

## Stochastic Gradient Algorithms

Study LMS (Least Mean Squares) algorithm  
and its variants.

Recall the optimal estimation problem

$X$  is random variable  
 $\underline{Y}$  is random vector (size:  $M \times 1$ ) } zero Mean

$$\hat{X}_{LMMSE} = \underline{w}_{opt}^* \underline{Y} \quad \text{where} \quad \underline{w}_{opt} = R_Y^{-1} R_{YX} \rightarrow (*)$$

$$\text{SDA: } \underline{w}(n+1) = \underline{w}(n) + \mu (R_{YX} - R_Y \underline{w}(n)) \rightarrow (**)$$

To find optimal filter directly or iteratively,  
(\*) (\*\*)

we need the second order statistics  $R_Y, R_{YX}$ .

In many situations,

(a)  $R_Y, R_{YX}$  may not be known

or

(b)  $R_Y, R_{YX}$  may change over time  
(non stationary)

## Problem Formulation

We are given realizations of  $X$  and  $Y$

$$\{x(0), x(1), \dots\}, \{y(0), y(1), \dots\}$$

$x(n) \rightarrow$  outcome of  $X$  at time instant  $n$

$y(n) \rightarrow$  outcome of  $Y$  at time  $n$

Using these outcomes (data), we want  
to find/learn the optimal filter.

### Approach:

$\rightarrow$  Approximate  $R_y$  &  $R_{yx}$  with  
instantaneous values

$\rightarrow$  Learn optimal filter iteratively

$\rightarrow$  Avoid the need for a priori  
statistical knowledge

$\rightarrow$  Provide a tracking mechanism

$x \rightarrow \text{-----} \rightarrow$

## LMS Algorithm -

$$R_Y = E \{ \underline{Y} \underline{Y}^* \}$$

$$\approx \frac{1}{N} \sum_{n=1}^N \underline{y}(n) \underline{y}^*(n)$$

$$\therefore R_{YX} = E \{ \underline{Y} \underline{X}^* \}$$

$$\approx \frac{1}{N} \sum_{n=1}^N \underline{y}(n) \underline{x}^*(n)$$

At time instant 'n'

$$\text{Instantaneous approximation } \hat{R}_Y(n) = \underline{y}(n) \underline{y}^*(n)$$

$$\hat{R}_{YX}(n) = \underline{y}(n) \underline{x}^*(n)$$

Apply these approximations in SDA.

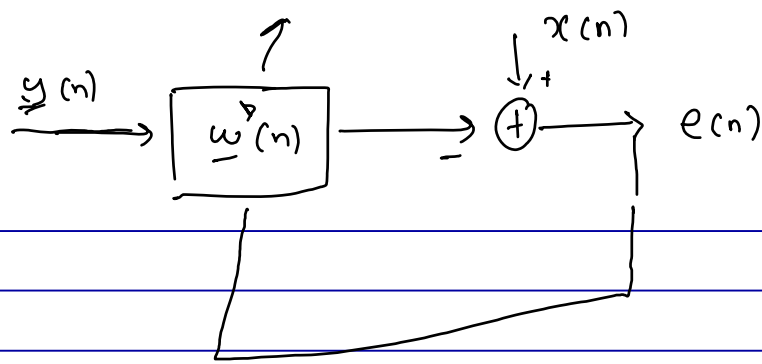
$$\underline{w}(n+1) = \underline{w}(n) + \mu ( R_{YX}(n) - \hat{R}_Y(n) \underline{w}(n) )$$

$$= \underline{w}(n) + \mu ( \underline{y}(n) \underline{x}^*(n) - \underline{y}(n) \underline{y}^*(n) \underline{w}(n) )$$

(Widrow & Hoff)

$$\text{LMS } \underline{w}(n+1) = \underline{w}(n) + \mu \underline{y}(n) [ \underline{x}^*(n) - \underline{y}^*(n) \underline{w}(n) ]$$

$$e^*(n)$$



Remarks

- $\mu > 0$  (usually very small)
- Computationally simple algorithm  
( $2M$  complex multiplications)
- \* Widely used algorithm

