## Placement

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## Agenda

- Introduction
- Placement Problem Footprints
- Placement Objective
- Algorithms
$\square$ Min-Cut Placement
$\square$ Analytic Placement
- Quadratic placement
- FDP
$\square$ Simulated Annealing
$\square$ Modern Placement Algorithms
- Summary


## Trends in Placement



- Chips are larger
- Footprints are more diverse
- Empty Space \% is growing
- Interconnect delays are larger percentage of chip cycle time
- Placement is no longer a point tool: It's part of a timing closure system.


Standard Cell sea of gates:

Data Path:

Mixed Data Path \& sea of gates:


## Core

Reserved areas

IP - Floorplanning


Perimeter IO

Area IO


| $\square$ | $\square$ | $\square$ | $\square$ |
| :---: | :---: | :---: | :---: |
| $\square$ | $\square$ | $\square$ | $\square$ |
| $\square$ | $\square$ | $\square$ | $\square$ |
| $\square$ | $\square$ | $\square$ | $\square$ |
| $\square$ | $\square$ | $\square$ | $\square$ |

## Placement Objective

- Find optimal relative ordering of cells
$\square$ minimize wire length and congestion
$\square$ maximize timing slack
- Find optimal spacing of cells
$\square$ eliminate wiring congestion problems
$\square$ provide space for post placement synthesis
- clock trees
- buffer insertion
- timing correction
- Find optimal Global Position


## Optimal Relative Order:

A B C

To spread ...

.. or not to spread


Place to the left

... or to the right

.. or near center


## Optimal Relative Order:

## - A B C

Without "free" space the problem is degenerate: Relative order dominates the solution space.

## Problems limited to Relative Order:

Capo HPNL $=1.3 \mathrm{Te}+06$



## Problem w/spacing / global position components


$\square$

## Optimization Objectives

| Total |
| :---: |
| Wirelength |



Number of Cut Nets



Signal
Delay


## Optimization Objectives - Total Wirelength

Wirelength estimation for a given placement
Half-perimeter
wirelength
(HPWL)

| Complete |
| :--- |
| graph |
| (clique) |

> Monotone chain

$H P W L=9$


Clique Length $=$
$(2 / p) \Sigma_{e \in \operatorname{clique}} d_{M}(e)=14.5$


Chain Length $=12$


Star Length $=15$

## Optimization Objectives - Total Wirelength

Wirelength estimation for a given placement (cont'd.)
Rectilinear
minimum
spanning
tree (RMST)
Rectilinear
Steiner
minimum
tree (RSMT)
Rectilinear
Steiner
arborescence
model (RSA)

Single-trunk Steiner tree (STST)


RMST Length $=11$


RSMT Length $=10$


RSA Length $=10$


STST Length $=10$

## Optimization Objectives - Total Wirelength

## Wirelength estimation for a given placement (cont‘d.)

Preferred method: Half-perimeter wirelength (HPWL)

- Fast (order of magnitude faster than RSMT)
- Equal to length of RSMT for 2- and 3-pin nets
- Margin of error for real circuits approx. 8\% [Chu, ICCAD 04]


RSMT Length $=10$


HPWL $=9$
$L_{\mathrm{HPWL}}=w+h$

## Optimization Objectives - Total Wirelength

Total wirelength with net weights (weighted wirelength)

- For a placement $P$, an estimate of total weighted wirelength is

$$
L(P)=\sum_{n e t \in P} w(\text { net }) \cdot L(\text { net })
$$

where $w(n e t)$ is the weight of net, and $L(n e t)$ is the estimated wirelength of net.

- Example:

Nets

$$
\begin{array}{ll}
N_{1}=\left(a_{1}, b_{1}, d_{2}\right) & w\left(N_{1}\right)=2 \\
N_{2}=\left(c_{1}, d_{1}, f_{1}\right) & w\left(N_{2}\right)=4 \\
N_{3}=\left(e_{1}, f_{2}\right) & w\left(N_{3}\right)=1
\end{array}
$$

$$
L(P)=\sum_{n e t \in P} w(n e t) \cdot L(n e t)=2 \cdot 7+4 \cdot 4+1 \cdot 3=33
$$

## Optimization Objectives - Number of Cut Nets

## Cut sizes of a placement

- To improve total wirelength of a placement $P$, separately calculate the number of crossings of global vertical and horizontal cutlines, and minimize

$$
L(P)=\sum_{v \in V_{P}} \psi_{P}(v)+\sum_{h \in H_{P}} \psi_{P}(h)
$$

where $\Psi_{P}(c u t)$ be the set of nets cut by a cutline cut

## Optimization Objectives - Number of Cut Nets

## Cut sizes of a placement

- Example:

