

### Problem 2:

The logic here looks for '01' pattern across the output of two consecutive comparators. So if the sampled input is different by 1 V<sub>LSB</sub>, it will cause a bubble error. Error in V<sub>in</sub> = Max. Slope X Skew in clock.

This gives maximum tolerable skew = 5 ps.

In the midterm problem, the logic is more "complicated". It looks for '011' pattern and hence the skew tolerance is relaxed. But then the comparators have to drive more gates.

### Problem 3:

Rds.C = 10% of 1ns = 100ps. C = 2pf => Rds = 50 ohm.

In order to find the Rds (or gds) of the nFET you can either do small signal simulation (as depicted in Fig. 2b) or just do operating point simulation.

W/L = 38um/0.35um

a) 1) Clock feed-through: The falling edge of the clock (VT("/clk")) causes drop in the sampled value. This is evident when signal (VT("/in")) is rising. When the signal is decreasing, rising edge of the clock appears in the clock feed-through - the tiny rising portions in VT("/out").

If the rise/fall time is changed, the clock feed-through changes.

2) The difference between sampled value and the held value is large when the clock switches off, then when the clock switches on (see markers A,B in the plots attached). This is due to charge injection. Under no-charge injection case, the difference between sampled value and held value would remain same, whether the clock is switching on or off.

3) There is finite difference between the input and the sampled signal - Due to finite V<sub>DS</sub> drop and finite time constant.

b) Pedestal error: input signal is 0.5V DC. Check the output signal. Ideally it should be at 0.5V.

Due to charge injection and clock feed-through, when the switch turns off, the output is not 0.5V. Pedestal error is about 40mV.

c) For  $f_{in} = 1\text{GHz}/64 = 15.625\text{ MHz}$ ,  $V_p = 0.1\text{ V}$

DC -> -5.893 dB (which corresponds to 507 mV ... close to 500 mV)

Fundamental @ 15.625 MHz -> -20.1 dB (close to 0.1 V input signal)

2nd Harmonic @ 31.25 MHz -> -80 dB

3rd Harmonic @ 46.875 MHz -> -100.7 dB

$V_p = 0.4\text{ V}$

DC -> -5.89 dB (which corresponds to 507.6 mV ... close to 500 mV)

Fundamental @ 15.625 MHz -> -8.039 dB (corresponds to 396.3 mV - close to 400 mV input signal)

2nd Harmonic @ 31.25 MHz -> -56.34 dB

3rd Harmonic @ 46.875 MHz -> -86.39 dB

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$f_{in} = 65/64 \text{ GHz} = 1.015625 \text{ GHz}$ ,  $V_p = 0.1 \text{ V}$

DC -> -6.004 dB (500.96 mV)

Fundamental @ 15.625 MHz -> -21.43 dB (84.8 mV as compared to 100 mV input)

2nd Harmonic @ 31.25 MHz -> -74.44 dB

3rd Harmonic @ 46.875 MHz -> -87.875 dB

$V_p = 0.4 \text{ V}$

DC -> -6.107 dB (495 mV)

Fundamental @ 15.625 MHz -> -9.318 dB (342 mV as compared to 400 mV input)

2nd Harmonic @ 31.25 MHz -> -54.98 dB

3rd Harmonic @ 46.875 MHz -> -63.16 dB

More distortion in the second case ( $f_{in} = 65/64 \text{ GHz}$ ).

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Problem 4:

Differential circuit - hence "cleaner" output - common mode errors are reduced.

Pedestal error reduced to 100s of fV.

$f_{in} = 1/64 \text{ GHz}$ ,  $V_p = 100 \text{ mV} \Rightarrow$  Differentially 200 mV input amplitude.

DC -> -142.27 dB

Fundamental @ 15.625 MHz -> -13.88 dB (202.3 mV)

2nd Harmonic @ 31.25 MHz -> -128.92 dB

3rd Harmonic @ 46.875 MHz -> -103.16 dB

$V_p = 400 \text{ mV} \Rightarrow$  Differentially 800 mV amplitude

DC -> -131.63 dB

Fundamental @ 15.625 MHz -> -1.842 dB (808.95 mV)

2nd Harmonic @ 31.25 MHz -> -106.86 dB

3rd Harmonic @ 46.875 MHz -> -78.34 dB

$f_{in} = 65/64 \text{ GHz}$ ,  $V_p = 100 \text{ mV} \Rightarrow$  Differentially 200 mV input amplitude.