16/9

### Least Perturbation Property of LMS

- LMS can be regarded as an exact solution to a local optimization problem
- Want to get an update  $\underline{w}(n+1)$ ,  
from  $\underline{w}(n)$ ,  $\underline{y}(n)$ ,  $x(n)$

$$\text{a priori error } e_a(n) = x(n) - \underline{w}(n) \underline{y}(n)$$

$$\text{posteriori error } e_p(n) = x(n) - \underline{w}(n+1) \underline{y}(n)$$

↓  
after filter update

Consider the optimization problem

$$\min \|\underline{w}(n+1) - \underline{w}(n)\|^2 \quad \text{subject to the}$$

$$\text{Constraint } e_p(n) = (1 - \mu \|\underline{y}(n)\|^2) e_a(n)$$

This constraint is relevant only if

$$|1 - \mu \|\underline{y}(n)\|^2| < 1 \quad \forall n$$

In this case  $|e_p(n)| < |e_a(n)|$

(meaning  $\underline{w}^{\rightarrow}(n+1) \underline{y}(n)$  is a better estimate of  $x(n)$  than  $\underline{w}^{\rightarrow}(n) \underline{y}(n)$ )

Solve the above optimization problem

$$e_a(n) - e_p(n) = \underbrace{\left( \underline{w}(n+1) - \underline{w}(n) \right)^* \underline{y}(n)}_{\underline{\delta w}}$$

from the constraint  $\rightarrow = \mu \|\underline{y}(n)\|^2 e_a(n)$

(local) optimization problem is

at time  $n$   $\min \|\underline{\delta w}\|^2$  such that

$$(*) \rightarrow \left( \underline{\delta w} \right)^* \underline{y}(n) = \mu \|\underline{y}(n)\|^2 e_a(n).$$

There are infinite solutions for (\*)

$$\|\underline{u}\|^2 = \underline{u}^* \underline{u}$$

Consider  $\underline{\delta w}_0 = \mu e_a^*(n) \underline{y}(n)$

This is a solution for (\*).

$$\left[ \begin{aligned} \underline{\delta w}_0^T \underline{y}(n) &= \mu e_a^*(n) \underline{y}^T(n) \underline{y}(n) \\ &= \mu e_a(n) \|\underline{y}(n)\|^2 \end{aligned} \right]$$

Let  $\underline{\delta w}_0 + \underline{z}$  be any other solution for (\*)

$$\text{So } (\underline{\delta w}_0 + \underline{z})^* \underline{y}(n) = \mu e_a(n) \|\underline{y}(n)\|^2$$

$$\cancel{\underline{\delta w}_0^T \underline{y}(n)} + \underline{z}^T \underline{y}(n) = \cancel{\mu e_a(n) \|\underline{y}(n)\|^2}$$

$$\Rightarrow \underline{z}^T \underline{y}(n) = 0$$

$$\text{Now, } \|\underline{\delta w}_0 + \underline{z}\|^2 = (\underline{\delta w}_0 + \underline{z})^* (\underline{\delta w}_0 + \underline{z})$$

$$= \underline{\delta w}_0^T \underline{\delta w}_0 + \underline{z}^T \underline{z} + \underline{\delta w}_0^T \underline{z} + \underline{z}^T \underline{\delta w}_0$$

$$= \|\underline{\delta w}_0\|^2 + \|\underline{z}\|^2 + \underbrace{\mu e_a(n) \underline{y}^T(n) \underline{z}}_0 + \underbrace{\underline{z}^T \underline{y}(n) \mu e_a(n)}_0$$

$$= \|\underline{\delta w}_0\|^2 + \|\underline{z}\|^2$$

$$\geq \|\underline{\delta w}_0\|^2$$

So  $\underline{w}_0$  is the min norm solution satisfying constraint (\*)

$$\underline{w}_0 = \mu e_a(n) y(n)$$

$$\underline{w}_{n+1} - \underline{w}_n = \mu e_a(n) y(n)$$

$$\underline{w}_{(n+1)} = \underline{w}_n + \mu y(n) [x^*(n) - y(n) \underline{w}_n]$$

↓  
LMS update.

x ————— x

## Applications of LMS

1. Channel estimation / System Identification
2. Channel Equalization

## Channel Estimation Problem.

- Channel is modelled as FIR filter (finite impulse response)

$$x(n) = \sum_{l=0}^L h_l^* y(n-l) + \underbrace{v(n)}_{\substack{\text{measurement} \\ \text{noise}}}$$

$\{h_0, h_1, \dots, h_L\}$  are unknown filter coefficients.