Nets

$$
\begin{aligned}
& N_{1}=\left(a_{1}, b_{1}, d_{2}\right) \\
& N_{2}=\left(c_{1}, d_{1}, f_{1}\right) \\
& N_{3}=\left(e_{1}, f_{2}\right)
\end{aligned}
$$

- Cut values for each global cutline $\psi_{\mathrm{P}}\left(v_{1}\right)=1 \psi_{\mathrm{P}}\left(v_{2}\right)=2$ $\psi_{\mathrm{P}}\left(h_{1}\right)=3 \psi_{\mathrm{P}}\left(h_{2}\right)=2$
- Total number of crossings in $P$

$\Psi_{P}\left(v_{1}\right)+\Psi_{P}\left(v_{2}\right)+\Psi_{P}\left(h_{1}\right)+\Psi_{P}\left(h_{2}\right)=1+2+3+2=8$
- Cut sizes
$X(P)=\max \left(\Psi_{P}\left(v_{1}\right), \Psi_{P}\left(v_{2}\right)\right)=\max (1,2)=2$
$Y(P)=\max \left(\Psi_{P}\left(h_{1}\right), \Psi_{P}\left(h_{2}\right)\right)=\max (3,2)=3$


## Optimization Objectives - Wire Congestion

Routing congestion of a placement

- Ratio of demand for routing tracks to the supply of available routing tracks
- Estimated by the number of nets that pass through the boundaries of individual routing regions



## Optimization Objectives - Wire Congestion

## Routing congestion of a placement

- Formally, the local wire density $\varphi_{P}(e)$ of an edge $e$ between two neighboring grid cells is

$$
\varphi_{P}(e)=\frac{\eta_{P}(e)}{\sigma_{P}(e)}
$$

where $\operatorname{sid}_{P}(e)$ is the estimated number of nets that cross $e$ and $\sigma_{P}(e)$ is the maximum number of nets that can cross $e$

- If $\varphi_{P}(e)>1$, then too many nets are estimated to cross $e$, making $P$ more likely to be unroutable.
- The wire density of $P$ is $\quad \Phi(P)=\max _{e \in E}\left(\varphi_{P}(e)\right)$
where $E$ is the set of all edges
■ If $\Phi(P)$ 1, then the design is estimated to be fully routable, otherwise routing will need to detour some nets through less-congested edges


## Optimization Objectives - Wire Congestion

Wire Density of a placement

$$
\begin{array}{ll}
\eta_{P}\left(h_{1}\right)=1 & \eta_{P}\left(v_{1}\right)=1 \\
\eta_{P}\left(h_{2}\right)=2 & \eta_{P}\left(v_{2}\right)=0 \\
\eta_{P}\left(h_{3}\right)=0 & \eta_{P}\left(v_{3}\right)=0 \\
\eta_{P}\left(h_{4}\right)=1 & \eta_{P}\left(v_{4}\right)=0 \\
\eta_{P}\left(h_{5}\right)=1 & \eta_{P}\left(v_{5}\right)=2 \\
\eta_{P}\left(h_{6}\right)=0 & \eta_{P}\left(v_{6}\right)=0
\end{array}
$$

Maximum:

$$
\eta_{P}(e)=2
$$



$$
\Phi(P)=\frac{\eta_{P}(e)}{\sigma_{P}(e)}=\frac{2}{3}
$$

## Optimization Objectives - Signal Delay

Circuit timing of a placement

- Static timing analysis using actual arrival time (AAT) and required arrival time (RAT)
$\square A A T(v)$ represents the latest transition time at a given node v measured from the beginning of the clock cycle
$\square R A T(v)$ represents the time by which the latest transition at $v$ must complete
in order for the circuit to operate correctly within a given clock cycle.

■ For correct operation of the chip with respect to setup (maximum path delay) constraints, it is required that $A A T(v) \leq R A T(v)$.

## Placement Algorithms

- Min-Cut Placement
- Analytic Placement
- Simulated Annealing

■ Modern Placement Algorithms

## Global Placement

- Partitioning-based algorithms:
$\square$ The netlist and the layout are divided into smaller sub-netlists and subregions, respectively
$\square$ Process is repeated until each sub-netlist and sub-region is small enough to be handled optimally
$\square$ Detailed placement often performed by optimal solvers, facilitating a natural transition from global placement to detailed placement
$\square$ Example: min-cut placement
- Analytic techniques:
$\square$ Model the placement problem using an objective (cost) function, which can be optimized via numerical analysis
$\square$ Examples: quadratic placement and force-directed placement
- Stochastic algorithms:
$\square$ Randomized moves that allow hill-climbing are used to optimize the cost function
$\square$ Example: simulated annealing