DC -> -133.57 dB

Fundamental @ 15.625 MHz -> -15.743 dB (163.25 mV)

2nd Harmonic @ 31.25 MHz -> -126.33 dB

3rd Harmonic @ 46.875 MHz -> -80.49 dB

$V_p = 400 \text{ mV} \Rightarrow$  Differentially 800 mV input amplitude.

DC -> -91.25 dB

Fundamental @ 15.625 MHz -> -3.783 dB (646.92 mV)

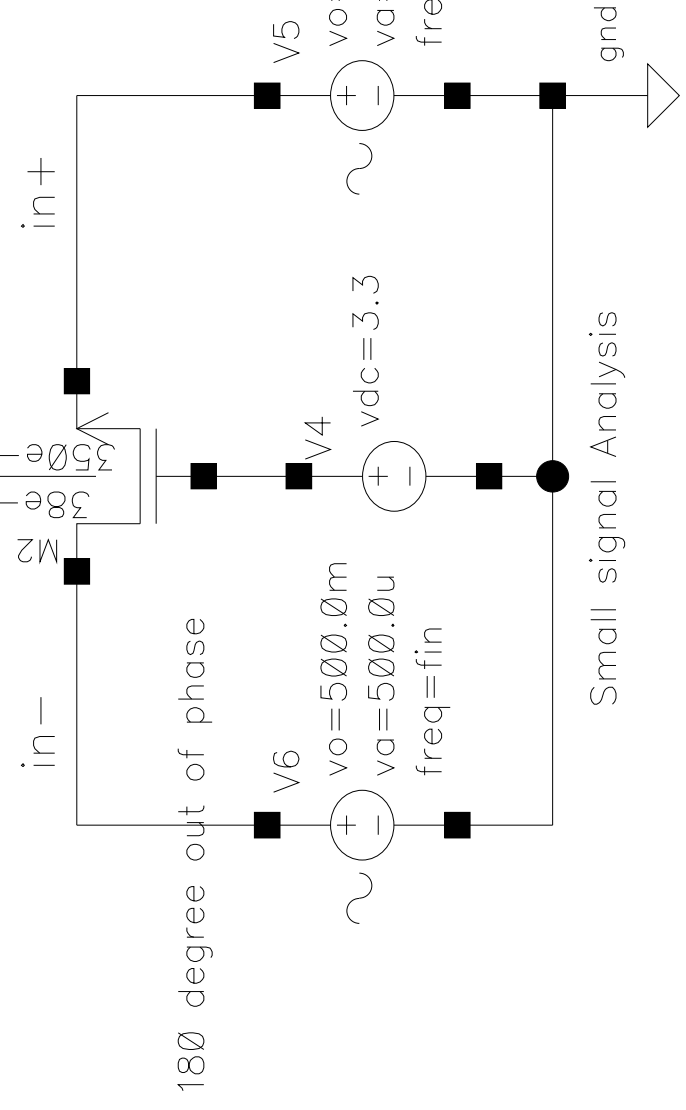
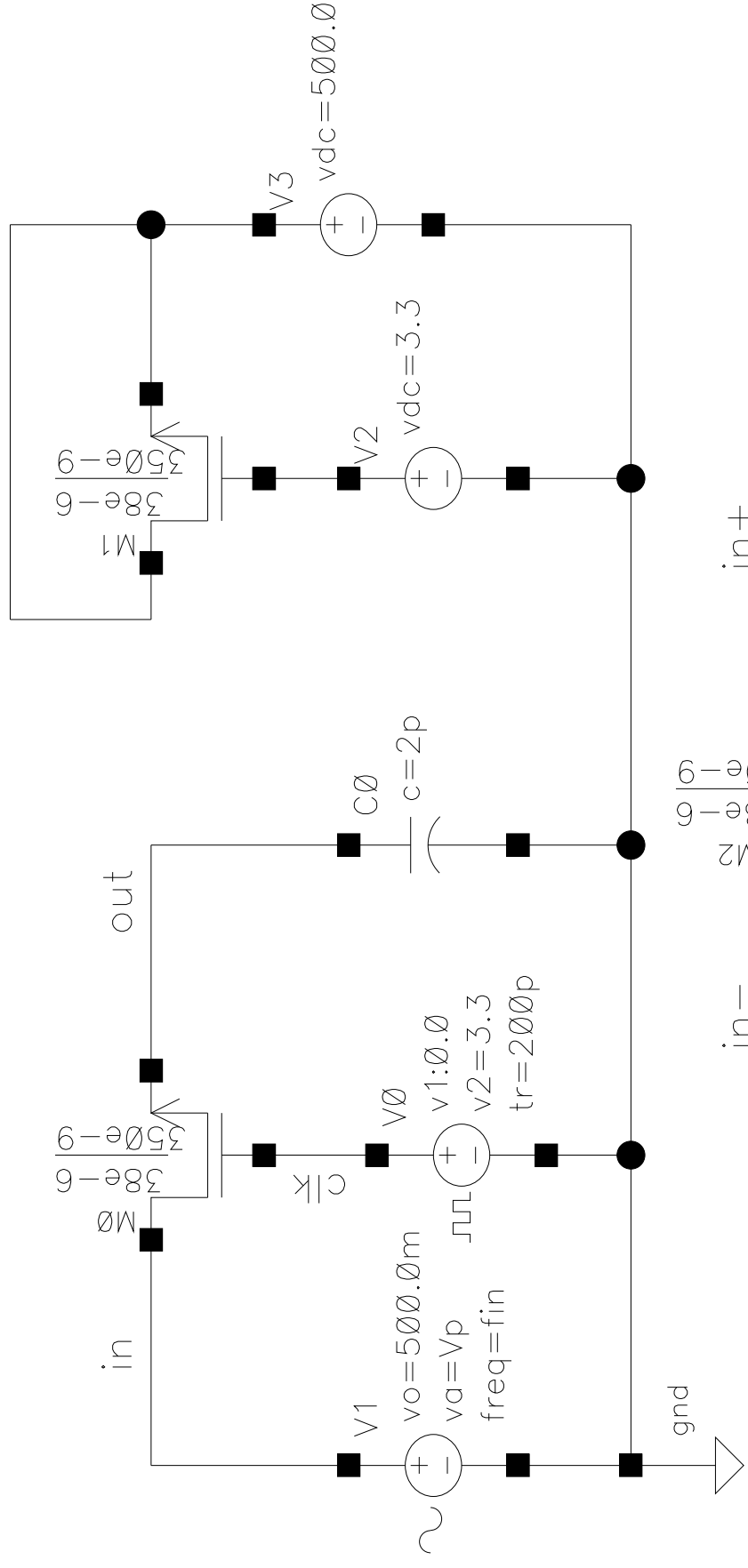
2nd Harmonic @ 31.25 MHz -> -85.63 dB

