_i \neq x_j$ if $i \neq j$.

Definition

A point process¹ in \mathbb{R}^2 is a **random variable** taking values in the space \mathbb{N} .

A simple representation: $\Phi = \sum_i \delta_{x_i}$

Notation:

1. Point process is denoted by Φ ; An instance of the point process is denoted by ϕ
2. Number of points of the point process in a set $A \subset \mathbb{R}^2$: $\Phi(A)$

¹D.J. Daley, D. Vere-Jones, *An introduction to the theory of point processes, Vol 1 and 2*, Springer

Example 1: An interesting but trivial point process

1. Contains only one point
2. The random point x is uniformly distributed in a bounded set A .

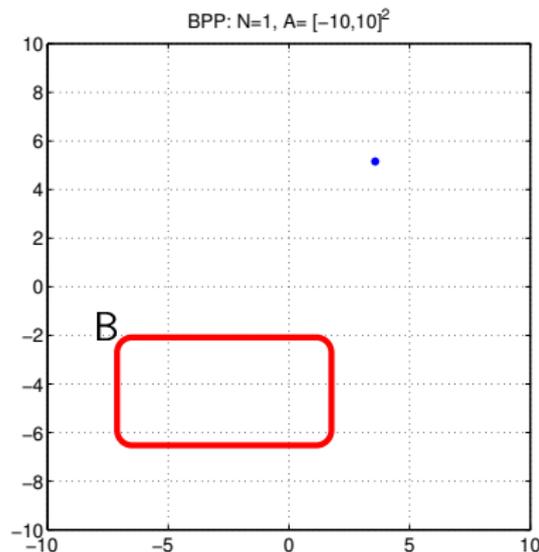
Uniform distribution

Let $B \subset A$

$$\mathbb{P}(x \in B) = \frac{|B|}{|A|},$$

where $|A|$ denotes the area of the set A .

Caveat: Defined only on bounded set, i.e., $|A| < \infty$.

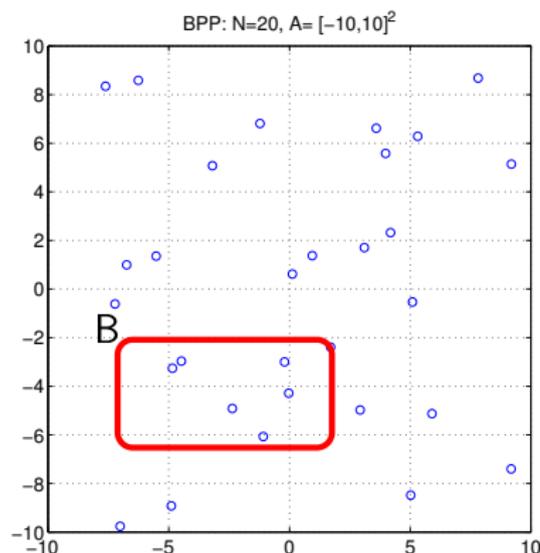


Example 2: Binomial point process (BPP)

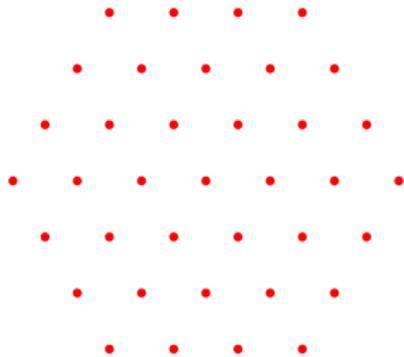
A BPP on a set A is the superposition of N independent uniformly distributed points on the set A .

Let $B \subset A$, then $\mathbb{P}(\Phi(B) = k) =$

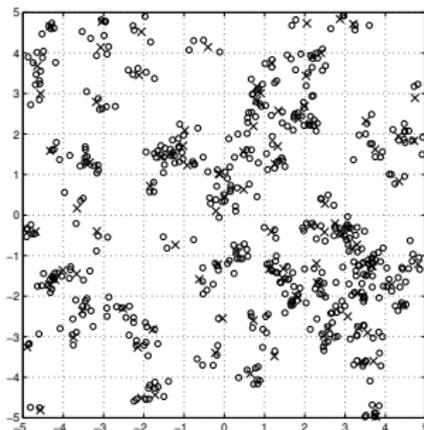
$$\binom{N}{k} \left(\frac{|B|}{|A|} \right)^k \left(1 - \frac{|B|}{|A|} \right)^{N-k}$$



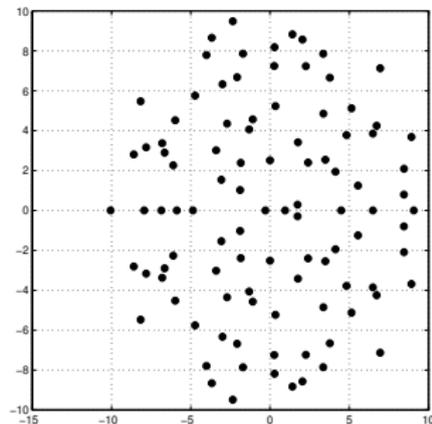
Other interesting examples



Hexagonal lattice



Thomas cluster process



Determinantal point process (eigenvalues of a Gaussian matrix)

Characterization of a point process

Given two point processes, is there a simple way to see if both of them are **equivalent**?

A simple point process Φ is determined by its void probabilities over all compact sets, *i.e.*, $\mathbb{P}(\Phi(K) = 0)$ for $K \subset \mathbb{R}^2$ and compact.

- ▶ This means that two point processes are equivalent if they have the same void probability distribution (for all sets).

Stationary point processes

Definition (Stationary point process)

A point process is stationary if its distribution is invariant with respect to translations.

- ▶ The point process looks statistically similar from any point in space.
- ▶ BPP is not a stationary point process.
- ▶ A stationary point process cannot be defined on a subset of \mathbb{R}^2

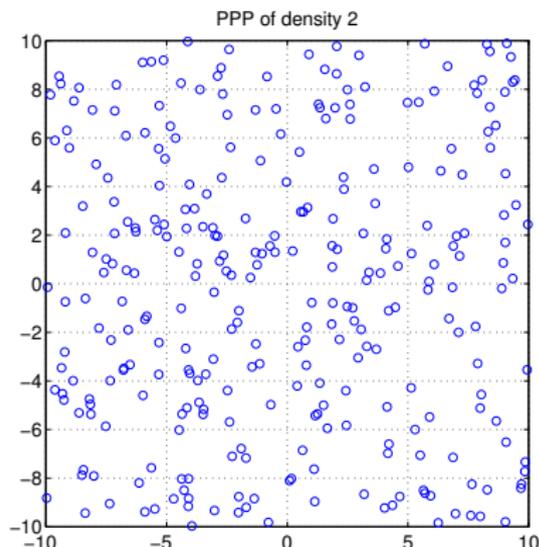
The density of a stationary point process Φ is defined as

$$\frac{\mathbb{E}[\Phi(B)]}{|B|}, \quad B \subset \mathbb{R}^2.$$

The RHS does not depend on the particular choice of the set B .

Stationary Poisson point process (PPP)

1. The most widely used model for spatial locations of nodes
 - ▶ Most amicable to analysis
 - ▶ "Gaussian of point processes"
2. No dependence between node locations
3. Random number of nodes
4. Can be defined on the entire plane
 - ▶ Limiting distribution of a BPP



PPP: Formal definition

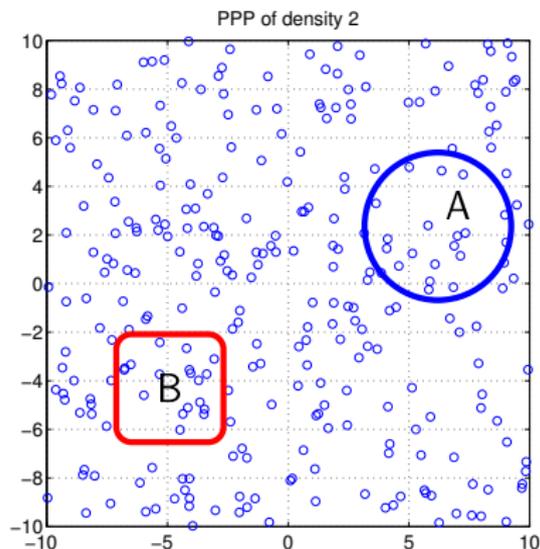
A stationary Poisson point process Φ of density λ is characterized by

1. The number of points in a bounded set $A \subset \mathbb{R}^2$ has a **Poisson distribution** with mean $\lambda|A|$, *i.e.*,

$$\mathbb{P}(\Phi(A) = n) = \exp(-\lambda|A|) \frac{(\lambda|A|)^n}{n!}$$

2. The number of points in disjoint sets are independent, *i.e.*, for $A \subset \mathbb{R}^2$, $B \subset \mathbb{R}^2$ and $A \cap B = \emptyset$,

$$\Phi(A) \perp \Phi(B)$$



A stationary PPP is completely characterized by a single number λ .

Properties of PPP

Lemma

The density of the PPP (as defined in previous slide) is λ .

Proof: Let $A \subset \mathbb{R}^2$. Then

$$\begin{aligned}\mathbb{E}[\Phi(A)] &= \sum_{n=0}^{\infty} n \exp(-\lambda|A|) \frac{(\lambda|A|)^n}{n!} \\ &= \lambda|A|,\end{aligned}$$

which follows from the mean of a Poisson random variable. Hence

$$\frac{\mathbb{E}[\Phi(A)]}{|A|} = \lambda.$$

Observe that the above expression does not depend on the set A . □

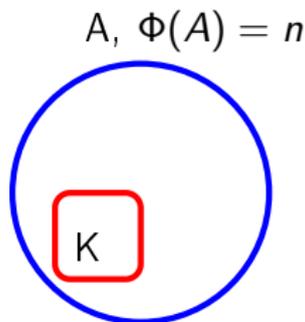
Properties...

Lemma

Let $A \subset \mathbb{R}^2$. Conditioned on the number of points $\Phi(A)$, the points are independently and uniformly distributed in the set A , i.e., the points form a BPP.

Proof: We consider the void probability of a set $K \subset A$.

$$\begin{aligned} \mathbb{P}(\Phi(K) = 0 | \Phi(A) = n) &= \frac{\mathbb{P}(\Phi(K) = 0 \cap \Phi(A) = n)}{\mathbb{P}(\Phi(A) = n)} \\ &= \frac{\mathbb{P}(\Phi(K) = 0) \mathbb{P}(\Phi(A \setminus K) = n)}{\mathbb{P}(\Phi(A) = n)} \\ &= \frac{e^{-\lambda|K|} e^{\lambda|A \setminus K|} (\lambda|A \setminus K|)^n / n!}{e^{\lambda|A|} (\lambda|A|)^n / n!} \\ &= \frac{|A \setminus K|^n}{|A|^n} = \left(1 - \frac{|K|}{|A|}\right)^n. \quad \square \end{aligned}$$



Simulation of a PPP

How to simulate a PPP of density λ on $A = [-L, L]^2$?

1. The number of points in the set A is a Poisson random variable with mean $\lambda|A|$.
2. Conditioned on the number of points, the points are uniformly distributed as a BPP.

Matlab code

```
 $N = \text{poissrnd}(\lambda|A|)$ 
```

```
Points = unifrnd(-L, L, N, 2)
```

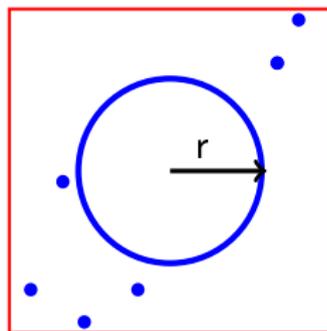
Distance properties of a PPP

First contact distribution

The CCDF of the distance of the nearest point of the process from the origin denoted by D is $\mathbb{P}(D \geq r) = \exp(-\lambda\pi r^2)$.