Define vectors  $\underline{y}(n) = \begin{bmatrix} y(n) \\ y(n-1) \\ \vdots \\ y(n-L) \end{bmatrix}$

$L \rightarrow$  length of FIR filter.

$$\underline{h} = \begin{bmatrix} h_0 \\ h_1 \\ \vdots \\ h_L \end{bmatrix}$$

$$x(n) = \underline{h}^* \underline{y}(n) + v(n)$$

17/19

Send a random sequence  $\{r(n)\}$  thru the system

Corresponding output

$$X(n) = \underline{h}^* \underline{r}(n) + V(n)$$

Assume that noise  $v(n)$  is uncorrelated with input  $r(n)$

Find  $\underline{w}$  such that

$$E \left| X(n) - \underline{w}^* \underline{r}(n) \right|^2 \text{ is minimum}$$

$$\left. \begin{aligned} R_Y(n) &= E \{ \underline{r}(n) \underline{r}^*(n) \} = R_Y \\ R_{YX}(n) &= E \{ \underline{r}(n) X^*(n) \} = R_{YX} \end{aligned} \right\} \begin{array}{l} \text{for (wide-sense)} \\ \text{stationary} \\ \text{sequences} \end{array}$$

Assuming stationarity.

$$\boxed{R_Y \underline{w}_{opt} = R_{YX}}$$



$$R_{rx} = E \{ \underline{y}(n) \underline{x}^*(n) \}$$

$$= E \{ \underline{y}(n) \cdot ( \underline{h}^* \underline{y}(n) + v(n) )^* \}$$

$$= E \{ \underline{y}(n) \underline{y}^*(n) \underline{h} + \underbrace{\underline{y}(n) v(n)^*}_{0} \}$$

(noise is uncorrelated with input)

$$= R_y \underline{h}$$

$$\underline{h} = R_y^{-1} R_{rx} = \underline{w}_{opt}$$

⇒ direct computation

of optimal filter

We know that SDA can be used

to compute  $\underline{w}_{opt}$  iteratively

(without computing  $R_y^{-1}$  explicitly)

start with  $\underline{w}(0)$  (initial vector)

SDA:

$$\underline{w}(n+1) = \underline{w}(n) + \mu (R_{rx} - R_y \underline{w}(n))$$

Suppose  $R_y$  &  $R_{rx}$  are unknown, then

we can use LMS algorithm to "learn"

the optimal filter (which is same as impulse response of system)

LMS: Instantaneous Approximation

$$R_y = \underline{y}(n) \underline{y}^*(n)$$

$$R_{YX} = \underline{y(n)} \underline{x(n)}^T$$

$$\underline{w(n+1)} = \underline{w(n)} + \mu \underline{y(n)} [ \underline{x(n)} - \underline{y(n)} \underline{w(n)} ]$$

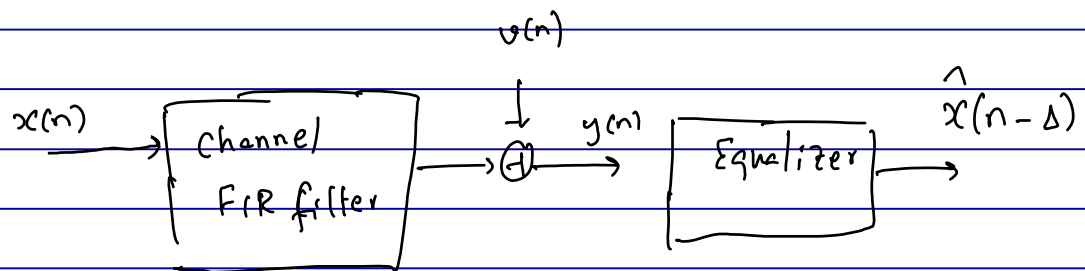
⇓

After convergence  $\underline{w(n)} \approx \underline{h}$

$\{y(n)\}$  is known training signal  
sent thru system.

$\{x(n)\}$  is corresponding output

### channel Equalization.



$x(n)$  is input to channel

$y(n)$  is channel output + noise

From  $y(n)$ , we want to recover  $x(n)$

$\Delta \rightarrow$  is the delay introduced due  
to filter & equalizer.

$X(n) \rightarrow$  random sequence sent thru channel

$V(n) \rightarrow$  noise uncorrelated with  $X(n)$

$$Y(n) = \sum_{i=0}^L h_i x(n-i) + V(n)$$

Let us collect  $M$  observations of  $\{Y(n)\}$

$$\underline{Y(n)} = \begin{bmatrix} Y(n) \\ Y(n-1) \\ \vdots \\ Y(n-M+1) \end{bmatrix}$$