## Global Placement



## Min-Cut Placement

- Uses partitioning algorithms to divide the netlist and the layout region into smaller sub-netlists and sub-regions
- Conceptually, each sub-region is assigned a portion of the original netlist
- Each cut heuristically minimizes the number of cut nets using, for example,
$\square$ Kernighan-Lin (KL) algorithm
$\square$ Fiduccia-Mattheyses (FM) algorithm


## Min-Cut Placement

Alternating cutline directions


Repeating cutline directions


## Min-Cut Placement

Input: netlist Netlist, layout area $L A$, minimum number of cells per region cells_min Output: placement $P$

$$
P=\varnothing
$$

regions $=$ ASSIGN(Netlist,LA)
while (regions != Ø)
region = FIRST_ELEMENT(regions)
REMOVE(regions, region)
if (region contains more than cell_min cells)
(sr1,sr2) = BISECT(region)

ADD_TO_END(regions,sr1)
ADD_TO_END(regions,sr2)
else
PLACE(region)
ADD(P,region)
// assign netlist to layout area
// while regions still not placed
// first element in regions
// remove first element of regions
// divide region into two subregions
// sr1 and sr2, obtaining the sub-
// netlists and sub-areas
$/ /$ add $s r 1$ to the end of regions
$/ /$ add $s r 2$ to the end of regions
// place region
// add region to $P$

## Min-Cut Placement - Example

Given:


Task: $4 \times 2$ placement with minimum wirelength using alternative cutline directions and the KL algorithm

$$
\begin{aligned}
& \square \square \square \square \\
& \square \square \square \square \square
\end{aligned}
$$



Vertical cut cut $_{1}: L=\{1,2,3\}, R=\{4,5,6\}$



Horizontal cut cut $_{2 \mathrm{~L}}: T=\{1,4\}, B=\{2,0\}$
Horizontal cut cut $_{2 \mathrm{R}}: T=\{3,5\}, B=\{6,0\}$


## Min-Cut Placement

- Advantages:
$\square$ Reasonably fast
$\square$ Objective function can be adjusted, e.g., to perform timing-driven placement
$\square$ Hierarchical strategy applicable to large circuits
- Disadvantages:
$\square$ Randomized, chaotic algorithms - small changes in input lead to large changes in output (Stability is poor)
$\square$ Optimizing one cutline at a time may result in routing congestion elsewhere


## Analytic Placement - Quadratic Placement

- Objective function is quadratic; sum of (weighted) squared Euclidean distance represents placement objective function

$$
L(P)=\frac{1}{2} \sum_{i, j=j}^{n} c_{i j}\left(\left(x_{i}-x_{j}\right)^{2}+\left(y_{i}-y_{j}\right)^{2}\right)
$$

where $n$ is the total number of cells, and $c(i, j)$ is the connection cost between cells $i$ and $j$.

- Minimize objective function by equating its derivative to zero which reduces to solving a system of linear equations


## Analytic Placement - Quadratic Placement

$$
L(P)=\frac{1}{2} \sum_{i, j=1}^{n} c_{i j}\left(\left(x_{i}-x_{j}\right)^{2}+\left(y_{i}-y_{j}\right)^{2}\right)
$$

where $n$ is the total number of cells, and $c(i, j)$ is the connection cost between cells $i$ and $j$.

- Each dimension can be considered ${ }_{n}$ independently:

$$
L_{x}(P)=\sum_{i=1, j=1}^{n} c(i, j)\left(x_{i}-x_{j}\right)^{2} \quad L_{y}(P)=\sum_{i=1, j=1}^{n} c(i, j)\left(y_{i}-y_{j}\right)^{2}
$$

- Convex quadratic optimization problem: any local minimum solution is also a global minimum
- Optimal $x$ - and $y$-coordinates can be found by setting the partial derivatives of $L_{x}(P)$ and $L_{y}(P)$ to zero


## Analytic Placement - Quadratic Placement

$$
L(P)=\frac{1}{2} \sum_{i, j=1}^{n} c_{i j}\left(\left(x_{i}-x_{j}\right)^{2}+\left(y_{i}-y_{j}\right)^{2}\right)
$$

where $n$ is the total number of cells, and $c(i, j)$ is the connection cost between cells $i$ and $j$.

$$
L_{x}(P)=\sum_{i=1, j=1}^{n} c(i, j)\left(x_{i}-x_{j}\right)^{2} \quad L_{y}(P)=\sum_{i=1, j=1}^{n} c(i, j)\left(y_{i}-y_{j}\right)^{2}
$$

- Each dim sion can be considered indepi lently:

$$
\frac{\partial L_{x}(P)}{\partial X}=A X-b_{x}=0 \quad \frac{\partial L_{y}(P)}{\partial Y}=A Y-b_{y}=0
$$

where $A$ is a matrix with $A[7][]=c(i, j)$ when $i \neq j$, and $A[][]=$ the sum of incident connection weights of cell $i$.
$X$ is a vector of all the $x$-coordinates of the non-fixed cells, and $b_{x}$ is a vector with $b_{x}[I]=$ the sum of $x$-coordinates of all fixed cells attached to $i$.
$Y$ is a vector of all the $y$-coordinates of the non-fixed cells, and $b_{y}$ is a vector with $b_{y}[1]=$ the sum of $y$-coordinates of all fixed cells attached to $i$.