Proof:

$$\begin{aligned}\mathbb{P}(D \geq r) &= \mathbb{P}(B(o, r) \text{ is empty}) \\ &= \exp(-\lambda|B(o, r)|) \\ &= \exp(-\lambda\pi r^2) \quad \square\end{aligned}$$



Hence the PDF equals $f_N(r) = 2\lambda\pi r \exp(-\lambda\pi r^2)$. The average distance is

$$\mathbb{E}[D] = \int_0^\infty r 2\lambda\pi r f_N(r) dr = \frac{1}{2\sqrt{\lambda}}.$$

N -th closest point

The CDF² of the N -th closest point to the origin equals

$$\mathbb{P}(D_n \geq r) = \sum_{k=0}^{n-1} e^{-\lambda\pi r^2} \frac{(\lambda\pi r^2)^k}{k!} = \frac{\Gamma(n, \lambda\pi r^2)}{(n-1)!}$$

The average distance to the N -th closest point equals

$$\mathbb{E}[D_n] = \frac{\Gamma(n + \frac{1}{2})}{\sqrt{\pi\lambda}\Gamma(n)} \sim \sqrt{\frac{n}{\pi\lambda}}$$

²M. Haenggi, "On Distances in Uniformly Random Networks," IEEE Transactions on Information Theory, vol. 51, pp. 3584-3586, Oct. 2005

Sums over PPP

Lemma (Campbells theorem)

Let Φ be a PPP of density λ and $f(x) : \mathbb{R}^2 \rightarrow \mathbb{R}^+$.

$$\mathbb{E}\left[\sum_{x \in \Phi} f(x)\right] = \lambda \int_{\mathbb{R}^2} f(x) dx$$

Proof: We have

$$\mathbb{E}\left[\sum_{x \in \Phi} f(x)\right] = \lim_{R \rightarrow \infty} \mathbb{E}\left[\sum_{x \in \Phi \cap B(o, R)} f(x)\right].$$

Let $n = \Phi(B(o, R))$. Conditioning on the number of points n ,

$$\mathbb{E}\left[\sum_{x \in \Phi \cap B(o, R)} f(x)\right] = \mathbb{E}_n \left[\mathbb{E}\left[\sum_{x \in \Phi \cap B(o, R)} f(x) \mid n\right]\right]$$

Since conditioned on the number of points, the points are i.i.d uniform

$$\mathbb{E}\left[\sum_{x \in \Phi \cap B(o, R)} f(x) \mid n\right] = n \int_{B(o, R)} \frac{f(x)}{|B(o, R)|} dx.$$

Averaging over n

$$\mathbb{E}\left[\sum_{x \in \Phi \cap B(o, R)} f(x)\right] = \mathbb{E}[n] \int_{B(o, R)} \frac{f(x)}{|B(o, R)|} dx$$

As $\mathbb{E}[n] = \lambda |B(o, R)|$, and tending $R \rightarrow \infty$ we obtain the result. \square

Let $g(x, y) : \mathbb{R}^2 \times \mathbb{R}^2 \rightarrow \mathbb{R}^+$. Then³

$$\mathbb{E} \sum_{x, y \in \Phi}^{\neq} g(x, y) = \lambda^2 \int_{\mathbb{R}^2} \int_{\mathbb{R}^2} g(x, y) dx dy.$$

³D. Stoyan, W. Kendall, and J. Mecke, "Stochastic Geometry and Its Applications", 2nd ed. John Wiley and Sons, 1996

Example: Mean and variance of interference

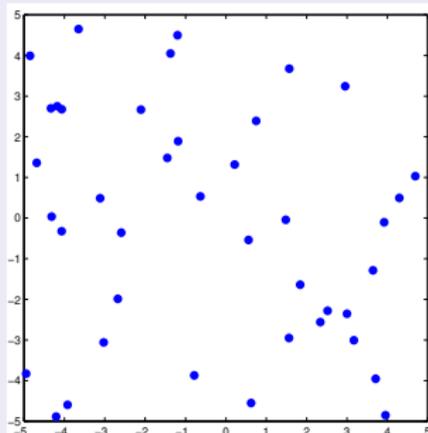
Let the transmitters be distributed as a PPP Φ of density λ .

Definition (Interference)

The interference (sum power) at location $y \in \mathbb{R}^2$ is

$$I(y) = \sum_{x \in \Phi} \ell(x - y),$$

where $\ell(x)$ is the path-loss function.



Mean of interference: By Campbell theorem,

$$\mathbb{E}[I(y)] = \mathbb{E}\left[\sum_{x \in \Phi} \ell(x - y)\right] = \lambda \int_{\mathbb{R}^2} \ell(x - y) dx = \lambda \int_{\mathbb{R}^2} \ell(x) dx$$

Variance of interference:

$$\begin{aligned}
 \mathbb{E}[I(y)^2] &= \mathbb{E} \left[\left(\sum_{x \in \Phi} \ell(x - y) \right)^2 \right] = \mathbb{E} \left[\left(\sum_{x \in \Phi} \ell(x - y) \right) \left(\sum_{z \in \Phi} \ell(z - y) \right) \right] \\
 &= \mathbb{E} \left[\sum_{x \in \Phi} \ell(x - y)^2 \right] + \mathbb{E} \left[\sum_{\substack{x, z \in \Phi \\ x \neq z}} \ell(x - y) \ell(z - y) \right] \\
 &= \lambda \int_{\mathbb{R}^2} \ell(x - y)^2 dx + \lambda^2 \int_{\mathbb{R}^2} \int_{\mathbb{R}^2} \ell(x - y) \ell(z - y) dx dz \\
 &= \lambda \int_{\mathbb{R}^2} \ell(x)^2 dx + (\lambda \int_{\mathbb{R}^2} \ell(x) dx)^2
 \end{aligned}$$

Hence variance equals,

$$\text{var}(I(y)) = \lambda \int_{\mathbb{R}^2} \ell(x)^2 dx.$$

Products over PPP

Lemma (Probability generating functional (PGFL))

Let Φ be a PPP of density λ and $f(x) : \mathbb{R}^2 \rightarrow [0, 1]$ be a real valued function. Then

$$\mathbb{E} \left[\prod_{x \in \Phi} f(x) \right] = \exp \left(-\lambda \int_{\mathbb{R}^2} (1 - f(x)) dx \right).$$

Proof: We prove the result for $\Psi_r = \Phi \cap B(o, r)$. Observe that Ψ_r is a PPP with number of points n distributed as a Poisson random variable with mean $\lambda \pi r^2$.

$$\begin{aligned} \mathbb{E} \left[\prod_{x \in \Psi_r} f(x) \right] &= \mathbb{E}_n \mathbb{E} \left[\prod_{x \in \Psi_r} f(x) \mid n \right] \\ &= \mathbb{E}_n \mathbb{E}[f(x)]^n \end{aligned}$$

But $\mathbb{E}[f(x)] = \frac{1}{\pi r^2} \int_{B(o,r)} f(x) dx$. Hence

$$\mathbb{E} \left[\prod_{x \in \Psi_r} f(x) \right] = \mathbb{E}_n \left[\left(\frac{1}{\pi r^2} \int_{B(o,r)} f(x) dx \right)^n \right].$$

Let $z > 0$. Let n be a Poisson random variable with mean a . Then

$$\mathbb{E}[z^n] = \exp(-a(1 - z)).$$

$$\begin{aligned} \mathbb{E} \left[\prod_{x \in \Psi_r} f(x) \right] &= \exp \left(-\lambda \pi r^2 \left(1 - \frac{1}{\pi r^2} \int_{B(o,r)} f(x) dx \right) \right) \\ &= \exp \left(-\lambda \int_{B(o,r)} (1 - f(x)) dx \right). \quad \square \end{aligned}$$

Application of PGFL: Laplace transform of interference

$$\mathbb{E}[\exp(-sI(y))] = \mathbb{E} \left[\exp \left(-s \sum_{x \in \Phi} \ell(x-y) \right) \right].$$

This can be rewritten as,

$$\mathbb{E}[\exp(-sI(y))] = \mathbb{E} \left[\prod_{x \in \Phi} \exp(-s\ell(x-y)) \right].$$

Using the PGFL,

$$\mathbb{E}[\exp(-sI(y))] = \exp \left(-\lambda \int_{\mathbb{R}^2} 1 - e^{-s\ell(x-y)} dx \right).$$

Substituting $x - y \rightarrow x$, we have

$$\mathcal{L}_I(s) = \exp \left(-\lambda \int_{\mathbb{R}^2} 1 - e^{-s\ell(x)} dx \right).$$

Transformations of PPP: Independent Thinning

Let Φ be a PPP of density λ

1. A node $x \in \Phi$ is coloured red with probability p and blue with probability $1 - p$.
2. Let Φ_r denote the red point process and Φ_b denote the blue point process. So we have $\Phi = \Phi_r \cup \Phi_b$.

Can be used to model ALOHA MAC protocol.

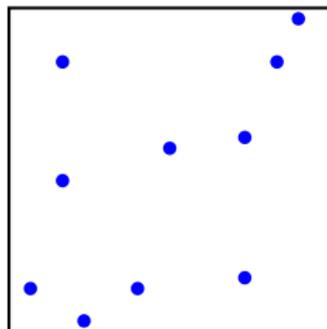
Lemma (Thinning)

1. Φ_r is a PPP of density λp
2. Φ_b is a PPP of density $\lambda(1 - p)$
3. Φ_r is independent of Φ_b .

Proof: Look at void probabilities of Φ_r and Φ_b . \square

Matern hard-core process: Dependent thinning

1. Begin with a PPP Φ of density λ .
2. To each $x \in \Phi$, associate a mark $m_x \sim U[0, 1]$ independent of every other point.
3. A node $x \in \Phi$ selected if it has the lowest mark among all the points in the ball $B(x, R)$.



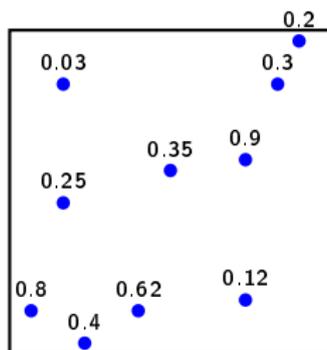
$$\Psi = \{y : y \in \Phi, m_y \leq m_x, \forall x \in B(y, R) \cap \Phi\}$$

A minimum distance process for modelling CSMA MAC.

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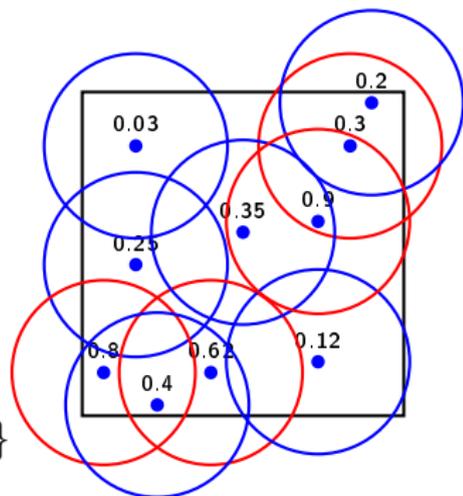


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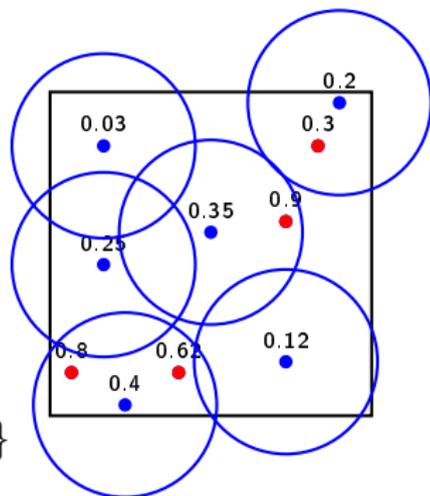


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A minimum distance process for modelling CSMA MAC.

Density of Matern hard-core process

A node $x \in \phi$ is retained with probability

$$p = \mathbb{P}(m_x \leq m_y, \forall y \in B(x, R) \cap \Phi).$$

- ▶ Let the mark of x be equal to $t \in [0, 1]$.
- ▶ Let n represent the number of points of Φ in $B(x, R)$.
 - ▶ $n \sim Poi(\lambda\pi R^2)$.
- ▶ Conditioned on the mark t , the probability that x is selected equals $\mathbb{E}[(1 - t)^n] = \exp(-\lambda\pi R^2 t)$.
- ▶ Averaging over t ,

$$p = \int_0^1 \exp(-\lambda\pi R^2 t) dt = \frac{1 - \exp(-\lambda\pi R^2)}{\lambda\pi R^2}.$$

So the final density of the process equals $\lambda_m = p\lambda$.

$$\lambda_m = \frac{1 - \exp(-\lambda\pi R^2)}{\pi R^2}$$

Conditioning: Reduced Palm probability

- ▶ Notion of a typical point, *i.e.*, conditioning on the existence of a node at a particular location

Nearest-neighbour distribution function

The distance of the nearest neighbour from a "typical" point.

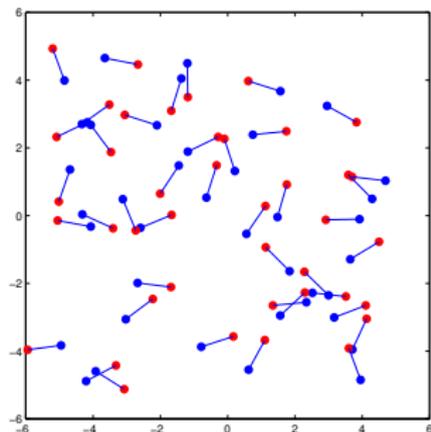
$$\begin{aligned} D(r) &= \mathbb{P}(\Phi(B(o, r)) = 1 | o \in \Phi) \\ &= \mathbb{P}^o(\Phi(B(o, r)) = 1) \\ &= \mathbb{P}^{!o}(\Phi(B(o, r)) = 0) \end{aligned}$$

Probability conditioned on there being a point at the origin (but not counting it).

A spatial average interpretation

The reduced Palm probability can also be interpreted as a spatial average

$$\mathbb{P}^{!o}(Y) = \lim_{R \rightarrow \infty} \mathbb{E} \sum_{x \in \Phi \cap B(o, R)} \frac{\mathbb{P}(\Phi_{-x} \setminus \{x\} \in Y)}{\lambda \pi R^2}.$$



Palm distribution of PPP: Slivnyak theorem

$$\mathbb{P}^{!o} = \mathbb{P},$$

i.e., reduced Palm distribution of a PPP equals the original distribution.