3rd Harmonic @ 46.875 MHz -> -55.89 dB

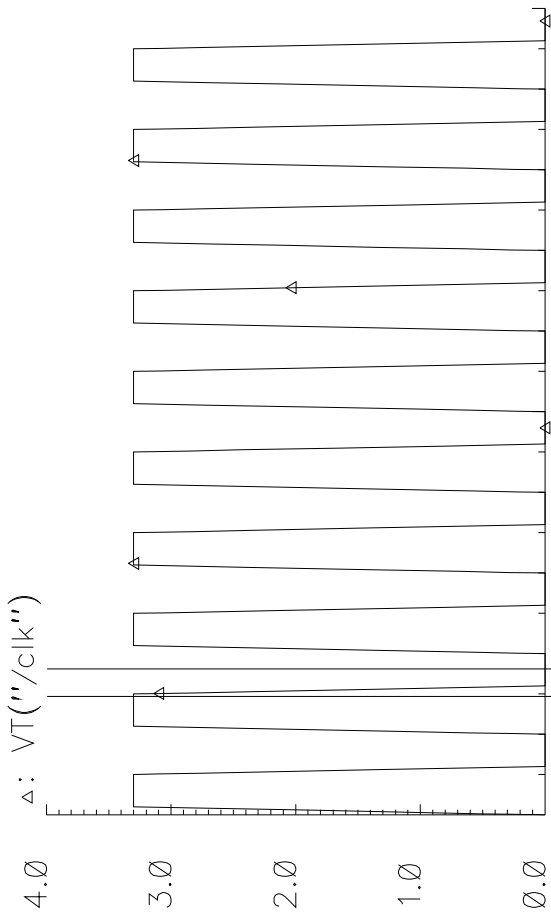
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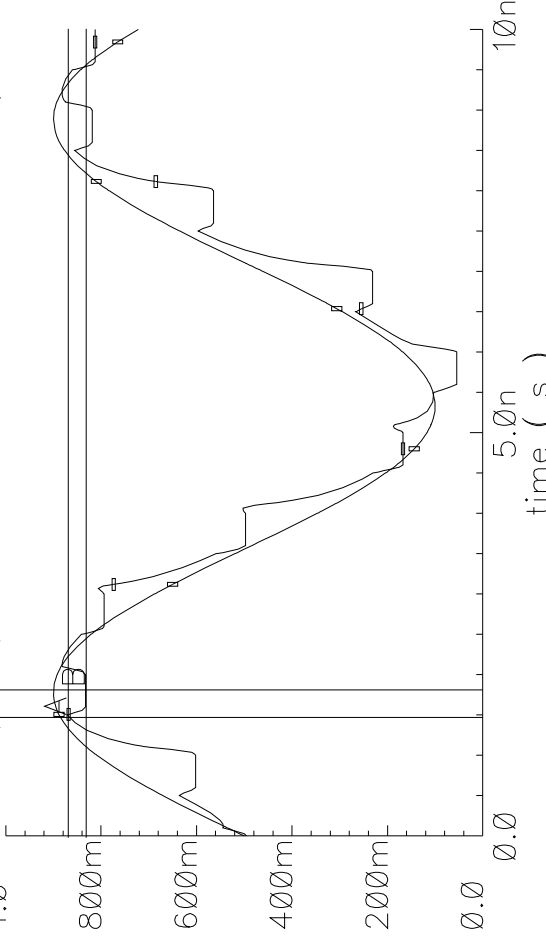
Operating point simulation