## Analytic Placement - Quadratic Placement

$$
L(P)=\frac{1}{2} \sum_{i, j=1}^{n} c_{i j}\left(\left(x_{i}-x_{j}\right)^{2}+\left(y_{i}-y_{j}\right)^{2}\right)
$$

where $n$ is the total number of cells, and $c(i, j)$ is the connection cost between cells $i$ and $j$.
$L_{x}(P)=\sum^{n} c(i, j)\left(x_{i}-x_{j}\right)^{2} \quad L_{y}(P)=\sum^{n} c(i, j)\left(y_{i}-y_{j}\right)^{2}$

- Each dimen, sion can be considered independdently:

$$
\frac{\partial L_{x}(P)}{\partial X}=A X-b_{x}=0
$$

$$
\frac{\partial L_{y}(P)}{\partial Y}=A Y-b_{y}=0
$$

■ System of linear equations for which iterative numerical methods can be used to find a solution

Why formulate the problem this way?

- Known techniques make solution easy to find
- There is only one solution
- The solution is a global optimum
- The solution conveys "relative order" information
- The solution conveys "global position" information


## Analytic Placement - Quadratic Placement

- Mechanical analogy: mass-spring system

$\square$ Squared Euclidean distance is proportional to the energy of a spring between these points
$\square$ Quadratic objective function represents total energy of the spring system; for each movable object, the $x(y)$ partial derivative represents the total force acting on that object
$\square$ Setting the forces of the nets to zero, an equilibrium state is mathematically modeled that is characterized by zero forces acting on each movable object
$\square$ At the end, all springs are in a force equilibrium with a minimal total spring energy; this equilibrium represents the minimal sum of squared wirelength
$\rightarrow$ Result: many cell overlaps


## Analytic Placement - Quadratic Placement

- Second stage of quadratic placers: cells are spread out to remove overlaps
- Methods:
$\square$ Adding fake nets that pull cells away from dense regions toward anchors
$\square$ Geometric sorting and scaling
$\square$ Repulsion forces, etc.



What does the solution look like?

- To get an intuitive feel for the solution, examine the relaxation method for solving $\mathrm{Ax}+\mathrm{B}=0$
- Actual program implementation may use other solution methods (that are generally less intuitive).

Solution of $Q$ uadratic Using nelaxation:


## Analytic Placement - Quadratic Placement

- Advantages:
$\square$ Captures the placement problem concisely in mathematical terms
$\square$ Leverages efficient algorithms from numerical analysis and available software
$\square$ Can be applied to large circuits without netlist clustering (flat)
$\square$ Stability: small changes in the input do not lead to large changes in the output
- Disadvantages:
$\square$ Connections to fixed objects are necessary: I/O pads, pins of fixed macros, etc.


## Analytical Constraint Generation: A Hybrid Approach

- Combine Quadratic techniques with MLP
- Use Quadratic solution to determine global position (ie balance)
- Use MLP to determine relative ordering of cells
[9] C. J. Alpert, G.-J. Nam, and P. G. Villarrubia, "Free Space Management for Cut-Based Placement"
Proc. IEEE Intl. Conf. on Computer-Aided Design\}, November, 2002.


## Analytical Constraint Generation




Select a command
Level PBLK: Visibility altered
Visibility has been altered for more than 1 level, only one modification is listed
Command>




Select a command
Level PBLK: Visibility altered
Visibility has been altered for more than 1 level, only one modification is listed
Command>



Select a command
Level PBLK: Visibility altered
Visibility has been altered for more than 1 level, only one modification is listed
Gommand:


Visibility has been altered for more than 1 level, only one modification is listed Select a command
Level FX: Visibility altered
Command)


Command $>$


Select a command
Level PBLK: Visibility altered
Visibility has been altered for more than 1 level, only one modification is listed
Command>


## Select a command

Level PBLK: Visibility altered
Visibility has been altered for more than 1 level, only one modification is listed

## Geometric Partitioning

J. Vygen, "Algorithms for Large-Scale Flat Placement",

Proc. 34th IEEE/ACM Design Automation Conference, 1988,pp 746-751

GymBrowser (3.08.03) Physical View(INT):GUL Mode:Browse(INT.int.all.Tier<*>) Loaded:/tmp/g| file_1_1