Hence for a PPP, a new point can be added to the process without disturbing other points of the process.

Campbell Mecke Theorem

Let $f(x, \phi) : \mathbb{R}^2 \times \mathbb{N} \rightarrow [0, \infty]$ be a real valued function,

$$\mathbb{E}\left[\sum_{x \in \Phi} f(x, \Phi \setminus \{x\})\right] = \lambda \int_{\mathbb{R}^2} \mathbb{E}^{!o}[f(x, \Phi)] dx.$$

Hence for a PPP,

$$\mathbb{E}\left[\sum_{x \in \Phi} f(x, \Phi \setminus \{x\})\right] = \lambda \int_{\mathbb{R}^2} \mathbb{E}[f(x, \Phi)] dx.$$

Analysis of Ad Hoc Networks

* Primer on Point Processes

* Ad hoc Networks

– *SINR* analysis

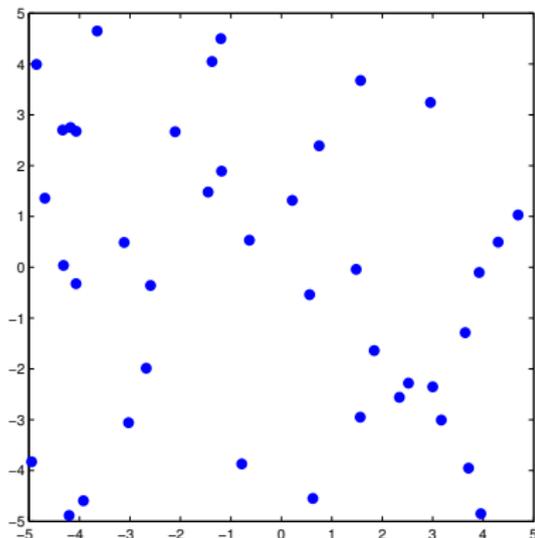
– Interference correlation

* Cellular Networks

* Heterogeneous Networks

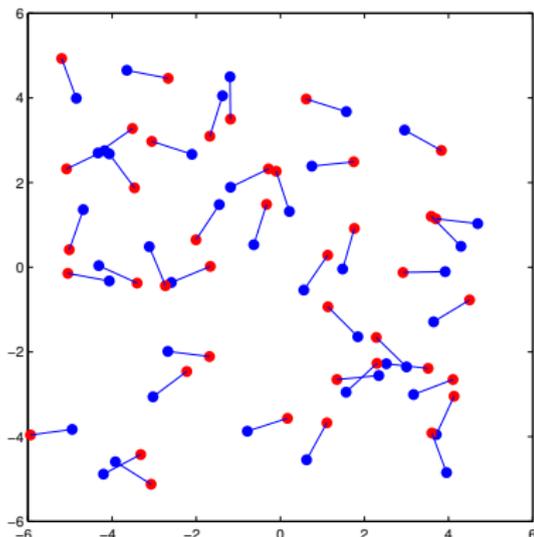
Dipole model

- ▶ The transmitters are distributed as a PPP Φ of density λ
- ▶ Each transmitter has a receiver at a distance d in a random direction
 - ▶ Not part of the process Φ .
- ▶ Path loss function is denoted by $\ell(x)$
 1. Examples: $\ell(x) = \|x\|^{-\alpha}$,
 $\ell(x) = \min\{1, \|x\|^{-\alpha}\}$.
 2. $\alpha > 2$
 3. Path loss between nodes x and y is $\ell(x - y)$.



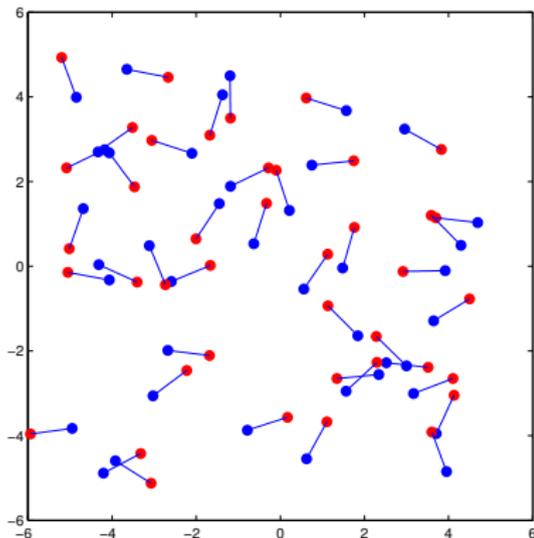
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System model

- ▶ All nodes transmit at the same power
- ▶ TX has N_t transmit antenna
- ▶ RX has N_r receive antenna
- ▶ Fading between any two nodes is i.i.d Rayleigh
 1. The fading power is exponentially distributed with unit mean.
 2. The fading power between nodes is denoted by h_{xy}
 - ▶ $\mathbb{P}(h_{xy} \geq z) = \exp(-z)$

What is the performance of a "typical" link?

³F. Baccelli, B. Blaszczyszyn, P. Muhlethaler, "An ALOHA protocol for multihop mobile wireless networks," Information Theory, IEEE Transactions on , vol.52, no.2, pp. 421- 436, Feb. 2006

SISO ad hoc network: $N_t = N_r = 1$

Typical link: By Slivnyaks theorem, we can add a new reference transmitter with its receiver at the origin

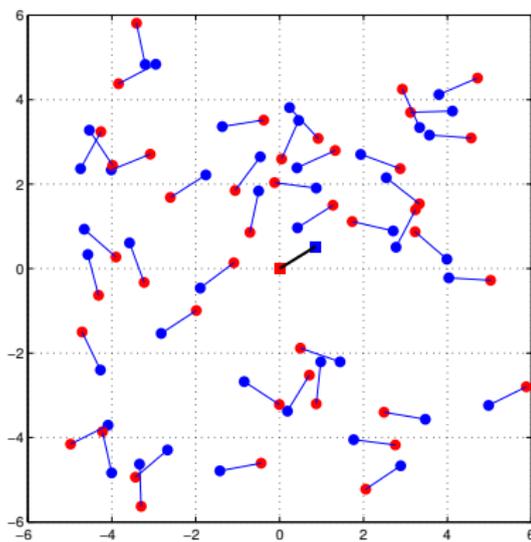
The Signal-to-interference-noise ratio (after processing) between the receiver at the origin and its corresponding transmitter is

$$\text{SINR} = \frac{h d^{-\alpha}}{\sigma^2 + I(o)},$$

where $I(o)$ is the interference at receiver at origin

$$I(o) = \sum_{y \in \Phi \setminus \{x\}} h_{y0} \|y\|^{-\alpha}.$$

How to compute $\mathbb{P}(\text{SINR} \geq \theta)$ for the reference link?



SINR distribution

$$\begin{aligned}
 p_s(\theta, \lambda) &= \mathbb{P}(\text{SINR}(o) > \theta) = \mathbb{P}\left(\frac{hd^{-\alpha}}{\sigma^2 + I(o)} \geq \theta\right) \\
 &= \mathbb{P}(h \geq d^\alpha \theta (\sigma^2 + I(o))) \\
 &= \mathbb{E} \exp(-d^\alpha \theta (\sigma^2 + I(o))) \\
 &= \exp(-d^\alpha \theta \sigma^2) \underbrace{\mathbb{E} \exp(-d^\alpha \theta I(o))}_{T_1 = \mathcal{L}_{I(o)}(d^\alpha \theta)}
 \end{aligned}$$

Observe that T_1 is the Laplace transform of $I(o)$ evaluated at $s = d^\alpha \theta$. We now evaluate the Laplace transform of interference

$$\begin{aligned}
 \mathcal{L}_{I(o)}(s) &= \mathbb{E} \exp\left(-s \sum_{y \in \Phi} h_{y_o} \|y\|^{-\alpha}\right) \\
 &= \mathbb{E} \prod_{y \in \Phi} \exp(-s h_{y_o} \|y\|^{-\alpha})
 \end{aligned}$$

Laplace transform of interference

$$\begin{aligned} \mathcal{L}_{I(o)}(s) &= \mathbb{E} \prod_{y \in \Phi} \exp(-sh_{y_o} \|y\|^{-\alpha}) \\ &\stackrel{(a)}{=} \mathbb{E} \prod_{y \in \Phi} \mathbb{E}_{h_{y_o}} \exp(-sh_{y_o} \|y\|^{-\alpha}) \\ &\stackrel{(b)}{=} \mathbb{E} \prod_{y \in \Phi} \frac{1}{1 + s \|y\|^{-\alpha}} \end{aligned}$$

(a) follows from the independence of the fading random variables and (b) follows from the Laplace transform of an exponential random variable.

Recall PGFL of a PPP

$$\mathbb{E} \prod_{y \in \Phi} f(x) = \exp \left(-\lambda \int_{\mathbb{R}^2} 1 - f(x) dx \right).$$

Using the PGFL of a PPP,

$$\begin{aligned}\mathcal{L}_{I(o)}(s) &= \exp\left(-\lambda \int_{\mathbb{R}^2} 1 - \frac{1}{1 + s\|y\|^{-\alpha}} dx\right) \\ &= \exp\left(-\lambda \int_{\mathbb{R}^2} \frac{1}{1 + s^{-1}\|y\|^\alpha} dx\right) \\ &= \exp(-\lambda s^{2/\alpha} C(\alpha)),\end{aligned}$$

where $C(\alpha) = \frac{2\pi^2}{\alpha \sin(2\pi/\alpha)}$.

The CCDF of SINR is

$$\begin{aligned}p_s(\theta, \lambda) = \mathbb{P}(\text{SINR}(o) > \theta) &= \exp(-d^\alpha \theta \sigma^2) \mathcal{L}_{I(o)}(d^\alpha \theta) \\ &= \underbrace{\exp(-d^\alpha \theta \sigma^2)}_{\text{Noise}} \underbrace{\exp(-\lambda d^2 \theta^{2/\alpha} C(\alpha))}_{\text{Interference}}\end{aligned}$$

Multiple antenna systems:

Post-processing SIR depends on the processing at the receiver

- ▶ *Maximal-ratio combining (MRC)*, $N_t = 1$, $N_r = n$: Let h_x denote $1 \times n$ channel vector from a node x to the receiver at the origin. The received signal is

$$Y = \frac{h_o d^{-\alpha/2}}{\sqrt{n}} a_o + \sum_{x \in \Phi} \frac{h_x \|x\|^{-\alpha/2}}{\sqrt{n}} a_x,$$

where a_x are the transmitted symbols. Hence the received SIR after multiplying with h_o^H is

$$\text{SIR} = \frac{\frac{1}{n} |h_o^H h_o|^2 d^{-\alpha}}{\frac{1}{n} \sum_{x \in \Phi} |h_o^H h_x|^2 \|x\|^{-\alpha}} = \frac{\|h_o\|^2 d^{-\alpha}}{\sum_{x \in \Phi} \left| \frac{h_o^H}{\|h_o\|} h_x \right|^2 \|x\|^{-\alpha}}$$

- ▶ $\|h_o\|^2$ is χ^2 distributed with $2n$ degrees of freedom.
- ▶ $\left| \frac{h_o^H}{\|h_o\|} h_x \right|^2$ is exponentially distributed with unit mean.

- ▶ Zero forcing receiver (ZF), $N_t = n, N_r = n, \# \text{ streams} = n$. Looking at the k -th stream the received vector is

$$\sqrt{n}Y_k = h_o(k)d^{-\alpha/2}a_{ok} + \sum_{i=1, i \neq k}^n h_o(i)d^{-\alpha/2}a_{oi} + \sum_{x \in \Phi} \sum_{i=1}^n h_x(n)\|x\|^{-\alpha/2}a_x$$

$h_x(i)$ is the i -th column of the channel matrix between the node x and the tagged receiver. Multiplying with the v that is orthogonal to the vectors $h_o(i), i \neq k$ the received SINR is

$$\text{SIR} = \frac{Sd^{-\alpha}}{\sum_{x \in \Phi} S_x \|x\|^{-\alpha}}$$

1. S is exponentially distributed.
2. S_x is also exponentially distributed.

The post process signal-to-interference ratio (after processing) is generally of the form

$$\text{SIR} = \frac{Sd^{-\alpha}}{\sigma^2 + I(o)},$$

where $I(o) = \sum_{y \in \Phi \setminus \{x\}} \hat{h}_{y0} \|y\|^{-\alpha}$ is the interference at receiver at origin. Let the CCDF of S be

$$F(x) = \sum_{k=0}^n a_k x^k \exp(-b_k x)$$

For example

1. When S is exponentially distributed, $n = 1$, $a_1 = 1$ and $b_1 = 1$
2. When S is χ^2 with $2m$ degrees of freedom, $n = m - 1$, $a_k = \frac{1}{k!}$ and $b_k = 1$.