$M \times 1$

$$\begin{bmatrix} Y(n) \\ Y(n-1) \\ \vdots \\ Y(n-M+1) \end{bmatrix} = \underbrace{\begin{bmatrix} h_0 & h_1 & \dots & h_L & 0 & 0 & \dots & 0 \\ 0 & h_0 & h_1 & \dots & h_L & 0 & \dots & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ 0 & 0 & h_0 & h_1 & \dots & h_L & \dots & 0 \end{bmatrix}}_H \begin{bmatrix} X(n) \\ X(n-1) \\ \vdots \\ X(n-M-L+1) \end{bmatrix} + \begin{bmatrix} V(n) \\ \vdots \\ V(n-M+1) \end{bmatrix}$$

$$\underline{Y(n)} = H \underline{X(n)} + \underline{V(n)}$$

From  $\underline{Y(n)}$ , let us recover  $\underline{X(n)}$

$\hat{\underline{X(n)}}$  is recovered from LMMSE  
estimate of  $\underline{X(n)}$  from  $\underline{Y(n)}$

$$\hat{\underline{X(n)}}_{\text{-LMMSE}} = R_{XY}^{-1}(n) \cdot R_Y(n) \underline{Y(n)}$$

Assume  $\{X(n), V(n)\}$  are stationary & uncorrelated.

$$R_Y = E \{ \underline{Y}(n) \underline{Y}(n)^* \}$$

$$= H R_X H^* + R_V$$

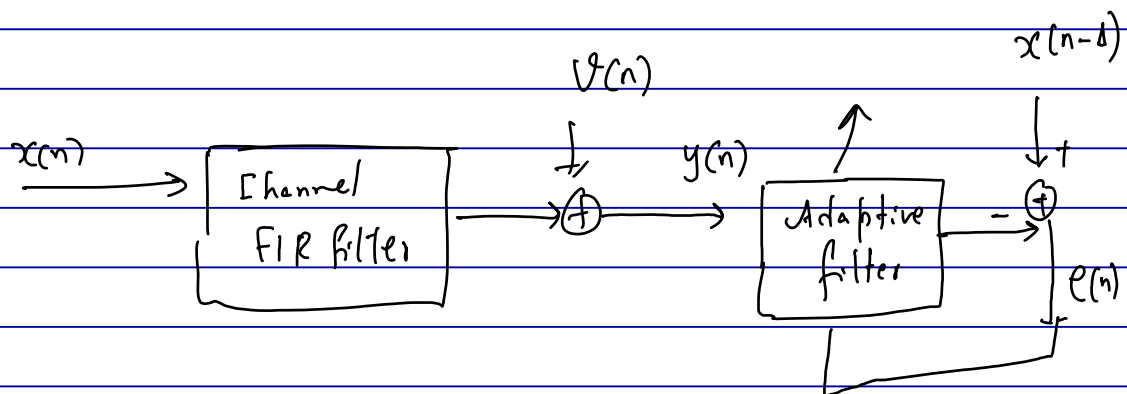
$$R_{YX} = H R_X = R_{XY}$$

$$\hat{\underline{X}}_{\text{LMMSE}}(n) = (H R_X)^* [H R_X H^* + R_V]^{-1} \underline{Y}(n)$$

linear MMSE equalizer

### Adaptive implementation of Equalizer

Send a training signal  $\{x(n)\}$  thru channel.



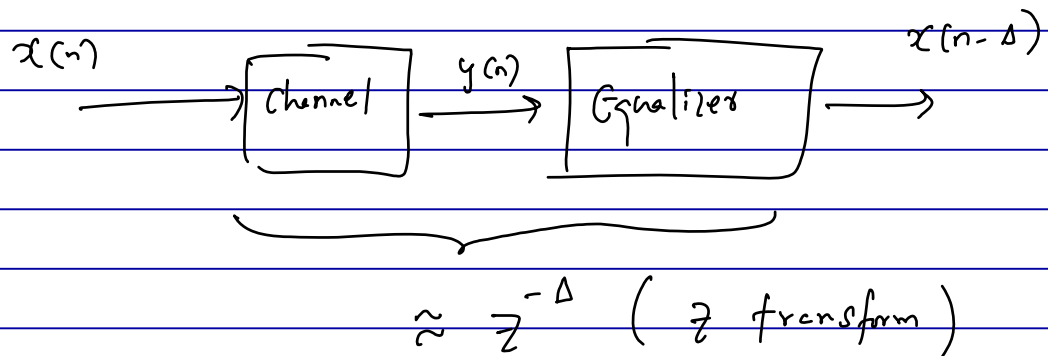
Start with weight vector  $\underline{w}(0)$  for adaptive filter.

LMS update:

$$\underline{w}(n+1) = \underline{w}(n) + \mu \underline{y}(n) [x(n-d) - \underline{y}(n) \underline{w}(n)]$$

After convergence,  $\underline{w}(n)$  will be close to  $\underline{w}_{opt}$ .

