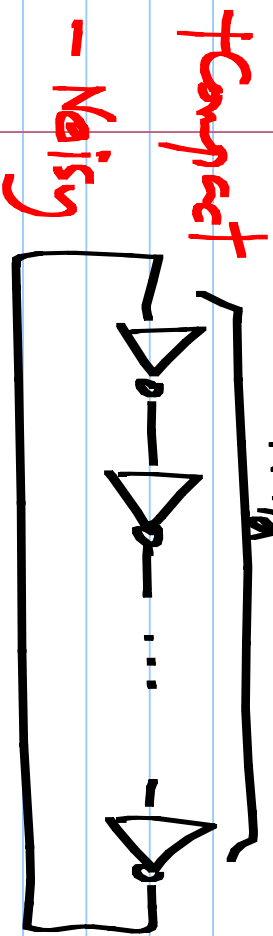


EE6322

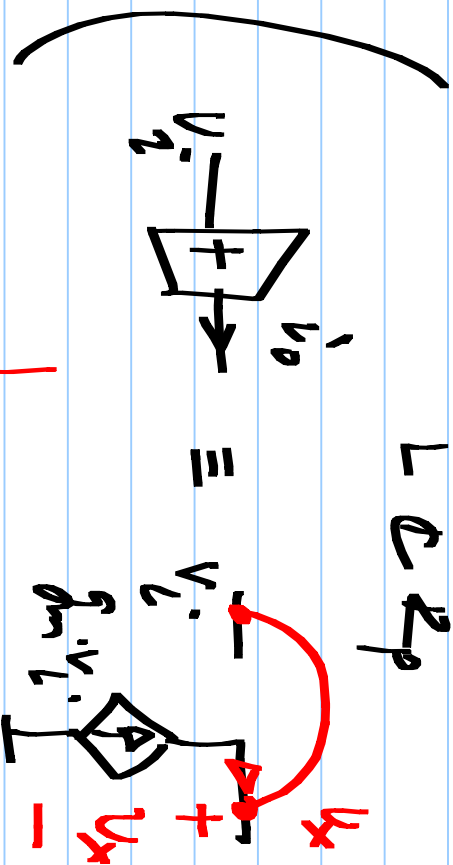
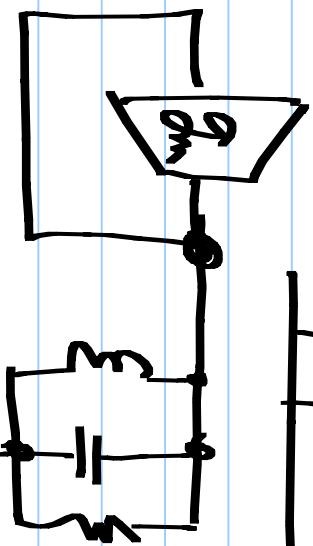
VCO phase noise