Laplace trick

$$\begin{aligned}
 p_s(\theta, \lambda) &= \mathbb{P}\left(\frac{Sd^{-\alpha}}{I(o)} \geq \theta\right) = \mathbb{P}(S \geq \theta d^\alpha I(o)) \\
 &= \mathbb{E}F(S \geq \theta d^\alpha I(o)) = \sum_{k=0}^n a_k \mathbb{E}\left[(\theta d^\alpha I(o))^k \exp(-b_k \theta d^\alpha I(o))\right] \\
 &= \sum_{k=0}^n a_k (\theta d^\alpha)^k \mathbb{E}\left[I(o)^k \exp(-b_k \theta d^\alpha I(o))\right]
 \end{aligned}$$

$$\mathbb{E}[x^k e^{-x}] = \mathbb{E}\left[(-1)^k \frac{d^k}{ds^k} e^{-xs} \Big|_{s=1}\right] = (-1)^k \frac{d^k}{ds^k} \mathcal{L}_X(s) \Big|_{s=1}$$

Hence,

$$p_s(\theta, \lambda) = \sum_{k=0}^n a_k \theta^k d^{k\alpha} (-1)^k \frac{d^k}{ds^k} \mathcal{L}_{I(o)}(s) \Big|_{s=b_k \theta d^\alpha}$$

Summary: Key steps in the SINR evaluation

$$\begin{aligned}
 p_s(\theta, \lambda) &= \mathbb{P}(h \geq d^\alpha \theta (\sigma^2 + I(o))) \stackrel{(a)}{=} \mathbb{E} \exp(-d^\alpha \theta (\sigma^2 + I(o))) \\
 &= \exp(-d^\alpha \theta \sigma^2) \mathbb{E} \exp(-d^\alpha \theta I(o)) \\
 &= \exp(-d^\alpha \theta \sigma^2) \mathbb{E} \prod_{y \in \Phi} \mathcal{L}_h(d^\alpha \theta \|y\|^{-\alpha}) \\
 &= \exp(-d^\alpha \theta \sigma^2) \mathbb{E} \prod_{y \in \Phi} \frac{1}{1 + d^\alpha \theta \|y\|^{-\alpha}} \\
 &\stackrel{(b)}{=} \exp(-d^\alpha \theta \sigma^2) \exp(-\lambda d^2 \theta^{2/\alpha} C(\alpha))
 \end{aligned}$$

1. The distribution of h being exponential in (a).
2. The distribution of the fading between the interferers and the tagged receiver is not crucial.
3. Using PGFL in (b).

Interference distribution

The Laplace transform of interference $I = \sum_{y \in \Phi} h_{y0} \|y\|^{-\alpha}$ is given by

$$\mathcal{L}_I(s) = \exp(-\lambda s^{2/\alpha} C(\alpha)), \quad \alpha > 2.$$

- ▶ The Laplace transform of an alpha stable distribution with parameter $2/\alpha$.
 1. Heavy tailed distribution. Not Gaussian⁴.
 2. No closed form expression for CDF
 3. Integer moments don't exist.
 - ▶ $\mathbb{E}[I] = \lambda \int_{\mathbb{R}^2} \|x\|^{-\alpha} dx = \infty$
 - ▶ An artefact of the singularity of path loss model $\ell(x) = \|x\|^{-\alpha}$ at $x = o$.
- ▶ $I = \sum_{y \in \Phi} h_{y0} \|y\|^{-\alpha}$ is also referred as shot noise (SN) process.

⁴R.K. Ganti and M. Haenggi. "Interference in ad hoc networks with general motion-invariant node distributions", ISIT, 2008

Tail bounds on interference⁵

Evaluate the CCDF of $\mathbb{P}(I \geq y)$, where $I = \sum_{y \in \Phi} h_{y0} \|y\|^{-\alpha}$.

- ▶ Divide the point process into two sets

$$\Phi_y = \{x \in \Phi, h_{x0} \|x\|^{-\alpha} > y\}$$

$$\Phi_y^c = \{x \in \Phi, h_{x0} \|x\|^{-\alpha} \leq y\}$$

- ▶ Lower bound:

$$\begin{aligned} \mathbb{P}(I \geq y) &= \mathbb{P}(I_{\Phi_y} + I_{\Phi_y^c} \geq y) \\ &\geq \mathbb{P}(I_{\Phi_y} \geq y) = 1 - \mathbb{P}(I_{\Phi_y} \leq y) \\ &= 1 - \mathbb{P}(\Phi_y = \emptyset) \end{aligned}$$

⁵S. Weber, X. Yang, J. G. Andrews and G. de Veciana, "Transmission Capacity of Wireless Ad Hoc Networks with Outage Constraints", IEEE Transactions on Information Theory, Vol. 51, No. 12, Dec. 2005



$$\begin{aligned}
 \mathbb{P}(\Phi_y = \emptyset) &= \mathbb{E} \prod_{x \in \Phi} \mathbf{1}(h_{x0} \|x\|^{-\alpha} < y) \\
 &= \mathbb{E} \prod_{x \in \Phi} \mathbb{E}_{h_{x0}} \mathbf{1}(h_{x0} \|x\|^{-\alpha} < y) = \mathbb{E} \prod_{x \in \Phi} 1 - e^{-y \|x\|^\alpha} \\
 &= \exp \left(-\lambda \int_{\mathbb{R}^2} e^{-y \|x\|^\alpha} dx \right) = \exp \left(-\lambda y^{-2/\alpha} \pi \Gamma(1 + 2/\alpha) \right)
 \end{aligned}$$

▶ Upper bound

$$\begin{aligned}
 \mathbb{P}(I \geq y) &= \mathbb{P}(I \geq y | I_{\Phi_y} > y) \mathbb{P}(I_{\Phi_y} > y) + \mathbb{P}(I \geq y | I_{\Phi_y} \leq y) \mathbb{P}(I_{\Phi_y} \leq y) \\
 &= \mathbb{P}(I_{\Phi_y} > y) + \mathbb{P}(I \geq y | I_{\Phi_y} \leq y) \mathbb{P}(I_{\Phi_y} \leq y) \\
 &= 1 - \mathbb{P}(\Phi_y = \emptyset) + \mathbb{P}(I \geq y | I_{\Phi_y} \leq y) \mathbb{P}(\Phi_y = \emptyset) \\
 &= 1 - (1 - \mathbb{P}(I \geq y | I_{\Phi_y} \leq y)) \mathbb{P}(\Phi_y = \emptyset)
 \end{aligned}$$

► Using Markov inequality

$$\begin{aligned}
 \mathbb{P}(I \geq y | I_{\Phi_y} \leq y) &= \mathbb{P}(I \geq y | \Phi_y = \emptyset) \\
 &\leq \frac{\mathbb{E}[I \geq y | \Phi_y = \emptyset]}{y} \\
 &= \frac{1}{y} \mathbb{E} \sum_{x \in \Phi} h_{x0} \|x\|^{-\alpha} \mathbf{1}(h_{x0} \|x\|^{-\alpha} \leq y) \\
 &= \frac{\lambda}{y} \int_{\mathbb{R}^2} \|x\|^{-\alpha} \mathbb{E}[h_{x0} \mathbf{1}(h_{x0} \|x\|^{-\alpha} \leq y)] dx \\
 &= \frac{\lambda}{y} \int_{\mathbb{R}^2} \|x\|^{-\alpha} \int_0^{y\|x\|^\alpha} h e^{-h} dh dx \\
 &= \frac{2\pi\Gamma(1 + 2/\alpha)}{2 - \alpha} y^{-2/\alpha}
 \end{aligned}$$

$$1 - e^{-\lambda y^{-2/\alpha} \mathbb{E}[h^{2/\alpha}]} \leq \mathbb{P}(I \geq y) \leq 1 - \left(1 - \frac{2\pi\mathbb{E}[h^{2/\alpha}]}{2 - \alpha} y^{-2/\alpha}\right) e^{-\lambda y^{-2/\alpha} \mathbb{E}[h^{2/\alpha}]}$$

Interference is heavy tailed

Lemma

When path loss is given by $\ell(x) = \|x\|^{-\alpha}$, interference is heavy tailed

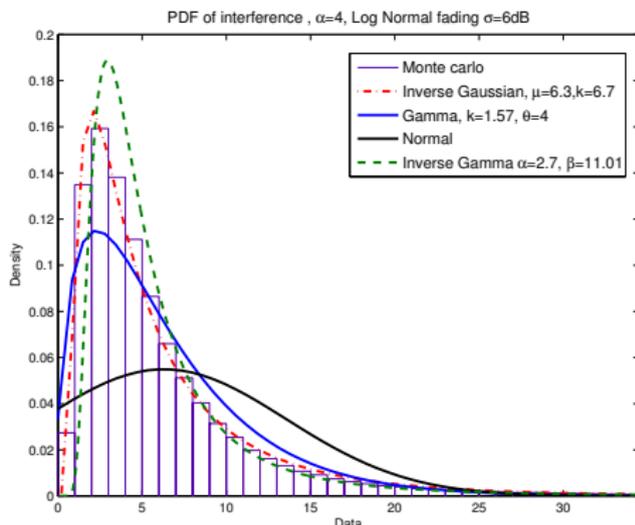
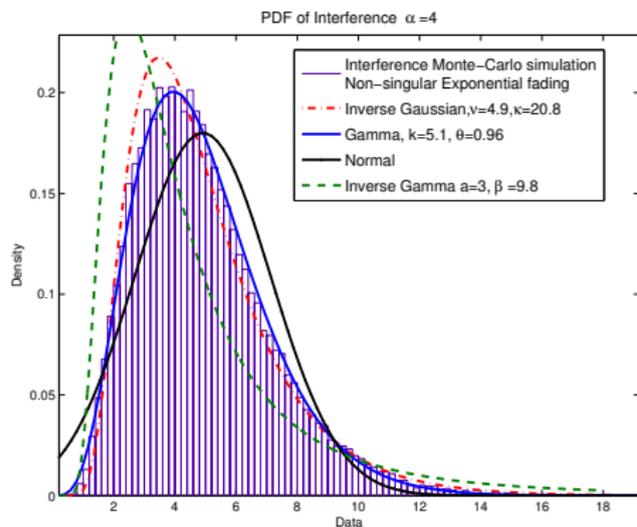
$$\mathbb{P}(I \geq y) \sim \lambda y^{-2/\alpha} \mathbb{E}[h^{2/\alpha}], \quad y \rightarrow \infty$$

Proof: We have

$$1 - e^{-\lambda y^{-2/\alpha} \mathbb{E}[h^{2/\alpha}]} \leq \mathbb{P}(I \geq y) \leq 1 - \left(1 - \frac{2\pi \mathbb{E}[h^{2/\alpha}]}{2 - \alpha} y^{-2/\alpha}\right) e^{-\lambda y^{-2/\alpha} \mathbb{E}[h^{2/\alpha}]}.$$

Use $\exp(-x) \sim 1 - x$, $x \rightarrow 0$. □

Is Gaussian modelling of interference appropriate?



PDF of interference for Rayleigh and log normal fading with path loss $\ell(x) = (1 + \|x\|^\alpha)^{-1}$.

Transmission capacity(TC)

Let $\epsilon \in (0, 1)$. TC is defined as

$$TC(\epsilon) = (1 - \epsilon) \arg \max_{\lambda > 0} \{p_s(\theta, \lambda) > 1 - \epsilon\}$$

1. $\arg \max_{\lambda > 0} \{p_s(\theta, \lambda) > 1 - \epsilon\}$ is the maximum density of transmitting nodes that can be supported for an outage constraint ϵ .
2. $1 - \epsilon$ fraction of these nodes are successful.
3. Hence TC measures the maximum spatial intensity of successful transmissions per unit area for a given outage capacity.
4. Can be related to area spectral efficiency (ASE) by multiplying with $\log_2(1 + \theta)$.

Transmission capacity of the PPP dipole network

Lemma

When $\sigma^2 \approx 0$, the TC of a PPP dipole network with $N_t = N_r = 1$ is

$$TC(\epsilon) = \frac{(1 - \epsilon)}{d^2 C(\alpha) \theta^{2/\alpha}} \ln \left(\frac{1}{1 - \epsilon} \right).$$

Proof: We have,

$$p(\theta, \lambda) = \exp(-\lambda d^2 \theta^{2/\alpha} C(\alpha))$$

Observe that $p(\theta, \lambda)$ increases with λ . Hence solving for

$$p(\theta, \lambda) > 1 - \epsilon,$$

we obtain the result. □

Sphere packing interpretation of TC

When $\epsilon \approx 0$, by Taylor series expansion, $\ln\left(\frac{1}{1-\epsilon}\right) = \epsilon + o(\epsilon)$. Hence

$$TC(\epsilon) = \frac{(1-\epsilon)}{d^2 C(\alpha) \theta^{2/\alpha}} \ln\left(\frac{1}{1-\epsilon}\right) \sim \frac{\epsilon}{d^2 C(\alpha) \theta^{2/\alpha}} = \frac{1}{\pi \left(d \sqrt{\frac{2\pi}{\epsilon \alpha \sin(2\pi/\alpha)}} \right)^2}.$$

Interpretation (heuristic)

Hence each transmission approximately requires an interference free disc of radius

$$R = d \left(\frac{2\pi}{\epsilon \alpha \sin(2\pi/\alpha)} \right)^{1/2}.$$

- ▶ The disc radius increases as $\frac{1}{\sqrt{\epsilon}}$.
- ▶ The disc radius decreases with increasing α
 - ▶ Higher path loss exponent \rightarrow better packing.