Repaint
EndCommand

## $\square$ Grid <br> $\downarrow$ Tic

## Color has been altered for more than 1 level, only one modification is listed

Fill has been altered for more than 1 level, only one modification is listed
Level CKTROW: Visibility altered

## Command

GymBrowser (3.08.03) Physical View(INT):GUL Mode:Browse(INT.int.all.Tier<*>) Loaded//tmp/g|1 file_3_1


Repaint



Repaint

## Grid

$\square$ Tic



Repaint EndCommand

## Grid

$\checkmark$ Tic



|  |  |
| :---: | :---: |
|  |  |
|  |  |
|  |  |
| Mag | Pan |
| ZmIn | ZmOut |
| Prev | Max |
| Fit | SS |
| $\mathrm{x}=$ | -614.7 |
| $\mathbf{y}=$ | 5950.8 |
| $\mathbf{r x}=$ | 0.0 |
| $\mathbf{r y}=$ | 25.0 |
| dx= | -614.7 |
| dy= | 5925.8 |

## Repaint



## Analytic Placement - Force-directed Placement

- Cells and wires are modeled using the mechanical analogy of a mass-spring system, i.e., masses connected to Hooke's-Law springs

- Attraction force between cells is directly proportional to their distance
- Cells will eventually settle in a force equilibrium
$\square$ minimized wirelength


## Analytic Placement - Force-directed Placement

- Given two connected cells $a$ and $b$, the attraction force $\overrightarrow{F_{a b}}$ exerted on $a$ by $b$ is

$$
\overrightarrow{F_{a b}}=c(a, b) \cdot(\vec{b}-\vec{a})
$$

where
$\square c(a, b)$ is the connection weight (priority) between cells a and $b$, and
$\square(\vec{b}-\vec{a})$ is the vector difference of the positions of $a$ and $b$ in the Euclidean plane

- The sum of forces exerted on a cell $i$ connected to other cells $1 \ldots j$ is $\quad \vec{F}_{i}=\sum_{c(i, j) \neq 0} \vec{F}_{i j}$
- Zero-force target (ZFT): position that minimizes this sum of forces


## Analytic Placement - Force-directed Placement

Zero-Force-Target (ZFT) position of cell $i$

$\min \overrightarrow{F_{i}}=c(i, a) \cdot(\vec{a}-\vec{i})+c(i, b) \cdot(\vec{b}-\vec{i})+c(i, c) \cdot(\vec{c}-\vec{i})+c(i, d) \cdot(\vec{d}-\vec{i})$

## Analytic Placement - Force-directed Placement

## Basic force-directed placement

- Iteratively moves all cells to their respective ZFT positions
- $x$ - and $y$-direction forces are set to zero:

$$
\sum_{c(i, j) \neq 0} c(i, j) \cdot\left(x_{j}^{0}-x_{i}^{0}\right)=0 \quad \sum_{c(i, j) \neq 0} c(i, j) \cdot\left(y_{j}^{0}-y_{i}^{0}\right)=0
$$

- Rearranging the variables to solve for $x_{i}^{0}$ and $y_{i}^{0}$ yields

$$
x_{i}^{0}=\frac{\sum_{c(i, j) \neq 0} c(i, j) \cdot x_{j}^{0}}{\sum_{c(i, j) \neq 0} c(i, j)} \quad y_{i}^{0}=\frac{\sum_{c(i, j) \neq 0} c(i, j) \cdot y_{j}^{0}}{\sum_{c(i, j) \neq 0} c(i, j)} \quad \begin{aligned}
& \text { Computation of } \\
& \text { ZFT position of cell } i \\
& \text { connected with } \\
& \text { cells } 1 \ldots j
\end{aligned}
$$

## Analytic Placement - Force-directed Placement

## Example: ZFT position

## Given:

$\square$ Circuit with NAND gate 1 and four I/O pads on a $3 \times 3$ grid
$\square$ Pad positions: In1 (2,2), In2 (0,2), In3 (0,0), Out $(2,0)$
$\square$ Weighted connections: $c(a, \ln 1)=8, \quad c(a, \ln 2)=10, \quad c(a, \ln 3)$

$$
=2, \quad c(a, O u t)=2
$$

Task: find the ZFT position of cell a


## Analytic Placement - Force-directed Placement

## Example: ZFT position

Given:
$\square$ Circuit with NAND gate 1 and four I/O pads on a $3 \times 3$ grid
$\square$ Pad positions: In1 (2,2), In2 (0,2), In3 (0,0), Out (2,0)