16/2/2018



+ Widely tunable
Ring oscillator

period of osc. = $2N \cdot T_d$



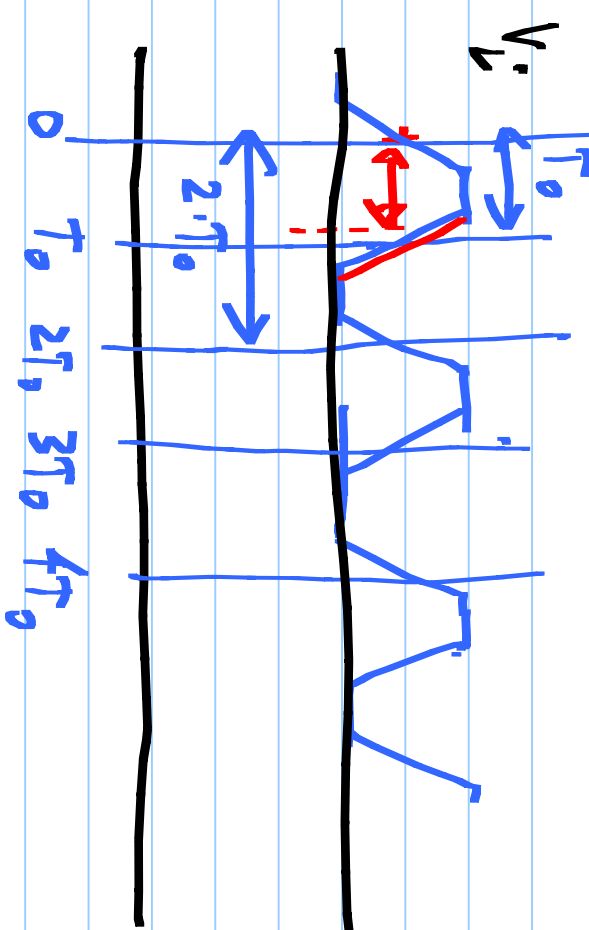
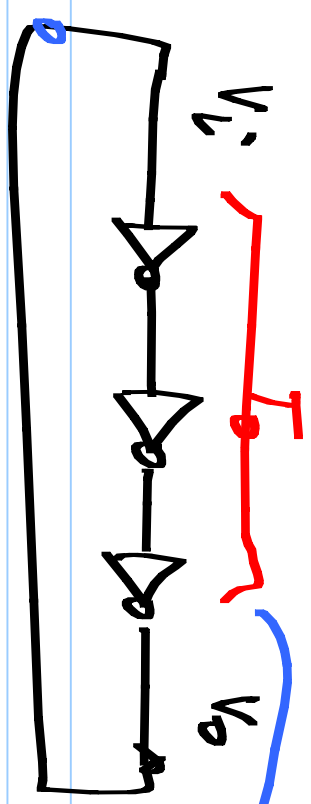
+ lower noise

- occupation a large area
- Narrow tuning range

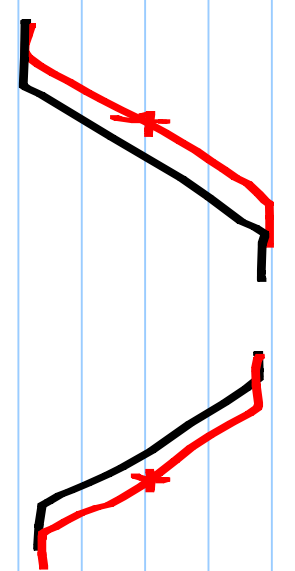
LC oscillator

$f_0 \sim \frac{1}{2\pi \sqrt{LC}}$

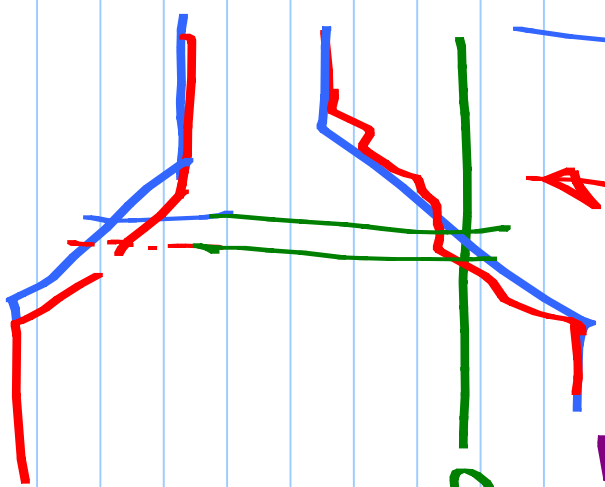
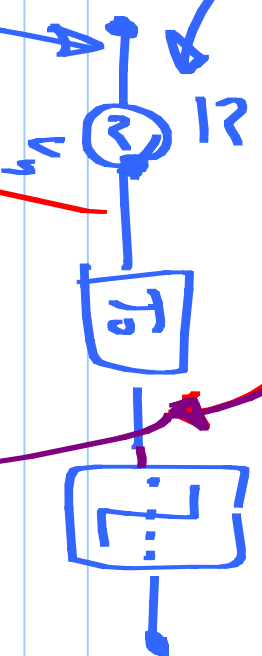
$\sim \frac{1}{g_m}$



$$\frac{v_n(T_0)}{S}$$



ideal periodic with period $2T_0$



Comparator threshold

First transition at $t=0$ (by defn.)