Transmission capacity for other schemes when $\epsilon \approx 0$

- ▶ SISO: $TC(\epsilon) = \frac{\epsilon}{d^2 C(\alpha) \theta^{2/\alpha}}$
- ▶ MIMO: MRC with n receive antenna

$$\frac{n^{2/\alpha} \epsilon}{d^2 C(\alpha) \theta^{2/\alpha}} \leq TC(\epsilon) \leq \frac{n^{2/\alpha} \Gamma(1 - 2/\alpha) \epsilon}{d^2 C(\alpha) \theta^{2/\alpha}}$$

- ▶ MIMO eigen-beamforming: m transmit and n receive

$$\frac{\max\{n, m\}^{2/\alpha} \epsilon}{d^2 C(\alpha) \theta^{2/\alpha}} \leq TC(\epsilon) \leq \frac{(nm)^{2/\alpha} \Gamma(1 - 2/\alpha) \epsilon}{d^2 C(\alpha) \theta^{2/\alpha}}$$

Can be used to analyse a multitude of systems with interference.

A. M. Hunter, J. G. Andrews and S. P. Weber, "Transmission Capacity of Ad Hoc Networks with Spatial Diversity",
 IEEE Transactions on Wireless Communications, Vol. 7, No. 12, pp. 5058-71, Dec. 2008

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▶ MIMO

R.H.Y. Louie, M. R. McKay, I. B. Collings , "Open-Loop Spatial Multiplexing and Diversity Communications in Ad Hoc Networks",/ IEEE Transactions on Information Theory, Vol. 57, No.1, pp. 317-344, Jan. 2011

▶ Power control

N. Jindal, S. Weber, and J. Andrews, "Fractional Power Control for Decentralized Wireless Networks, IEEE Trans. Wireless Communications, Vol. 7, No. 12, pp. 5482-5492, Dec. 2008

X. Zhang and M. Haenggi, "Random Power Control in Poisson Networks," IEEE Transactions on Communications, 2012.

▶ Multihop

R. Vaze, "Throughput-Delay-Reliability Tradeoff with ARQ in Wireless Ad Hoc Networks", Wireless Communications, IEEE Transactions on , vol.10, no.7, pp.2142-2149, July 2011

Monographs

F. Baccelli and B. Blaszczyszyn "Stochastic Geometry and Wireless Networks" NOW Publishers

S. Weber and J. G. Andrews, "Transmission Capacity of Wireless Networks", NOW Publishers

M. Haenggi and R. K. Ganti, "Interference in Large Wireless Networks", NOW Publishers

Spatial and temporal correlation of interference in PPPs with ALOHA

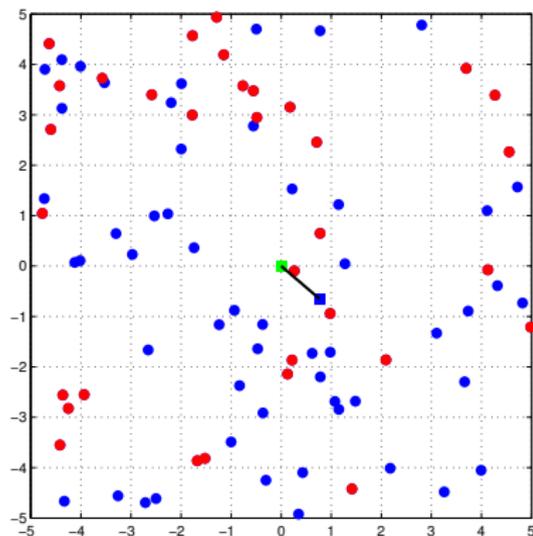
- ▶ At each time instant each node transmits with probability p
- ▶ Let Φ_k denote the set of active transmitters at time k . *i.e.*,

$$\Phi_k = \{x; x \in \Phi, x \text{ is on at time } k\}$$

- ▶ The interference is

$$I(\Phi_k, z) = \sum_{x \in \Phi_k} h_{xz}[k] \ell(x - z)$$

- ▶ We assume that the fading is independent across space and time.



Time instant 1.

Spatial and temporal correlation of interference in PPPs with ALOHA

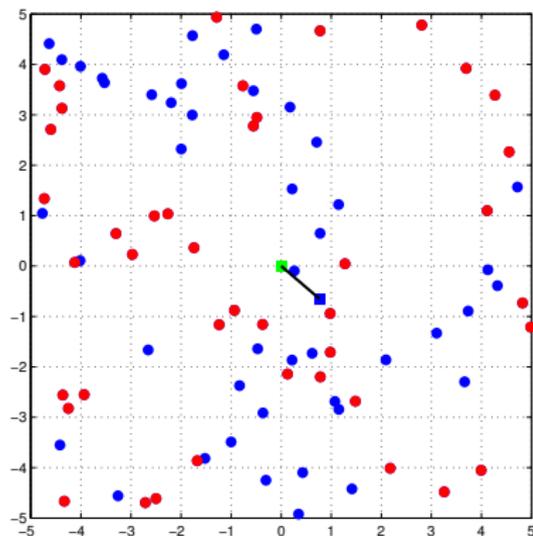
- ▶ At each time instant each node transmits with probability p
- ▶ Let Φ_k denote the set of active transmitters at time k . *i.e.*,

$$\Phi_k = \{x; x \in \Phi, x \text{ is on at time } k\}$$

- ▶ The interference is

$$I(\Phi_k, z) = \sum_{x \in \Phi_k} h_{xz}[k] \ell(x - z)$$

- ▶ We assume that the fading is independent across space and time.



Time instant 2.

Spatial and temporal correlation of interference in PPPs with ALOHA

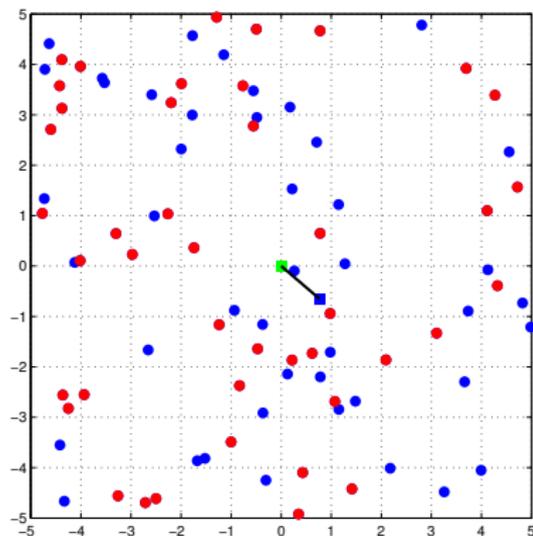
- ▶ At each time instant each node transmits with probability p
- ▶ Let Φ_k denote the set of active transmitters at time k . *i.e.*,

$$\Phi_k = \{x; x \in \Phi, x \text{ is on at time } k\}$$

- ▶ The interference is

$$I(\Phi_k, z) = \sum_{x \in \Phi_k} h_{xz}[k] \ell(x - z)$$

- ▶ We assume that the fading is independent across space and time.



Time instant 3.

Spatio-temporal correlation coefficient

$$\text{Correlation coefficient, } \rho_{X,Y} = \frac{\text{cov}(X, Y)}{\sigma_X \sigma_Y}.$$

Lemma

The spatio-temporal correlation coefficient of the interferences $I(\Phi_m, u)$ and $I(\Phi_n, v)$, $m \neq n$ for ALOHA and path loss functions $\ell(x)$ satisfying $\int_{\mathbb{R}^2} \ell(x) dx < \infty$, is

$$\zeta(u, v) = \frac{\rho \int_{\mathbb{R}^2} \ell(x) \ell(x - \|u - v\|) dx}{\mathbb{E}[h^2] \int_{\mathbb{R}^2} \ell^2(x) dx}.$$

Proof: Follows⁶ from Campbell's theorem. \square

⁶R. K. Ganti and M. Haenggi. "Spatial and temporal correlation of the interference in ALOHA ad hoc networks", IEEE Communications Letters, 13(9):631 -633, September 2009.

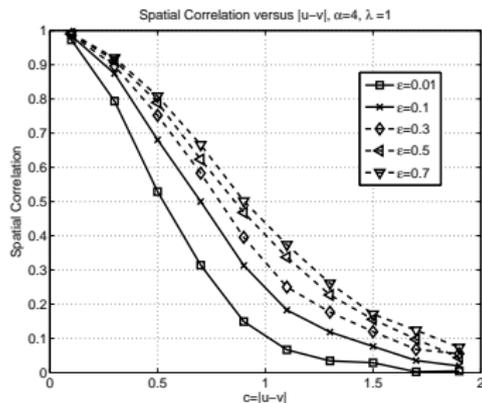
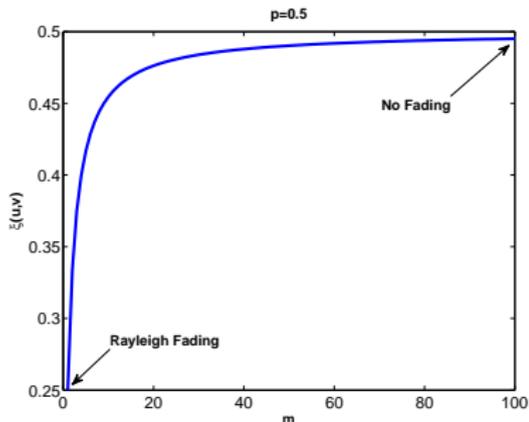
When $u = v$, the temporal correlation is equal to

$$\frac{\rho}{\mathbb{E}[h^2]}.$$

Hence for Nakagami- m fading, it is equal to $\frac{\rho m}{1+m}$.

When the path loss is given by $\ell(x) = 1/(\epsilon + \|x\|^\alpha)$ and $u \neq v$ the correlation is equal to

$$\lim_{\epsilon \rightarrow 0} \xi(u, v) = 0.$$



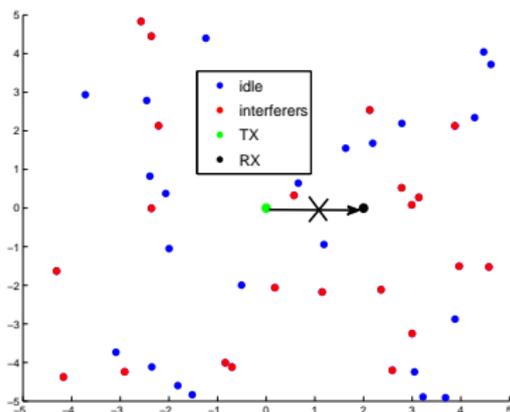
Interference is spatio-temporally correlated.

Link formation delay

- ▶ A TX at x can connect to a RX at y if

$$\text{SIR}(x, y) = \frac{h_{xy}[k]\ell(d)}{I(\Phi_k, y)} \geq \theta$$

- ▶ ALOHA MAC with access probability p
- ▶ D : No of attempts required for a connection to form.



First attempt

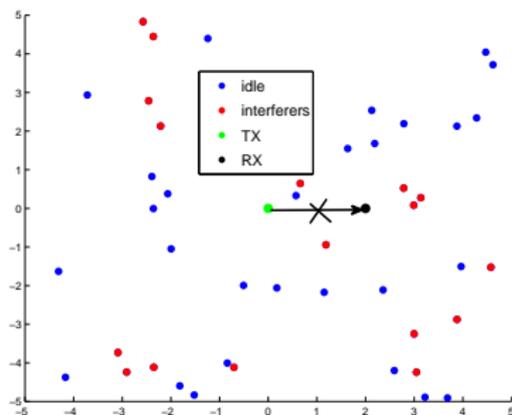
No connection

Link formation delay

- ▶ A TX at x can connect to a RX at y if

$$\text{SIR}(x, y) = \frac{h_{xy}[k]\ell(d)}{I(\Phi_k, y)} \geq \theta$$

- ▶ ALOHA MAC with access probability p
- ▶ D : No of attempts required for a connection to form.



Second attempt

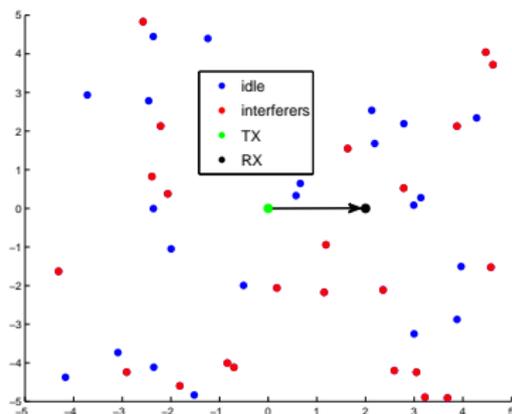
No connection

Link formation delay

- ▶ A TX at x can connect to a RX at y if

$$\text{SIR}(x, y) = \frac{h_{xy}[k]\ell(d)}{I(\Phi_k, y)} \geq \theta$$

- ▶ ALOHA MAC with access probability p
- ▶ D : No of attempts required for a connection to form.