Solution:

$$
\begin{aligned}
& x_{a}^{0}=\frac{\sum_{c(i, j) \neq 0} c(a, j) \cdot x_{j}^{0}}{\sum_{c(i, j) \neq 0} c(a, j)}=\frac{c(a, \operatorname{In} 1) \cdot x_{\text {In } 1}+c(a, \operatorname{In} 2) \cdot x_{\text {In } 2}+c(a, \operatorname{In} 3) \cdot x_{\text {In } 3}+c(a, \text { Out }) \cdot x_{\text {Out }}}{c(a, \operatorname{In} 1)+c(a, \operatorname{In} 2)+c(a, \operatorname{In} 3)+c(a, \text { Out })}=\frac{8 \cdot 2+10 \cdot 0+2 \cdot 0+2 \cdot 2}{8+10+2+2}=\frac{20}{22} \approx 0.9 \\
& y_{a}^{0}=\frac{\sum_{c(i, j) \neq 0} c(a, j) \cdot y_{j}^{0}}{\sum_{c(i, j) \neq 0} c(a, j)}=\frac{c(a, \operatorname{In} 1) \cdot y_{\text {In } 1}+c(a, \operatorname{In} 2) \cdot y_{\operatorname{In} 2}+c(a, \operatorname{In} 3) \cdot y_{\text {In } 3}+c(a, \text { Out }) \cdot y_{\text {Out }}}{c(a, \operatorname{In} 1)+c(a, \operatorname{In} 2)+c(a, \operatorname{In} 3)+c(a, \text { Out })}=\frac{8 \cdot 2+10 \cdot 2+2 \cdot 0+2 \cdot 0}{8+10+2+2}=\frac{36}{22} \approx 1.6
\end{aligned}
$$

ZFT position of cell $a$ is $(1,2)$

## Analytic Placement - Force-directed Placement

Example: ZFT position
Given:
$\square$ Circuit with NAND gate 1 and four I/O pads on a $3 \times 3$ grid
$\square$ Pad positions: In1 (2,2), In2 (0,2), In3 (0,0), Out $(2,0)$

Solution:

ZFT position of cell $a$ is $(1,2)$


## Analytic Placement - Force-directed Placement

Input: set of all cells $V$
Output: placement $P$

```
P = PLACE(V)
loc = LOCATIONS(P)
foreach (cell c\inV)
    status[c] = UNMOVED
while (ALL_MOVED(V) || !STOP())
    c = MAX_DEGREE(V,status)
    ZFT_pos = ZFT_POSITION(c)
    if (loc[ZFT_pos] == Ø)
        loc[ZFT_pos] = c
    else
        RELOCATE(c,loc)
    status[c] = MOVED
```

// arbitrary initial placement
// set coordinates for each cell in $P$
// continue until all cells have been
// moved or some stopping
// criterion is reached
// unmoved cell that has largest
// number of connections
// ZFT position of $c$
// if position is unoccupied,
// move $c$ to its ZFT position
// use methods discussed next
// mark c as moved

## Analytic Placement - Force-directed Placement

- Finding a valid location for a cell with an occupied ZFT position ( $p$ : incoming cell, $q$ : cell in $p$ s ZFT position)
- If possible, move $p$ to a cell position close to $q$.
- Chain move: cell $p$ is moved to cells $q$ 's location.
$\square$ Cell $q$, in turn, is shifted to the next position. If a cell $r$ is occupying this space, cell $r$ is shifted to the next position.
$\square$ This continues until all affected cells are placed.
- Compute the cost difference if $p$ and $q$ were to be swapped. If the total cost reduces, i.e., the weighted connection length $L(P)$ is smaller, then swap $p$ and $q$.


## Analytic Placement - Force-directed Placement (Example)

Given:

```
Nets
N}=(\mp@subsup{b}{1}{},\mp@subsup{b}{3}{}
N}=(\mp@subsup{b}{2}{},\mp@subsup{b}{3}{}
\(N_{1}=\left(b_{1}, b_{3}\right)\)
\(N_{2}=\left(b_{2}, b_{3}\right)\)
```

Weight
$\mathrm{c}\left(N_{1}\right)=2$
$c\left(N_{2}\right)=1$


## Analytic Placement - Force-directed Placement (Example)

Given:

```
Nets
N}=(\mp@subsup{b}{1}{},\mp@subsup{b}{3}{}
N}=(\mp@subsup{b}{2}{},\mp@subsup{b}{3}{}
Weight
\(\mathrm{c}\left(N_{1}\right)=2\)
\(c\left(N_{2}\right)=1\)
```




## Analytic Placement - Force-directed Placement (Example)

Given:

Nets
$N_{1}=\left(b_{1}, b_{3}\right)$
$N_{2}=\left(b_{2}, b_{3}\right)$
$\mathrm{c}\left(N_{1}\right)=2$
$c\left(N_{2}\right)=1$