Second transition at $T_0 + \frac{v_n(t_0)}{s} \times \frac{2\pi}{2T_0}$

Third " " $2T_0 + \frac{v_n(t_0)}{s} - \frac{v_n(2T_0)}{s}$

4th " " $3T_0 + \frac{v_n(t_0)}{s} - \frac{v_n(2T_0)}{s} + \frac{v_n(3T_0)}{s}$

\rightarrow White noise; variance $\sigma_{v_n}^2$
 v_n : Input ref. noise of amplifier/delay line/

samples of v_n every $\frac{T_0}{2T_0}$ comparator

$$f_{osc} = \frac{1}{2T_0}$$

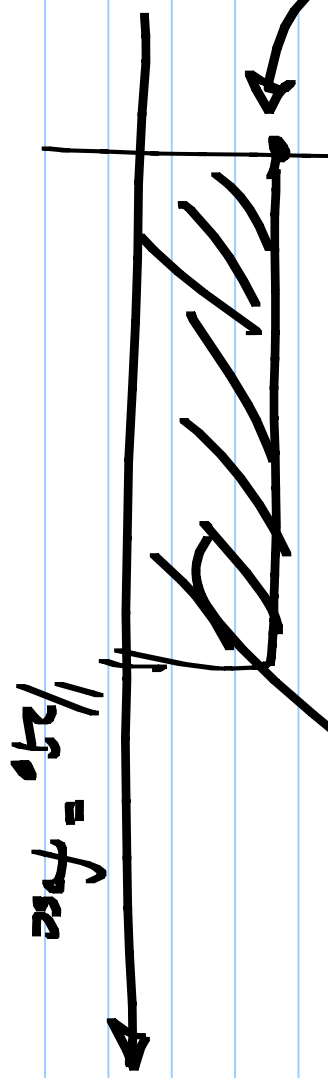
PSD of v_n samples

$$v_n(kT_0) : \sigma_{v_n}^2$$

$$2T_0 \cdot \sigma_{v_n}^2$$

white

$$\sigma_{v_n}^2$$



Cont. time noise

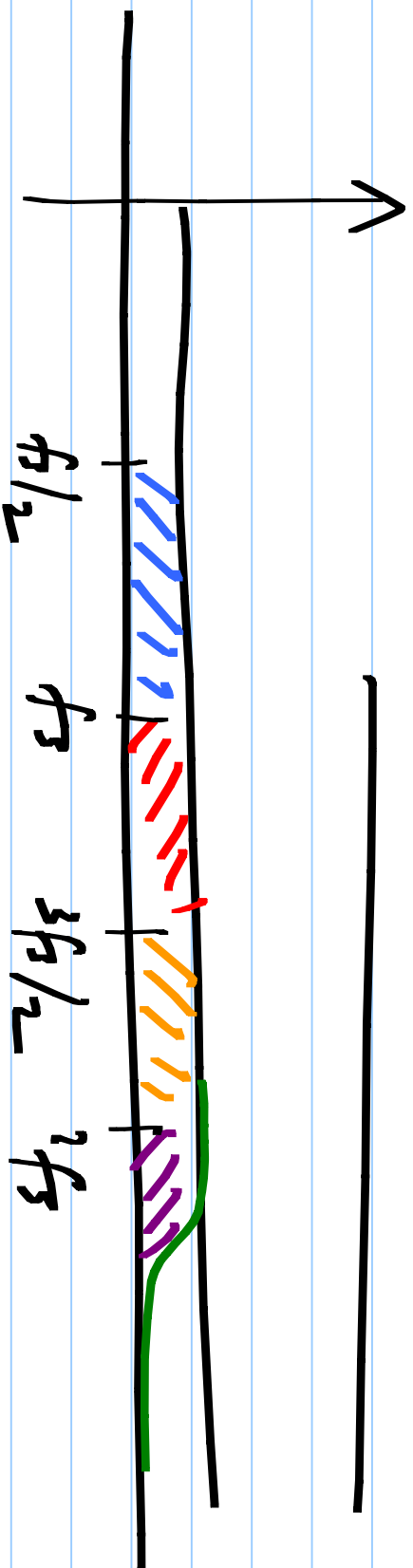
$$\int_{-\infty}^{\infty} S_x(f) df = \sigma_x^2$$