Third attempt

Connected

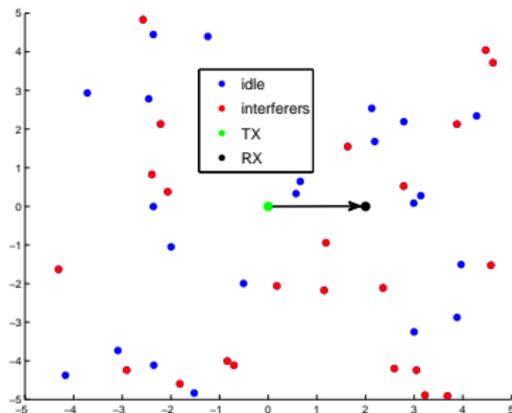
$$D = 3$$

Link formation delay

- ▶ A TX at x can connect to a RX at y if

$$\text{SIR}(x, y) = \frac{h_{xy}[k]\ell(d)}{I(\Phi_k, y)} \geq \theta$$

- ▶ ALOHA MAC with access probability p
- ▶ D : No of attempts required for a connection to form.



What is the average delay $\mathbb{E}[D]$?

Average delay $\mathbb{E}[D]$: Neglecting interference correlation

Recall

- ▶ ALOHA corresponds to independent thinning
 - ▶ Φ_m (the set of transmitters at time m) is a PPP of density λp .
- ▶ The probability of link formation in a PPP network with density $p\lambda$ is

$$p_s(\theta, p\lambda) = \exp(-p\lambda d^2 \theta^{2/\alpha} C(\alpha))$$

1. At each time instant a link is formed with probability $p_s(\theta, \lambda)$ independent of every other time
2. So the delay D is a geometric random variable with mean $\frac{1}{p_s(\theta, \lambda)}$.

$$\mathbb{E}[D] = \frac{1}{p_s(\theta, \lambda)} = \exp(p\lambda d^2 \theta^{2/\alpha} C(\alpha))$$

3. Observe that the delay increases with p .

Average delay $\mathbb{E}[D]$: With interference correlation

- ▶ Let E_k denote the event $\frac{h[k]\ell(d)}{I(\Phi_k, o)} \geq \theta$
- ▶ Then $\mathbb{P}(D > k) = \mathbb{P}(E_1^c \cap E_2^c \cap \dots \cap E_k^c)$ Fail for k times
- ▶ $\mathbb{E}[D] = \sum_{k=0}^{\infty} \mathbb{P}(D > k)$ Average of a positive random variable
 - ▶ $\mathbb{E}[D] = \mathbb{E}_{\Phi} \mathbb{E}[D|\Phi] = \mathbb{E}_{\Phi} [\sum_{k=0}^{\infty} \mathbb{P}(D > k|\Phi)]$ Conditioning on Φ
- ▶ $\mathbb{P}(D > k|\Phi) = \mathbb{P}(E_1^c|\Phi)\mathbb{P}(E_2^c|\Phi)\dots\mathbb{P}(E_k^c|\Phi)$ Conditional Independence
- ▶ Probability that a link is not formed at time m is

$$\begin{aligned} \mathbb{P}(E_m^c|\Phi) &= \mathbb{P}\left(\frac{h[m]d^{-\alpha}}{I(\Phi_m, y)} \leq \theta \mid \Phi\right) \\ &= 1 - \underbrace{\mathbb{E}[\exp(-d^\alpha \theta I(\Phi_m, o)) | \Phi]}_{T_1} \end{aligned}$$

- ▶ Two sources on randomness: Fading and ALOHA MAC

$$\begin{aligned}
 T_1 &= \mathbb{E} e^{-d^\alpha \theta \sum_{x \in \Phi} h_{x_0}[k] \|x\|^{-\alpha} \mathbf{1}(x \text{ is Tx at time } m)} \\
 &= \mathbb{E} \prod_{x \in \Phi} e^{-d^\alpha \theta h_{x_0}[k] \|x\|^{-\alpha} \mathbf{1}(x \text{ is Tx})} \\
 &= \mathbb{E} \prod_{x \in \Phi} \left[e^{-d^\alpha \theta h_{x_0}[k] \|x\|^{-\alpha} \mathbf{1}(x \text{ is Tx})} + 1 - \mathbf{1}(x \text{ is Tx}) \right]
 \end{aligned}$$

- ▶ First averaging over ALOHA,

$$T_1 = \mathbb{E} \prod_{x \in \Phi} \left[e^{-d^\alpha \theta h_{x_0}[k] \|x\|^{-\alpha} p} + 1 - p \right]$$

- ▶ Averaging over fading,

$$\mathbb{P}(E_m^c | \Phi) = 1 - T_1 = 1 - \prod_{x \in \Phi} \left[1 - \frac{p}{1 + d^\alpha \theta \|x\|^\alpha} \right]$$

Observe that $\mathbb{P}(E_m^c | \Phi)$ does not depend on the time index m .

$$\begin{aligned}\mathbb{P}(D > k|\Phi) &= \mathbb{P}(E_1^c|\Phi)\mathbb{P}(E_2^c|\Phi)\dots\mathbb{P}(E_k^c|\Phi) \\ &= \left(1 - \prod_{x \in \Phi} \left[1 - \frac{p}{1 + d^{\alpha}\theta\|x\|^{\alpha}}\right]\right)^k\end{aligned}$$

- Hence the conditional average of delay is

$$\begin{aligned}\mathbb{E}[D|\Phi] &= \sum_{k=0}^{\infty} \mathbb{P}(D > k|\Phi) = \sum_{k=0}^{\infty} \left(1 - \prod_{x \in \Phi} \left[1 - \frac{p}{1 + d^{\alpha}\theta\|x\|^{\alpha}}\right]\right)^k \\ &= \frac{1}{\prod_{x \in \Phi} \left[1 - \frac{p}{1 + d^{\alpha}\theta\|x\|^{\alpha}}\right]} \\ &= \prod_{x \in \Phi} \left[1 - \frac{p}{1 + d^{\alpha}\theta\|x\|^{\alpha}}\right]^{-1}\end{aligned}$$

Using PGFL of a PPP,

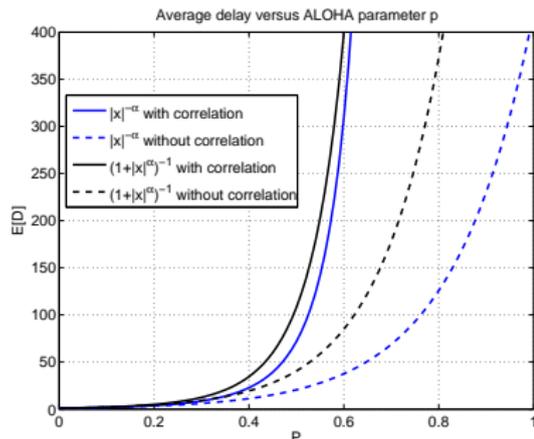
$$\mathbb{E} \prod_{x \in \Phi} \left[1 - \frac{p}{1 + d^{\alpha} \theta \|x\|^{\alpha}} \right]^{-1} \\ = \exp \left(-\lambda \int_{\mathbb{R}^2} 1 - \left[1 - \frac{p}{1 + d^{\alpha} \theta \|x\|^{\alpha}} \right]^{-1} dx \right)$$

With correlation

$$\mathbb{E}[D] = \exp \left(\frac{p \lambda d^2 \theta^{2/\alpha} C(\alpha)}{(1-p)^{1-2/\alpha}} \right)$$

Observe $\mathbb{E}[D] = \infty$ for $p = 1$.

- ▶ Relying on fading is not sufficient for ARQ to succeed.
- ▶ **Correlation of interference cannot be neglected.**



Recall: Without considering correlation

$$\mathbb{E}[D] = \exp \left(p \lambda d^2 \theta^{2/\alpha} C(\alpha) \right)$$

F. Baccelli, B. Baszczyszyn, "A New Phase Transitions for Local Delays in MANETs," INFOCOM, 2010

Proceedings IEEE, vol., no., pp.1-9, 14-19 March 2010

Analysis of Cellular Networks

* Primer on Point Processes

* Ad hoc Networks

* Cellular Networks

– SINR distribution

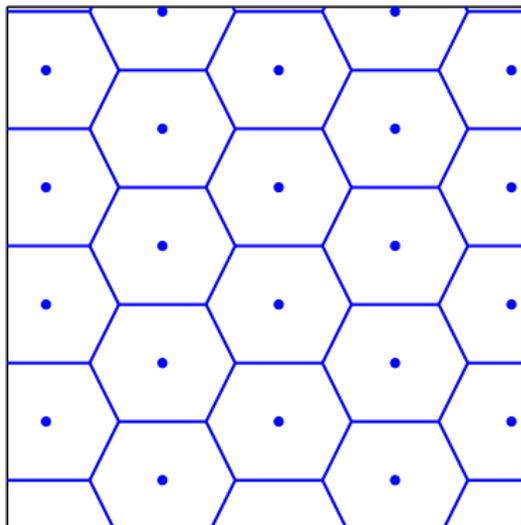
– Frequency reuse

* Heterogeneous Networks

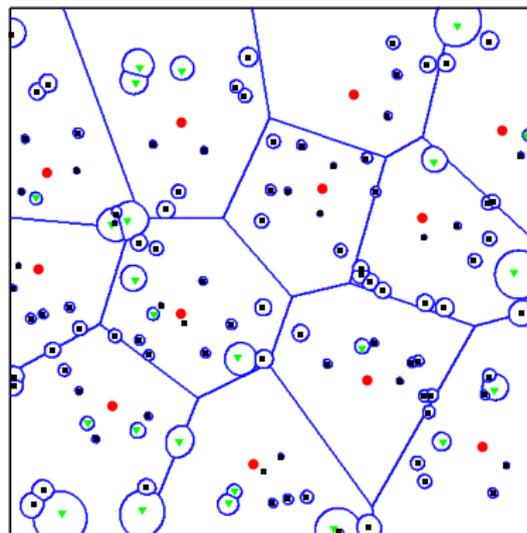
Cellular Trends

- ▶ Interference is a main challenge in cellular systems
 1. SINR more important than SNR
 - ▶ Universal frequency reuse
 - ▶ Denser and denser deployments
 2. BS cooperation and other interference-suppression techniques require good models for other-cell interference
- ▶ Networks are becoming unplanned, decentralized and heterogeneous
 - ▶ Picocells placed strategically in high-traffic areas
 - ▶ Femtocells/relays being placed randomly
 - ▶ BS deployments increasingly driven by capacity needs rather than coverage needs

Emerging cellular networks



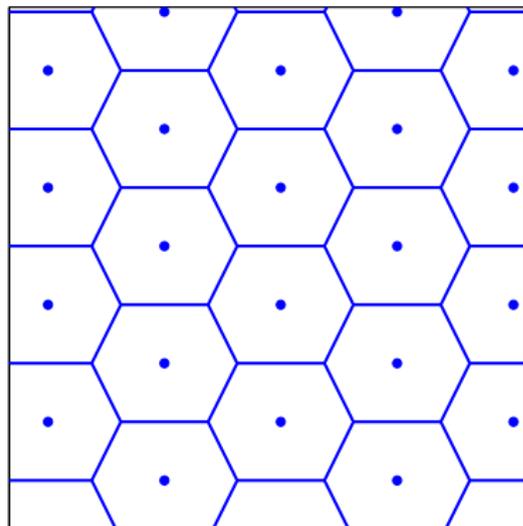
What we think of



4G+femto/pico

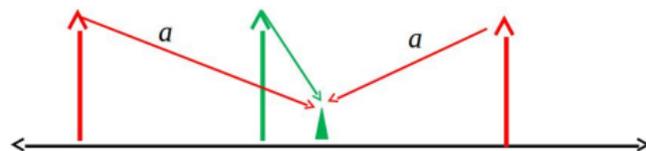
Cells getting smaller, more random and chaotic

Some current models



Hexagonal model

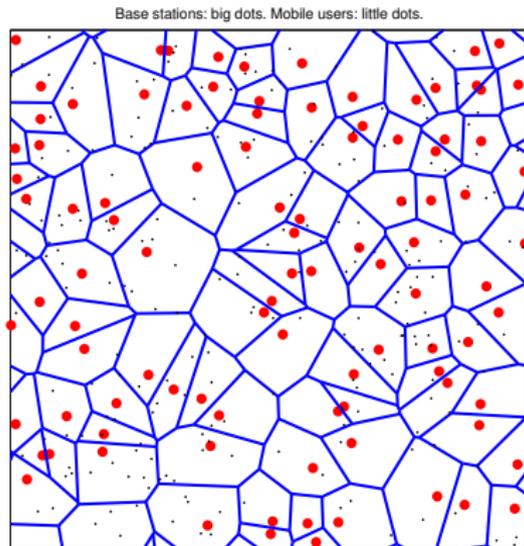
1. Is it accurate?
2. Not very tractable.