Cell $q \quad \begin{aligned} & L(P) \\ & \text { before } \\ & \text { move }\end{aligned} \quad \begin{aligned} & L(P) / \text { placement } \\ & \text { after move }\end{aligned}$
$b_{1} \quad L(P)=5 \quad L(P)=5$

$\rightarrow$ No swapping of $b_{3}$ and $b_{1}$

$$
\begin{aligned}
& b_{2} \quad x_{b_{2}}^{0}=\frac{\sum_{c\left(b_{2}, j\right) \neq 0} c\left(b_{2}, j\right) \cdot x_{j}^{0}}{\sum_{c\left(b_{2}, j\right) \neq 0} c\left(b_{2}, j\right)}=\frac{1 \cdot 2}{1}=2 \quad b_{3} \quad L(P)=5 \quad L(P)=3 \quad b_{1} \ldots b_{3} \quad b_{2} \\
& \rightarrow \text { Swapping of } b_{2} \text { and } b_{3}
\end{aligned}
$$

- 6 Movable Objects
- 2 PIOs
- Connections between objects (Nets)



## FDP Flow

(2)

Step 1: Solve convex quadratic program


## FDP Flow

Step 2: Spread objects to reduce overlap


## FDP Flow

(4)

Step 3: Add spreading forces to objects for next quadratic program


## Addition of Spreading Forces



## Importance of Spreading

Solve initial convex quadratic program (QP)
While target density is not met
Spread objects to reduce overlap
Add spreading forces to QP formulation
Solve the convex quadratic program
end while

■ Need to carefully control the magnitude of the spreading forces
$\square$ Fast spreading will severely degrade wirelength
$\square$ Slow spreading affects turn-around-time

## Force Directed Placement



## Analytic Placement - Force-directed Placement

- Advantages:
$\square$ Conceptually simple, easy to implement
$\square$ Primarily intended for global placement, but can also be adapted to detailed placement
- Disadvantages:
$\square$ Does not scale to large placement instances
$\square$ Is not very effective in spreading cells in densest regions
$\square$ Poor trade-off between solution quality and runtime
- In practice, FDP is extended by specialized techniques for cell spreading
$\square$ This facilitates scalability and makes FDP competitive


## Force Directed Placement Example



## Force-vector modulation

Proposes to modulate the spreading force vectors within analytical placement


## FDP Flow with Modulation

Solve initial convex quadratic program (QP)
While target density is not met
Spread objects to reduce overlap
Order objects based on spreading force magnitude Modulate spreading forces for top x\% of objects
Add spreading forces to QP formulation
Solve the convex quadratic program
end while

## Spreading forces



## Nullify Top x\% of Spreading Forces



Rank Modules based on the spreading force magnitude

Nullify the spreading force magnitude for top $\mathrm{x} \%$ of modules

> Typically $x=5-10 \%$ within RQL

## Advantages of modulation

## Improves Wirelength

## Reorders modules at a global scale

No Impact on spreading

Can be incorporated within any analytical placer

## Simulated Annealing



- Analogous to the physical annealing process
$\square$ Melt metal and then slowly cool it
$\square$ Result: energy-minimal crystal structure
- Modification of an initial configuration (placement) by moving/exchanging of randomly selected cells
$\square$ Accept the new placement if it improves the objective function
$\square$ If no improvement: Move/exchange is accepted with temperaturedependent (i.e., decreasing) probability


## Simulated Annealing - Algorithm

Input: set of all cells $V$
Output: placement $P$

```
T=T
P = PLACE(V)
while (T> T Tmin}
// set initial temperature
// arbitrary initial placement
```

```
while (!STOP())
```

while (!STOP())
new_P = PERTURB(P)
new_P = PERTURB(P)
\Deltacost = COST(new_P) - COST(P)
\Deltacost = COST(new_P) - COST(P)
if ( }\Delta\mathrm{ cost < 0)
if ( }\Delta\mathrm{ cost < 0)
P= new_P
P= new_P
else
else
r= RANDOM(0,1)
r= RANDOM(0,1)
if (r<e-\DeltacostT)
if (r<e-\DeltacostT)
P = new_P
P = new_P
T=\alpha\cdotT
T=\alpha\cdotT

```
// not yet in equilibrium at T
```

// not yet in equilibrium at T
// cost improvement
// cost improvement
// accept new placement
// accept new placement
// no cost improvement
// no cost improvement
// random number [0,1)
// random number [0,1)
// probabilistically accept
// probabilistically accept
// reduce T, 0<\alpha<1

```
    // reduce T, 0<\alpha<1
```