Then  $\underline{w}^*(n) y(n)$  will be close to  $x(n-\Delta)$



Once the training phase is over,

we can send (unknown) inputs thru channel

& use the equalizer to get  
back the inputs.

x ————— x

### Normalized LMS algorithm

We had steepest descent algorithm

and used instantaneous approximation to get LMS

$$\text{SDA: } \underline{w}(n+1) = \underline{w}(n) + \mu (R_{yx} - R_y \underline{w}(n))$$

Newton's method:

$$\underline{w}(n+1) = \underline{w}(n) + \mu R_y^{-1} (R_{yx} - R_y \underline{w}(n))$$

Newton's method with regularization:

• If  $R_T$  is not invertible or

$R_T$  is close to a singular matrix

then replace  $R_T^{-1}$  with  $(R_T + \epsilon I)^{-1}$

$$\epsilon > 0$$

$$\underline{w}(n+1) = \underline{w}(n) + \mu (R_T + \epsilon I)^{-1} (R_T x - R_T \underline{w}(n))$$

We use instantaneous approximation on this

Newton's method with regularization.

$$R_T = \underline{y}(n) \underline{y}^*(n) \quad R_T x = \underline{y}(n) x^*(n)$$

$$\underline{w}(n+1) = \underline{w}(n) + \mu [\epsilon I + \underline{y}(n) \underline{y}^*(n)]^{-1} (\underline{y}(n) x^*(n) - \underline{y}(n) \underline{y}^*(n) \underline{w}(n))$$

Using Matrix inversion lemma.

$$[\epsilon I + \underline{y}(n) \underline{y}^*(n)]^{-1} = \epsilon^{-1} I - \frac{\epsilon^{-2} \underline{y}(n) \underline{y}^*(n)}{1 + \epsilon^{-1} \|\underline{y}(n)\|^2}$$

$$\underline{w}(n+1) = \underline{w}(n) + \mu \left( \epsilon^{-1} I - \frac{\epsilon^{-2} \underline{y}(n) \underline{y}^*(n)}{1 + \epsilon^{-1} \|\underline{y}(n)\|^2} \right) \underline{y}(n) [x^*(n) - \underline{y}^*(n) \underline{w}(n)]$$

$$= \underline{w}(n) + \mu \left( \epsilon^{-1} \underline{y}(n) - \frac{\epsilon^{-2} \underline{y}(n) \|\underline{y}(n)\|^2}{1 + \epsilon^{-1} \|\underline{y}(n)\|^2} \right) (x^*(n) - \underline{y}^*(n) \underline{w}(n))$$

$$\underline{w}(n+1) = \underline{w}(n) + \frac{\mu}{\epsilon + \|\underline{y}(n)\|^2} \underline{y}(n) (x^*(n) - \underline{y}^T(n) \underline{w}(n))$$

⇓  
Normalized LMS. (NLMS)

This can be considered as a variable step size LMS with  $\mu(n) = \frac{\mu}{\epsilon + \|\underline{y}(n)\|^2}$

\* ————— \*

Least Perturbation property of  $\epsilon$ -NLMS

Similar to the LMS.

We can verify that  $\epsilon$ -NLMS <sup>update</sup> is the solution to the following optimization problem.

min  $\|\underline{w}(n+1) - \underline{w}(n)\|^2$  subject to the

constraint that

$$e_p(n) = \left( 1 - \frac{\mu \|\underline{y}(n)\|^2}{\epsilon + \|\underline{y}(n)\|^2} \right) e_a(n)$$

$$e_a(n) = x(n) - \underline{w}^T(n) \underline{y}(n)$$

$$e_p(n) = x(n) - \underline{w}^T(n+1) \underline{y}(n)$$

We need  $\mu$  such that

$$\left| 1 - \frac{\mu \|y(n)\|^2}{\varepsilon + \|y(n)\|^2} \right| < 1$$

Equivalently

$$0 < \mu < \frac{2(\varepsilon + \|y(n)\|^2)}{\|y(n)\|^2}$$

It is easy to see that choosing

$0 < \mu < 2$  the above condition is

satisfied irrespective of  $\|y(n)\|^2$

x ————— x

## Various Cost Functions

- So far we have studied MSE cost.

$$\min_{\underline{w}} J(\underline{w}) = E\{|x - \underline{w}^T \underline{y}|^2\}$$

\* Direct computation  $\underline{w}_{opt} = R_y^{-1} R_{yx}$   $-\nabla J(\underline{w})$

Compute  $\nabla_{\underline{w}} J(\underline{w})$

\* SDA :  $\underline{w}(n+1) = \underline{w}(n) + \mu (R_{yx} - R_y \underline{w}(n))$