Wyner model

1. Fixed background interference
2. Highly inaccurate (averaging)

Proposed Model: PPP Base Stations



- ▶ BS locations are drawn from a PPP of density λ
- ▶ Each mobile associates with the closest BS
 - ▶ Cells are Voronoi tessellations

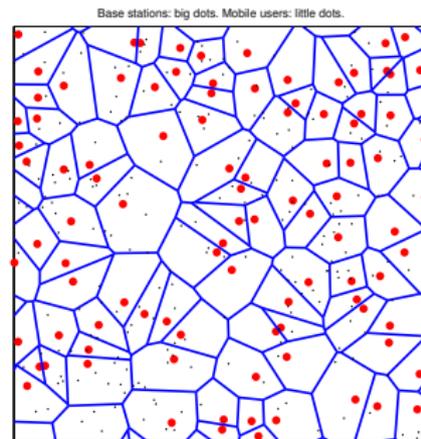
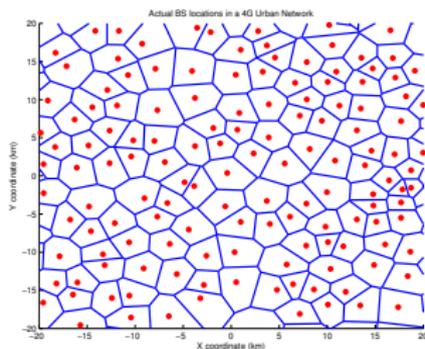
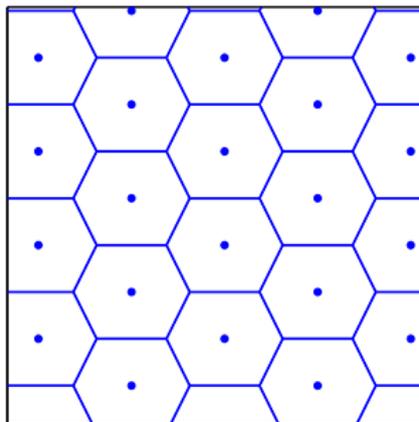
Advantages

- ▶ Non uniform cell sizes
- ▶ Tractable?

Disadvantages

- ▶ BSs might get very close

Comparison with real deployment



System model: Downlink

- ▶ The BSs are spatially distributed as a PPP of density λ
- ▶ Mobile users connect to the nearest (geographical) BS
- ▶ The path loss is given by $\ell(x) = \|x\|^{-\alpha}$, $\alpha > 2$.
- ▶ All BSs transmit at the same power P
- ▶ The fading between a BS x and a mobile y is denoted by h_{xy}

What is the SINR distribution of a typical mobile user?

Without loss of generality, we can assume the typical mobile user to be at the origin o .

Analysis of SINR

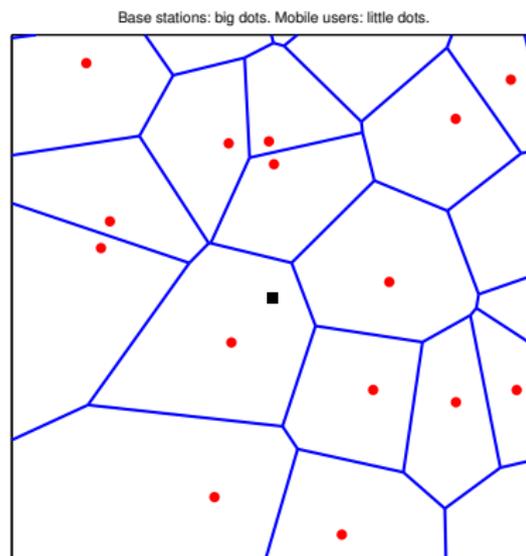
Let $x \in \Phi$ be the BS that is closest to the reference MS at the origin.

The downlink SINR of the MS at the origin is

$$\text{SINR} = \frac{h_{x0} r^{-\alpha}}{\sum_{y \in \Phi \setminus \{x\}} h_{y0} \|y\|^{-\alpha}}$$

where $r = \|x\|$.

Compute $\mathbb{P}(\text{SINR} > \theta)$



Analysis of SINR

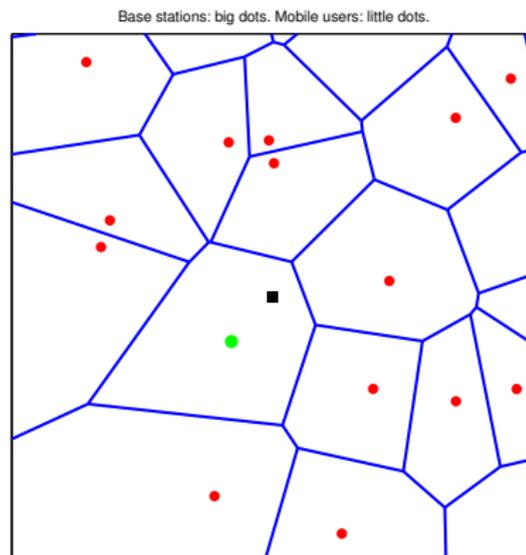
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where $r = \|x\|$.

Compute $\mathbb{P}(\text{SINR} > \theta)$



Analysis of SINR

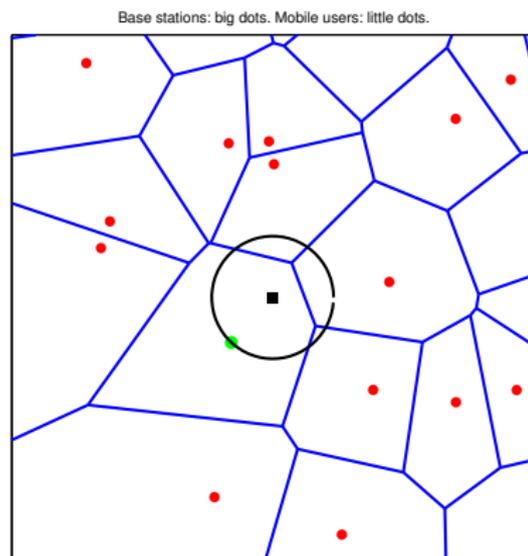
Let $x \in \Phi$ be the BS that is closest to the reference MS at the origin.

The downlink SINR of the MS at the origin is

$$\text{SINR} = \frac{h_{x0}r^{-\alpha}}{\sum_{y \in \Phi \setminus \{x\}} h_{y0}\|y\|^{-\alpha}}$$

where $r = \|x\|$.

Compute $\mathbb{P}(\text{SINR} > \theta)$



Analysis of SINR

Recall that the distribution of the nearest BS equals the first contact distribution

$$f(r) = \lambda 2\pi r \exp(-\lambda\pi r^2)$$

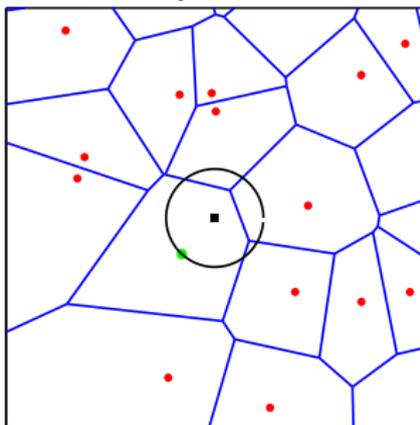
- ▶ We first condition on the distance to the nearest BS r .

$$\begin{aligned} \mathbb{P}(\text{SINR} > \theta) &= \mathbb{E}_r \mathbb{P}(\text{SINR} > \theta | r) \\ &= \int_0^\infty \mathbb{P}(\text{SINR} > \theta | r) \lambda 2\pi r \exp(-\lambda\pi r^2) dr \end{aligned}$$

- ▶ Focusing on $\mathbb{P}(\text{SINR} > \theta | r)$ which we denote by p_r

$$\begin{aligned} p_r &= \mathbb{P} \left(\frac{h_{x_0} r^{-\alpha}}{\sum_{y \in \Phi \setminus \{x\}} h_{y_0} \|y\|^{-\alpha}} \geq \theta \mid r \right) \\ &= \mathbb{E} \left[\exp \left(-\theta r^\alpha \sum_{y \in \Phi \setminus \{x\}} h_{y_0} \|y\|^{-\alpha} \right) \mid r \right] \end{aligned}$$

Base stations: big dots. Mobile users: little dots.



$$p_r = \mathbb{E} \left[\prod_{y \in \Phi \setminus \{x\}} \exp(-\theta r^\alpha h_{y_0} \|y\|^{-\alpha}) \mid r \right]$$

- ▶ Since the fades are independent and exponentially distributed with unit mean,

$$p_r = \mathbb{E} \left[\prod_{y \in \Phi \setminus \{x\}} \frac{1}{1 + \theta r^\alpha \|y\|^{-\alpha}} \mid r \right]$$

- ▶ Using PGFL on $\Phi \cap B(o, r)^c$

$$p_r = e^{-\lambda \int_{B(o, r)^c} \frac{1}{1 + \theta r^\alpha \|y\|^{-\alpha}} dy}.$$

- ▶ Un-conditioning on r ,

$$\mathbb{P}(\text{SINR} > \theta) = \int_0^\infty e^{-\lambda \int_{B(o, r)^c} \frac{1}{1 + \theta r^\alpha \|y\|^{-\alpha}} dy} f(r) dr.$$

SINR distribution

The SIR distribution in a PPP BS network where a MS connects to the nearest BS is given by⁷

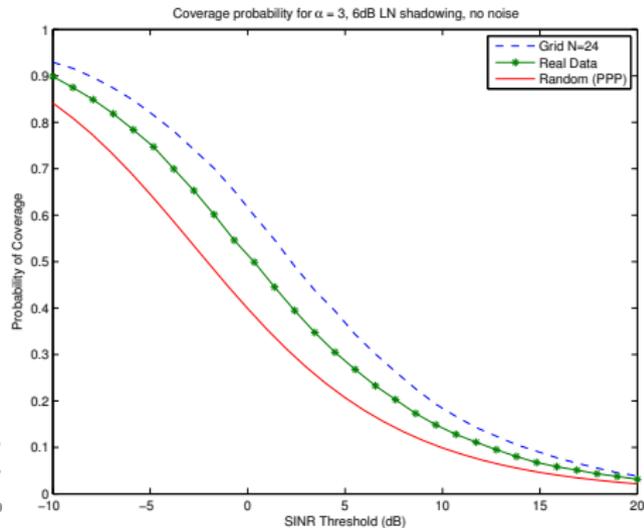
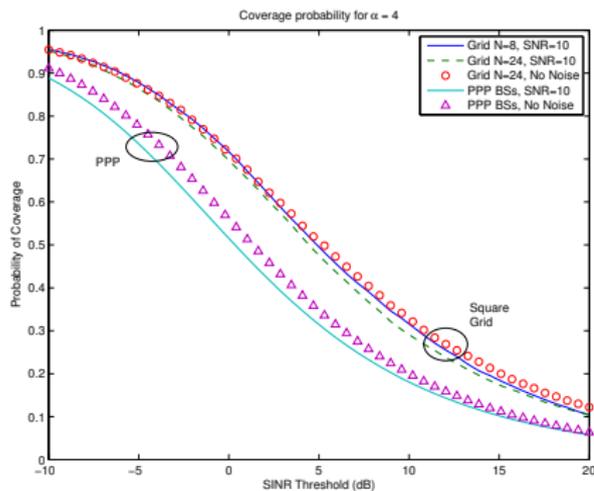
$$\mathbb{P}(\text{SIR} \geq \theta) = \frac{1}{1 + \rho(\theta, \alpha)}$$

where $\rho(\theta, \alpha) = \theta^{2/\alpha} \int_{\theta^{-2/\alpha}}^{\infty} \frac{1}{1+u^{\alpha/2}} du$.

- ▶ Does not depend on the density of BSs λ .
 - ▶ Increasing the density of BSs does not increase the coverage probability.
- ▶ Simple expression
 - ▶ For $\alpha = 4$ reduces to $(1 + \sqrt{\theta}(\pi/2 - \arctan(1/\sqrt{\theta})))^{-1}$.
- ▶ Extensions to general fading distributions and noise possible.
- ▶ Since the PDF of SIR is known the average ergodic can be computed
 - ▶ For $\alpha = 4$, the computed ergodic rate is 1.49nats/sec/Hz

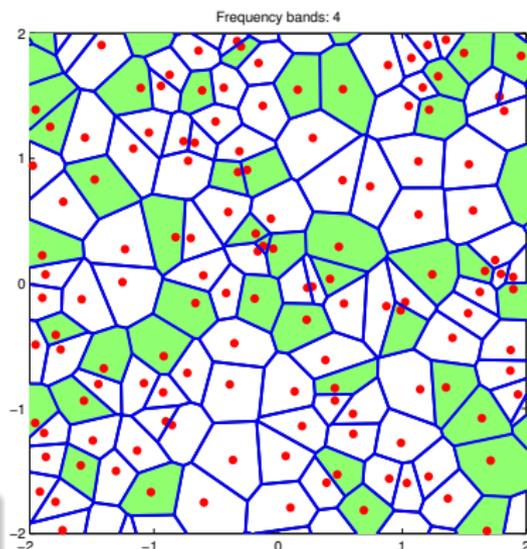
⁷J. G. Andrews, F. Baccelli, and R. K. Ganti. A tractable approach to coverage and rate in cellular networks. IEEE Trans. on Communications, Nov. 2011

Numerical results



Frequency reuse

- ▶ Frequency planning necessary for increasing the coverage.
- ▶ Assume that there are δ frequency bands
 - ▶ PPP: random allocation of bands
 - ▶ Corresponds to thinning of a PPP Φ
 - ▶ Density of Φ_m is λ/δ .
- ▶ We first consider the BS $x \in \Phi$ to which the MS at the origin connects.
 - ▶ $\|x\| = r \sim \lambda 2\pi r \exp(-\lambda\pi r^2)$
 - ▶ The interferers density is now λ/δ .