Transient Response

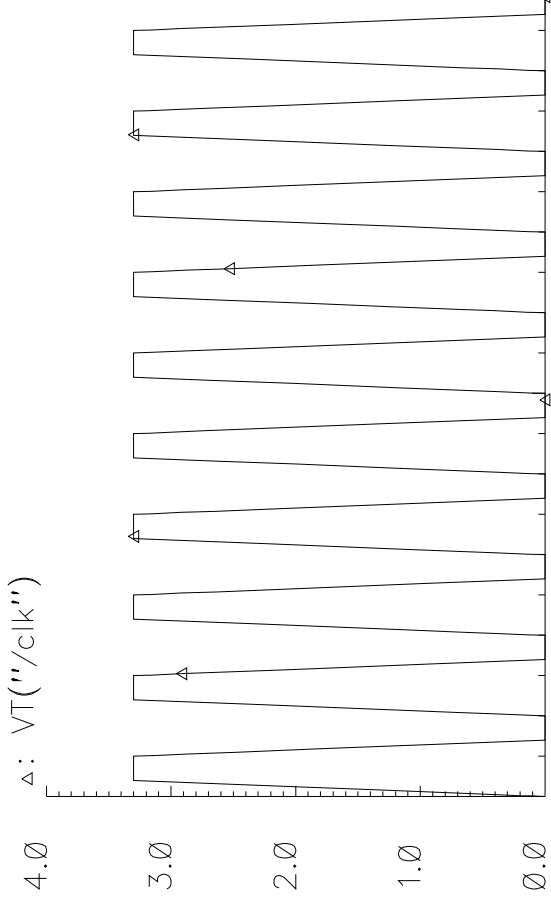


clock rise/fall time = 100ps

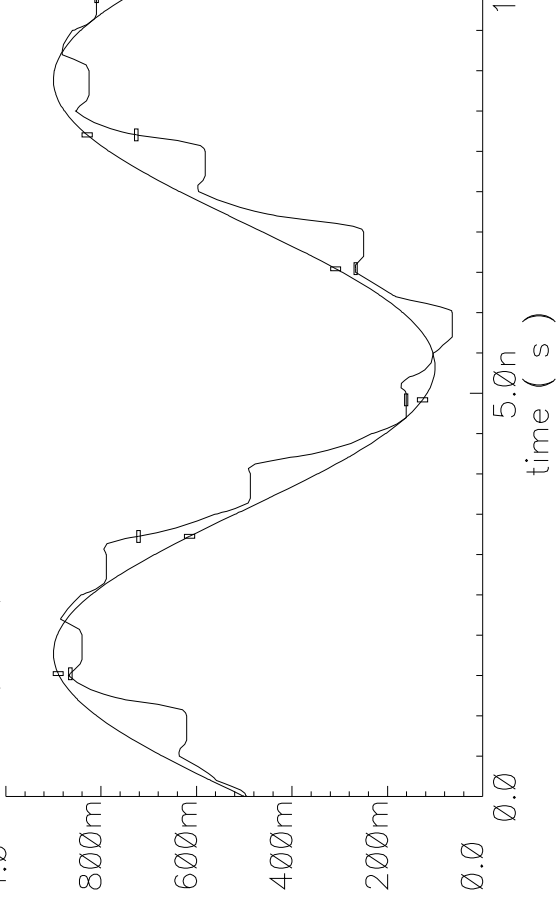


A: (1.49052n 869.339m)  
B: (1.82927n 832.875m)  
delta: (338.75p -36.4644m)  
slope: -107.644M