\* LMS :  $\underline{w}(n+1) = \underline{w}(n) + \mu \underline{y}(n) [x(n) - \underline{y}(n)^T \underline{w}(n)]$

instantaneous approx. to  $R_y, R_{yx}$



## Variants of LMS

\* Signed-LMS.

$$\min_{\underline{w}} E \left\{ |x - \underline{w}^T \underline{r}| \right\}$$

↪ absolute error cost.

$$\underline{w}(n+1) = \underline{w}(n) + \mu y(n) \cdot \text{sign}(x(n) - y(n) \underline{w}(n))$$

$$\text{where } \text{sign}(a + jb) = \text{sign}(a) + j \text{sign}(b)$$

\* Leaky LMS.

$$\min_{\underline{w}} \alpha \|\underline{w}\|^2 + E \left\{ |x - \underline{w}^T \underline{r}|^2 \right\}$$

for some  $\alpha > 0$ .

$$\underline{w}(n+1) = (1 - \mu\alpha) \underline{w}(n) + \mu y(n) [x(n) - y(n) \underline{w}(n)]$$

22/9

\* Least Mean Fourth (LMF) Algorithm.

$$\text{Cost } J(\underline{w}) = E \left\{ |x - \underline{w}^T \underline{r}|^4 \right\}$$

Corresponding update equation is

$$\underline{w}(n+1) = \underline{w}(n) + \mu y(n) [e(n) |e(n)|^2]$$

where  $e(n) = x(n) - \underline{\omega}^T(n) \underline{y}(n)$

\* Least mean - mixed norm [LMMN] Algorithm.

let  $E = X - \underline{\omega}^T Y$

$$J(\underline{\omega}) = E \left\{ \delta |E|^2 + \frac{(1-\delta)|E|^4}{2} \right\}$$

$$0 < \delta < 1$$

update equation

$$\underline{\omega}(n+1) = \underline{\omega}(n) + \mu \underline{y}(n) e^*(n) \left[ \delta + (1-\delta) |e(n)|^2 \right]$$

where  $e(n) = x(n) - \underline{\omega}^T(n) \underline{y}(n)$

x  x

## Blind Algorithms.

- So far, adaptive filters we saw need training data  $\{x(n), x(n), \dots\}$   
&  $\{y(n), y(n), \dots\}$

- In some applications, you may not have access to  $\{x(n)\}$

- In this case, you can still use some properties you know (a priori) about  $x(n)$  to train filter.

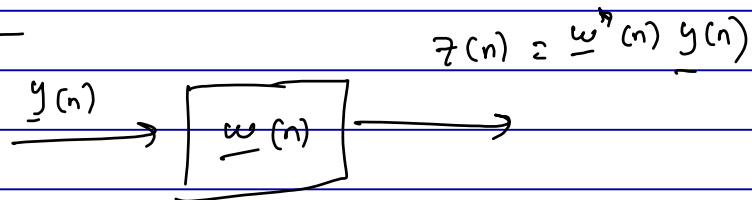
x \_\_\_\_\_ x

## Constant Modulus Algorithms (CMA)

- \* In digital communication, the symbols  $\{x(n)\}$  sent thru channel from transmitter are from a specific constellation. Note that if constellation is PSK, then all symbols  $\{x(n)\}$  have same modulus.

CMA algorithms exploit this constant modulus property of  $\{x(n)\}$  to blindly train the adaptive filter

\* CMA 1



$$\min_{\underline{w}} J(\underline{w}) = E \left\{ |r - |z||^2 \right\}$$

$$z = \underline{w}^T \underline{y}$$

where  $r > 0$

$\rightarrow$  real positive constant.

Update equation:

$$\underline{w}(n+1) = \underline{w}(n) + \mu \underline{y}(n) \left[ r \frac{z(n)}{|z(n)|} - z(n) \right]$$

Note that  $\underline{w}(n+1) = \underline{w}(n)$

if  $|z(n)| = r$

$\epsilon$ -Normalized CMA

$$\underline{w}(n+1) = \underline{w}(n) + \frac{\mu}{E \{ \|\underline{y}(n)\|^2 \}} \underline{y}(n) \left[ r \frac{z(n)}{|z(n)|} - z(n) \right]$$

• CMA 2

$$J(\underline{\omega}) = E \left\{ |r - |z|^2|^2 \right\}$$

Update equation is  $z = \underline{\omega}^T \underline{y}$

$$\underline{\omega}(n+1) = \underline{\omega}(n) + \mu \underline{y}(n) z(n) [r - |z(n)|^2]$$

$$z(n) = \underline{\omega}^T(n) \underline{y}(n).$$

Other Variants of CMA.