The coverage probability is

$$\mathbb{P}(\text{SIR} \geq \theta) = \frac{1}{1 + \frac{1}{\delta}\rho(\theta, \alpha)}$$

Coverage versus rate

1. The coverage probability is

$$\mathbb{P}(\text{SIR} \geq \theta) = \frac{1}{1 + \frac{1}{\delta}\rho(\theta, \alpha)} \quad \Uparrow \delta$$

2. The ergodic rate equals $\frac{1}{\delta}\mathbb{E}[\ln(1 + \text{SIR})]$, which equals

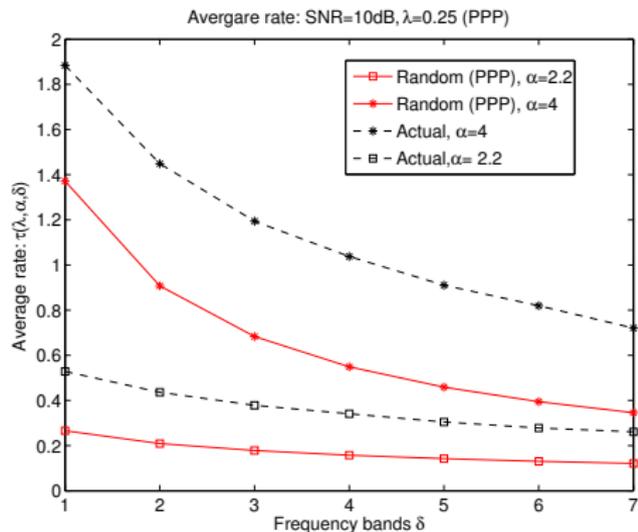
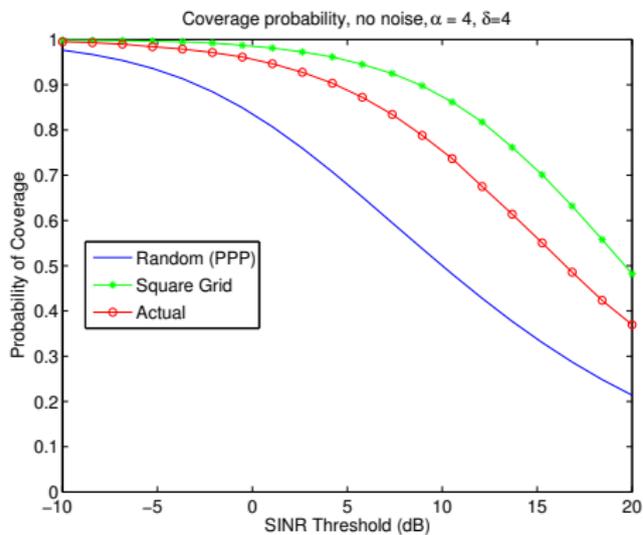
$$\frac{1}{\delta} \int_0^\infty \mathbb{P}(\ln(1 + \text{SIR}) > \theta) d\theta = \frac{1}{\delta} \int_0^\infty \frac{1}{1 + \frac{1}{\delta}\rho(e^\theta - 1, \alpha)} d\theta,$$

which equals

$$R = \int_0^\infty \frac{1}{\delta + \rho(e^\theta - 1, \alpha)} d\theta \quad \Downarrow \delta$$

Coverage increases with δ while the average rate decreases with δ .

Numerical results

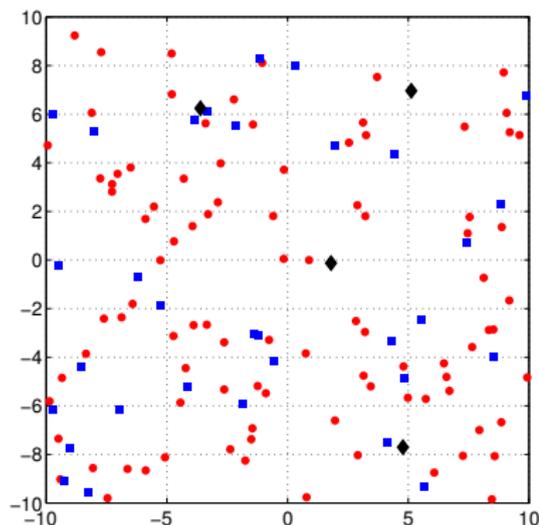


Analysis of Heterogeneous Networks

- * Primer on Point Processes
- * Ad hoc Networks
- * Cellular Networks
- * Heterogeneous Networks

System Model

- ▶ K tier network
- ▶ The BS locations in tier i are modelled by a PPP Φ_i
 - ▶ Density of Φ_i is λ_i
 - ▶ BSs in tier i transmit with power P_i
 - ▶ A mobile can connect to a BS in tier i if the received SIR is greater than θ_i
- ▶ All tiers transmit in the same frequency band and hence contribute to interference
- ▶ Fading is i.i.d Rayleigh
- ▶ Path loss is given by $\|x\|^{-\alpha}$, $\alpha > 2$



An example of $k = 3$ network.
 Red: FemtoBS, blue: PicoBS,
 black: MacroBS

Assumption

The SIR thresholds $\theta_i > 1$.

Max SIR model

Connectivity model

A mobile user can connect to a BS of any tier provided that the SIR constraint is satisfied, *i.e.*, to connect to a BS of tier i , the SIR should be greater than θ_i .

Lemma

If $\theta_i > 1$, a mobile user can connect to at most one BS

Proof.

Let a_i denote the received power from a BS i , then only one of the following terms can be greater than 1

$$\frac{a_i}{\sum_{j \neq i} a_j}, \quad i = 1, 2, \dots$$



Coverage probability

The receive SIR of a mobile at origin and a BS $x \in \Phi_m$ is

$$\text{SIR}(x) = \frac{h_{x0} \|x\|^{-\alpha}}{\sum_{i=1}^K \sum_{y \in \Phi_i} h_{y0} \|y\|^{-\alpha} - h_{x0} \|x\|^{-\alpha}}$$

Let p_c denote the coverage probability.

$$\begin{aligned} 1 - p_c &= \mathbb{E} \left(\prod_{m=1}^K \prod_{x \in \Phi_m} \mathbf{1}(\text{SIR}(x) < \theta_m) \right) \\ &= \mathbb{E} \left(\prod_{m=1}^K \prod_{x \in \Phi_m} 1 - \mathbf{1}(\text{SIR}(x) > \theta_m) \right) \end{aligned}$$

Contd...

Expanding the inner product,

$$1 - p_c = 1 - \mathbb{E} \sum_{m=1}^K \sum_{x \in \Phi_m} \mathbf{1}(\text{SIR}(x) > \theta_m) \\ + \mathbb{E} \underbrace{\sum \mathbf{1}(\text{SIR}(x) > \theta_m) \mathbf{1}(\text{SIR}(y) > \theta_n)}_{T_2} - (\text{three terms}) \dots$$

The term T_2 and higher order terms are zero since the MS can connect to at most 1 BS. Hence

$$p_c = \sum_{m=1}^K \mathbb{E} \sum_{x \in \Phi_m} \mathbf{1}(\text{SIR}(x) > \theta_m)$$

Contd...

How to evaluate $\mathbb{E} \sum_{x \in \Phi_m} \mathbf{1}(\text{SIR}(x) > \theta_m)$.

Recall Campbell Mecke theorem

For a PPP of density λ

$$\mathbb{E} \sum_{x \in \Phi} f(x, \Phi \setminus \{x\}) = \lambda \int_{\mathbb{R}^2} \mathbb{E}[f(x, \Phi)] dx$$

$$\mathbb{E} \sum_{x \in \Phi_m} \mathbf{1}(\text{SIR}(x) > \theta_m) = \lambda_m \int_{\mathbb{R}^2} \mathbb{P}(\text{SIR}(x) > \theta_m) dx$$

As before,

$$\mathbb{P} \left(\frac{P_m h_{x_0} \|x\|^{-\alpha}}{I} > \theta_m \right) = \mathcal{L}_I(\|x\|^\alpha \theta_m P_m^{-1}).$$

Contd...

Observe that the total interference is the sum of the interference from each tier which are independent. Hence

$$\mathcal{L}_I(\|x\|^\alpha \theta_m P_m^{-1}) = \prod_{j=1}^K \mathcal{L}_{I_j}(\|x\|^\alpha \theta_m P_m^{-1})$$

Recall that the Laplace transform of interference $I_j = \sum_{x \in \Phi_j} P_j h_{x0} \|x\|^{-\alpha}$ is

$$\mathcal{L}_{I_j}(s) = \exp(-\lambda_j P_j^{2/\alpha} s^{2/\alpha} C(\alpha)),$$

where $C(\alpha) = \frac{2\pi^2}{\alpha \sin(2\pi/\alpha)}$

Hence

$$\mathcal{L}_I(\|x\|^\alpha \theta_m P_m^{-1}) = \exp(-\|x\|^2 \theta_m^{2/\alpha} P_m^{-2/\alpha} \sum_{i=1}^K \lambda_i P_i^{2/\alpha} C(\alpha))$$

Coverage results⁸

Combining everything,

$$p_c = \sum_{m=1}^k \lambda_m \int_{\mathbb{R}^2} \exp(-\|x\|^2 \theta_m^{2/\alpha} P_m^{-2/\alpha} \sum_{i=1}^K \lambda_i P_i^{2/\alpha} C(\alpha)) dx$$

Lemma

The coverage probability in a K tier heterogeneous network is

$$p_c = \frac{\pi}{C(\alpha)} \frac{\sum_{m=1}^K \lambda_m P_m^{2/\alpha} \theta_m^{-2/\alpha}}{\sum_{m=1}^K \lambda_m P_m^{2/\alpha}}, \quad \theta_i > 1$$

1. A simple expression. Convex combination of $\theta_m^{-2/\alpha}$
2. p_c does not depend on the densities if all the thresholds θ_m are equal
 - ▶ Can add more tiers without changing the coverage

⁷ H. Dhillon, R. K. Ganti, F. Baccelli, and J. G. Andrews. "Modelling and analysis of k-tier downlink heterogeneous cellular networks". IEEE JSAC, April 2012

Fraction of users connected to j -th tier is

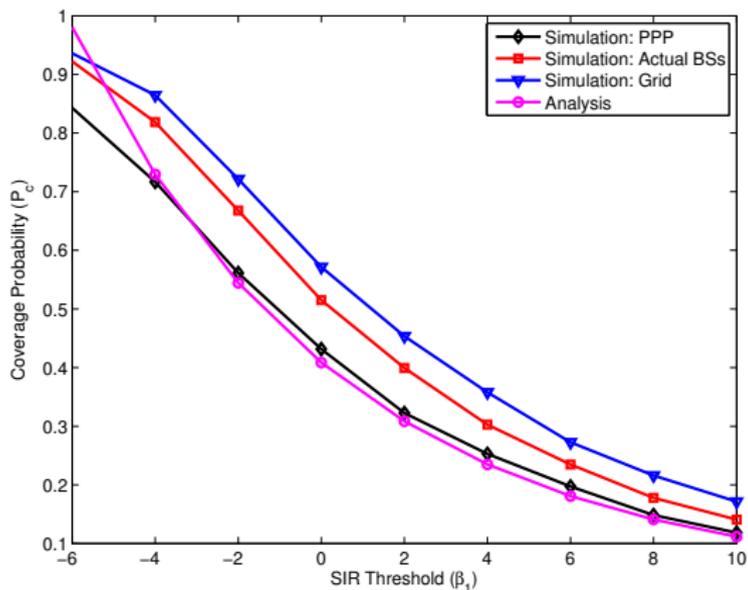
$$\beta_j = \frac{\lambda_j P_j^{2/\alpha} \theta_j^{-2/\alpha}}{\sum_{m=1}^K \lambda_m P_m^{2/\alpha} \theta_m^{-2/\alpha}}$$

Lemma (Closed access)

When the user is allowed to connect to only a subset of tiers $B \subset \{1, 2, 3, \dots, K\}$, the coverage probability is

$$p_c = \frac{\pi}{C(\alpha)} \frac{\sum_{m \in B} \lambda_m P_m^{2/\alpha} \theta_m^{-2/\alpha}}{\sum_{m=1}^K \lambda_m P_m^{2/\alpha}}, \quad \theta_i > 1$$

Numerical results



A two-tier HCN, $K = 2$, $\alpha = 3$, $P_1 = 100P_2$, $\lambda_2 = 2\lambda_1$, $\beta_2 = 1\text{dB}$

Conclusions

- ▶ Networks are getting more random and chaotic
- ▶ Random spatial models are necessary for modelling current networks.
- ▶ A rich set of mathematical tools are provided by stochastic geometry