one-sided spectral density

Discrete-time noise

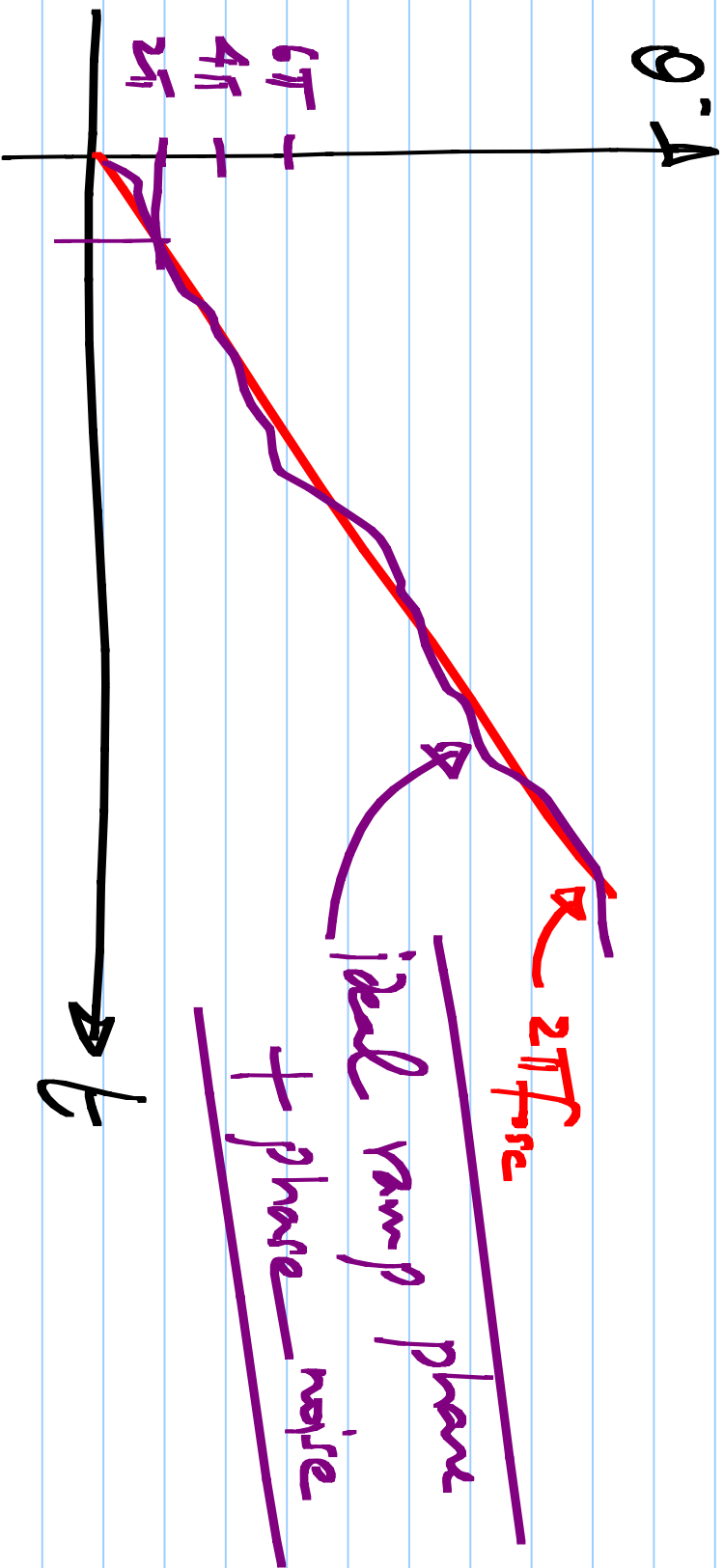
$$\int_{-f_s/2}^{f_s/2} S_x(f) df = \sigma_x^2$$

(Samples @ f_s)



Phase noise: shift in transition instants
(zero-crossing)