\* Reduced Constellation Algorithm (RCA)

$$J(\underline{\omega}) = E \left\{ \left| \underline{\omega}^T \underline{y} - r \operatorname{sign}(\underline{\omega}^T \underline{y}) \right|^2 \right\}$$

Update Equation

$$\underline{\omega}(n+1) = \underline{\omega}(n) + \mu \underline{y}(n) [r \operatorname{sign}(z(n)) - z(n)]$$

$$z(n) = \underline{\omega}^T(n) \underline{y}(n)$$

\* Multi-Modulus Algorithm (MMA)

$$J(\underline{\omega}) = E \left\{ \left[ \left( \operatorname{Re} \{ \underline{\omega}^T \underline{y} \} \right)^2 - r \right]^2 + \left[ \left( \operatorname{Im} \{ \underline{\omega}^T \underline{y} \} \right)^2 - r \right]^2 \right\}$$

Update equation:

$$z(n) = \underline{w}(n) \underline{y}(n)$$

$$a(n) = \operatorname{Re}(z(n))$$

$$b(n) = \operatorname{Im}(z(n))$$

$$e(n) = a(n) [\gamma - a^2(n)] + j b(n) [\gamma - b^2(n)]$$

$$\underline{w}(n+1) = \underline{w}(n) + \mu \underline{y}(n) e^*(n)$$

\*

\_\_\_\_\_ \*

↑  
Portions  
for  
quiz 2

Affine Projection Algorithm.

• We go back to the MSE cost

$$J(\underline{w}) = E \{ |X - \underline{w}^T \underline{r}|^2 \}$$

LMS Approximation

$$R_{\underline{r}} = \hat{R}_{\underline{r}}(n) = \underline{y}(n) \underline{y}^T(n)$$

Recall Newton method with

regularization

$$(*) \quad \underline{w}(n+1) = \underline{w}(n) + \mu \left( \epsilon \underline{I} + R_{\underline{r}} \right)^{-1} [R_{\underline{r}} \underline{w}(n) - R_{\underline{r}} \underline{w}(n)]$$

At  $n$ th instant, we use past  $k$

$$\text{realizations } \left\{ \underline{y}(n), \underline{y}(n-1), \dots, \underline{y}(n-k+1) \right\}$$
$$\left\{ \underline{x}(n), \underline{x}(n-1), \dots, \underline{x}(n-k+1) \right\}$$

to estimate  $R_y$  and  $R_{yx}$

$$\hat{R}_y = \frac{1}{k} \sum_{i=n-k+1}^n \underline{y}(i) \underline{y}^{\Delta}(i)$$

$$\hat{R}_{yx} = \frac{1}{k} \sum_{i=n-k+1}^n \underline{y}(i) \underline{x}^{\Delta}(i)$$

Using these estimates in Newton's method,  
we get affine projection algorithm.

23/9

Let us define matrix

$$\rightarrow Y_n = \begin{bmatrix} \underline{y}(n) & \underline{y}(n-1) & \dots & \underline{y}(n-k+1) \end{bmatrix}$$

$M \times k$

$$\rightarrow X_n = \begin{bmatrix} \underline{x}(n) & \underline{x}(n-1) & \dots & \underline{x}(n-k+1) \end{bmatrix}$$

$k \times 1$

$$Y_n Y_n^* = \begin{bmatrix} \underline{y}(n) & \dots & \underline{y}(n-k+1) \end{bmatrix} \begin{bmatrix} \underline{y}^{\Delta}(n) \\ \underline{y}^{\Delta}(n-1) \\ \vdots \\ \underline{y}^{\Delta}(n-k+1) \end{bmatrix}$$

$$= \underline{y}(n) \underline{y}^{\star}(n) + \underline{y}(n-1) \underline{y}^{\star}(n-1) + \dots + \underline{y}(n-k+1) \underline{y}^{\star}(n-k+1)$$

$$\hat{R}_Y = \frac{\underline{Y}_n \underline{Y}_n^{\star}}{k}$$

$$\hat{R}_{YX} = \frac{1}{k} \underline{Y}_n \underline{X}_n^{\star}$$

using these approximations in Newton's method

Take  $\epsilon^l$  in (\*) as  $\epsilon/k$

Newton's method becomes

$$\underline{w}(n+1) = \underline{w}(n) + \mu \left( \frac{\epsilon}{k} \mathbf{I} + \frac{1}{k} \underline{Y}_n \underline{Y}_n^{\star} \right)^{-1} \left( \frac{1}{k} \underline{Y}_n \underline{X}_n^{\star} - \frac{1}{k} \underline{Y}_n \underline{Y}_n^{\star} \underline{w}(n) \right)$$

ie)

$$\underline{w}(n+1) = \underline{w}(n) + \mu \left( \epsilon \mathbf{I} + \underline{Y}_n \underline{Y}_n^{\star} \right)^{-1} \left( \underline{Y}_n \underline{X}_n^{\star} - \underline{Y}_n \underline{Y}_n^{\star} \underline{w}(n) \right)$$