## Simulated Annealing

- Advantages:
$\square$ Can find global optimum (given sufficient time)
$\square$ Well-suited for detailed placement
- Disadvantages:
$\square$ Very slow
$\square$ To achieve high-quality implementation, laborious parameter tuning is necessary
$\square$ Randomized, chaotic algorithms - small changes in the input lead to large changes in the output
- Practical applications of SA:
$\square$ Very small placement instances with complicated constraints
$\square$ Detailed placement, where SA can be applied in small windows (not common anymore)
$\square$ FPGA layout, where complicated constraints are becoming a norm


## Modern Placement Algorithms

- Predominantly analytic algorithms
- Solve two challenges: interconnect minimization and cell overlap removal (spreading)
- Two families:


Non-convex optimization placers

## Modern Placement Algorithms



Quadratic placers

Non-convex
optimization placers

- Solve large, sparse systems of linear equations (formulated using force-directed placement) by the Conjugate Gradient algorithm
- Perform cell spreading by adding fake nets that pull cells away from dense regions toward carefully placed anchors


## Modern Placement Algorithms

## Quadratic placers

- Model interconnect by sophisticated differentiable functions, e.g., log-sum-exp is the popular choice
- Model cell overlap and fixed obstacles by additional (non-convex) functional terms
- Optimize interconnect by the non-linear Conjugate Gradient algorithm
- Sophisticated, slow algorithms
- All leading placers in this category use netlist clustering to improve computational scalability (this further complicates the implementation)


## Modern Placement Algorithms



Non-convex<br>optimization placers

Pros and cons:

- Quadratic placers are simpler and faster, easier to parallelize
- Non-convex optimizers tend to produce better solutions
- As of 2011, quadratic placers are catching up in solution quality while running 5-6 times faster


## Legalization and Detailed Placement

- Global placement must be legalized
$\square$ Cell locations typically do not align with power rails
$\square$ Small cell overlaps due to incremental changes, such as cell resizing or buffer insertion
- Legalization seeks to find legal, non-overlapping placements for all placeable modules
- Legalization can be improved by detailed placement techniques, such as
$\square$ Swapping neighboring cells to reduce wirelength
$\square$ Sliding cells to unused space
- Software implementations of legalization and detailed placement are often bundled


## Legalization and Detailed Placement

Legal positions of standard cells between VDD and GND rails


## Summary

- Row-based standard-cell placement
$\square$ Cell heights are typically fixed, to fit in rows (but some cells may have double and quadruple heights)
$\square$ Legal cell sites facilitate the alignment of routing tracks, connection to power and ground rails
- Wirelength as a key metric of interconnect
$\square$ Bounding box half-perimeter (HPWL)
$\square$ Cliques and stars
$\square$ RMSTs and RSMTs
- Objectives: wirelength, routing congestion, circuit delay
$\square$ Algorithm development is usually driven by wirelength
$\square$ The basic framework is implemented, evaluated and made competitive on standard benchmarks
$\square$ Additional objectives are added to an operational framework


## Summary

- Combinatorial optimization techniques: min-cut and simulated annealing
$\square$ Can perform both global and detailed placement
$\square$ Reasonably good at small to medium scales
$\square$ SA is very slow, but can handle a greater variety of constraints
$\square$ Randomized and chaotic algorithms - small changes at the input can lead to large changes at the output
- Analytic techniques: force-directed placement and non-convex optimization
$\square$ Primarily used for global placement
$\square$ Unrivaled for large netlists in speed and solution quality
$\square$ Capture the placement problem by mathematical optimization
$\square$ Use efficient numerical analysis algorithms
$\square$ Ensure stability: small changes at the input can cause only small changes at the output
$\square$ Example: a modern, competitive analytic global placer takes 20mins for global placement of a netlist with 2.1 M cells (single thread, 3.2 GHz Intel CPU)


## Legalization and Detailed Placement

- Legalization ensures that design rules \& constraints are satisfied
$\square$ All cells are in rows
$\square$ Cells align with routing tracks
$\square$ Cells connect to power \& ground rails
$\square$ Additional constraints are often considered, e.g., maximum cell density
- Detailed placement reduces interconnect, while preserving legality
$\square$ Swapping neighboring cells, rotating groups of three
$\square$ Optimal branch-and-bound on small groups of cells
$\square$ Sliding cells along their rows
$\square$ Other local changes
- Extensions to optimize routed wirelength, routing congestion and circuit timing
- Relatively straightforward algorithms, but high-quality, fast implementation is important
- Most relevant after analytic global placement, but are also used after min-cut placement
- Rule of thumb: $50 \%$ runtime is spent in global placement, $50 \%$ in detailed placement


## State-of-the-art Analytical Placers

|  | APlace | mPL6 | NTUP3 |
| :---: | :---: | :---: | :---: | FDP

The mountain hike analogy