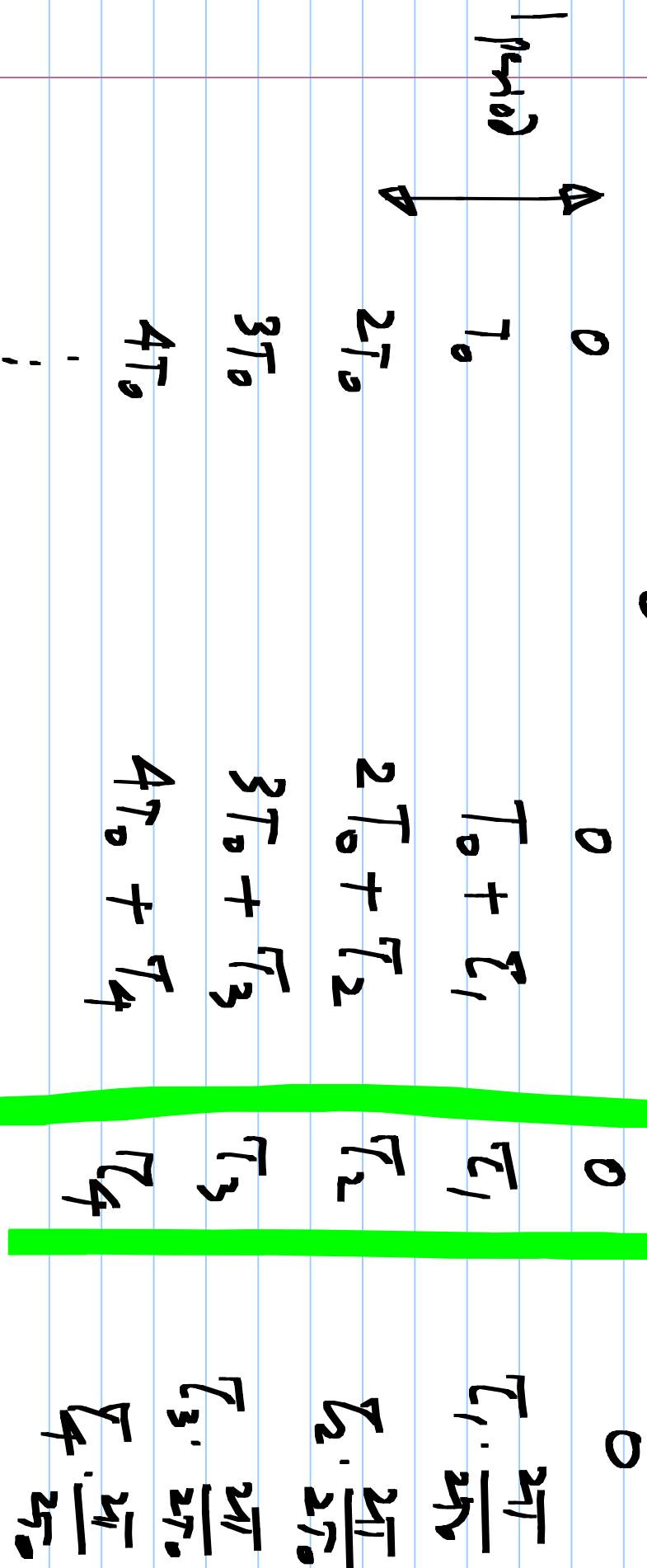
Compared to an ideal periodic signal



1 pole zero-crossings

Actual

Derivative (phase)



$$T_1 = \frac{v_n(\tau_0)}{s}$$

$$T_2 = \frac{v_n(\tau_0)}{s} - \frac{v_n(2\tau_0)}{s}$$

$$T_3 = \frac{v_n(\tau_0)}{s} - \frac{v_n(2\tau_0)}{s} + \frac{v_n(3\tau_0)}{s}$$

⋮

$$T_k - T_{k-1}$$

$$= \frac{v_n(k\tau_0)}{s}$$

$$\frac{v_n(k\tau_0)}{s}$$

Same

FSD on

$$\frac{v_n(k\tau_0)}{s}$$

$$r_k - r_{k-1} = \frac{v_n'(kT_0)}{s} \quad \mathcal{Z}(z)(1-z^{-1}) = \frac{v_n'(z)}{s}$$

width of k^{th} half-period = $(kT_0 + r_k) - ((k-1)T_0 + r_{k-1})$

$$= T_0 + r_k - r_{k-1}$$

k^{th} half-period jitter: uncorrelated / white

$$\mathcal{Z}(z) = \frac{v_n'(z)}{s(1-z^{-1})}$$