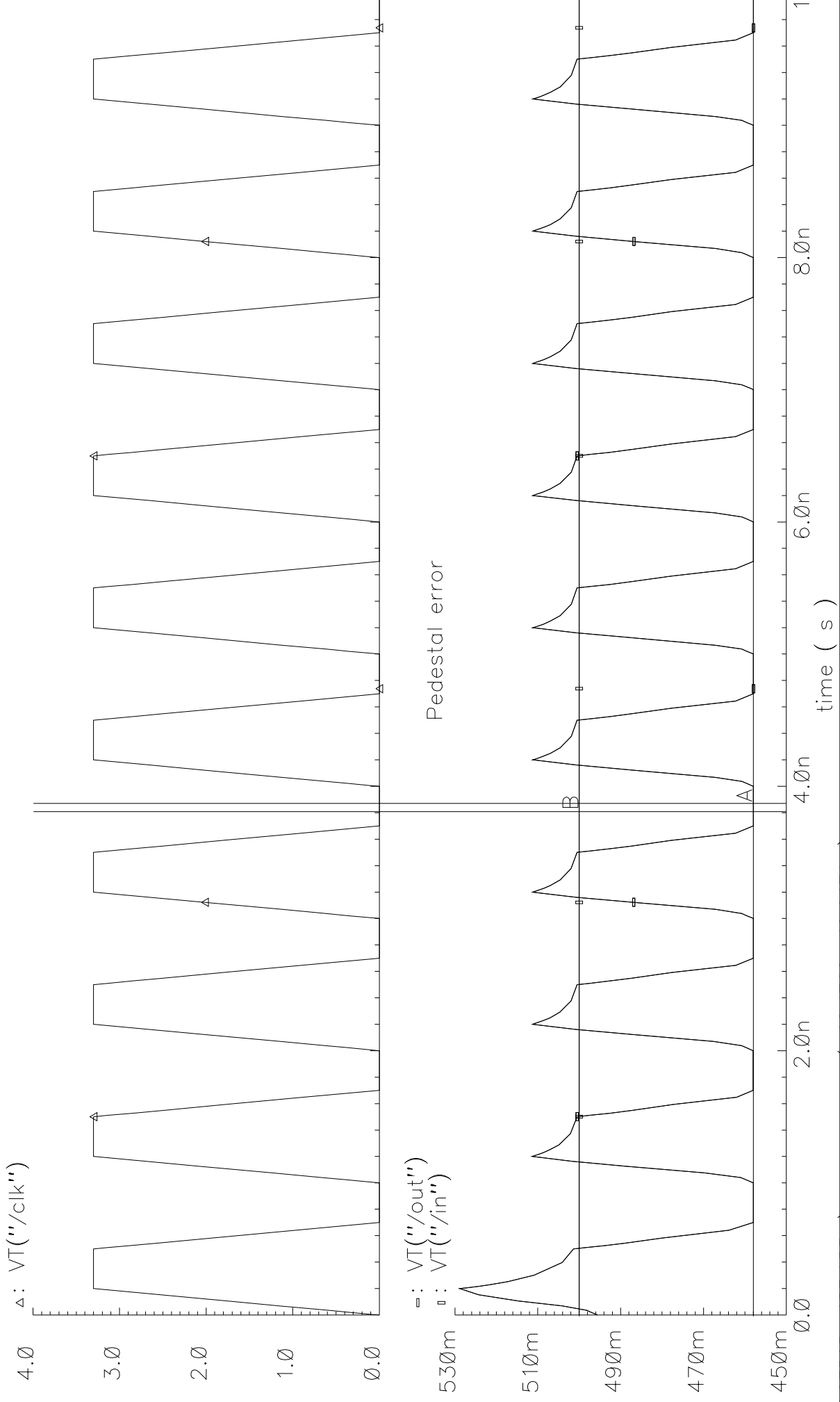
Transient Response



clock rise/fall time = 200ps



### Transient Response



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