$\Downarrow$

$\epsilon$  - Affine projection algorithm (APA)

$\epsilon \mathbf{I} + \underline{Y}_n \underline{Y}_n^{\star} \rightarrow M \times M$  matrix

Using Matrix inversion lemma, we can

write this inverse in terms of  $k \times k$  inverse.  
(Typically  $k \leq M$ )



$$(A + BCD)^{-1} = \bar{A}^{-1} - \bar{A}^{-1} B (\bar{C}^{-1} + D \bar{A}^{-1} B)^{-1} D \bar{A}^{-1}$$

$$(\varepsilon I + Y_n Y_n^*)^{-1} = ?$$

$$A = \varepsilon I \quad B = Y_n \quad C = I \quad D = Y_n^*$$

$$(\varepsilon I + Y_n Y_n^*)^{-1} = \frac{1}{\varepsilon} I - \frac{1}{\varepsilon} I Y_n \left( I + \frac{Y_n^* I Y_n}{\varepsilon} \right)^{-1} \cdot \left( \frac{Y_n^* I}{\varepsilon} \right)$$

$$= \frac{1}{\varepsilon} \left( I - Y_n \underbrace{\left( \varepsilon I + Y_n^* Y_n \right)^{-1}}_{k \times k \text{ matrix}} Y_n^* \right)$$

$$(\varepsilon I + Y_n Y_n^*)^{-1} Y_n = \frac{1}{\varepsilon} \left( I - Y_n \left( \varepsilon I + Y_n^* Y_n \right)^{-1} Y_n^* \right) Y_n$$

$$= \frac{Y_n (I)}{\varepsilon} - \frac{Y_n}{\varepsilon} \left( \varepsilon I + Y_n^* Y_n \right)^{-1} Y_n^* Y_n$$

$$= \frac{Y_n}{\varepsilon} \left( \varepsilon I + Y_n^* Y_n \right)^{-1} \left( \varepsilon I + Y_n^* Y_n \right)$$

$$- \frac{Y_n}{\varepsilon} \left( \varepsilon I + Y_n^* Y_n \right)^{-1} Y_n^* Y_n$$

$$= \frac{Y_n}{\varepsilon} \left( \varepsilon I + Y_n^* Y_n \right)^{-1} \left( \varepsilon I + Y_n^* Y_n - Y_n^* Y_n \right)$$

$$= \gamma_n (\epsilon I + \gamma_n^* \gamma_n)^{-1}$$

Update eqn of  $\epsilon$ -APA becomes

$$\underline{w}(n+1) = \underline{w}(n) + \mu \gamma_n (\epsilon I + \gamma_n^* \gamma_n)^{-1} (\underline{x}_n^* - \gamma_n^* \underline{w}(n))$$

↓

involves  $k \times k$  matrix inversion

which requires  $(O(k^3))$  operations

(additions & multiplications)

For  $k=1$ , we get  $\epsilon$ -NLMS

$$\underline{w}(n+1) = \underline{w}(n) + \mu \frac{\underline{y}(n)}{\epsilon + \|\underline{y}(n)\|^2} (\underline{x}_n^* - \underline{y}(n) \underline{w}(n))$$

\*  $\epsilon$ -APA uses vector valued estimation error

$$\left( \underline{x}_n^* - \gamma_n^* \underline{w}(n) \right)$$

\* We are reusing data. in update eqn.

\*  $k$  is referred as order of  $\epsilon$ -APA.

\*  $\xrightarrow{\hspace{10em}}$

Least Perturbation Property.

Similar to (LMS,  $\epsilon$ -NLMS) the  $\epsilon$ -APA can also be seen as an exact solution to a local optimization problem

Let us define a priori & posteriori estimation error vectors.

$$\underline{e}_a(n) = \underline{x}(n) - \underline{Y}_n \underline{w}(n)$$

$$\underline{e}_p(n) = \underline{x}(n) - \underline{Y}_n \underline{w}(n+1)$$

$\epsilon$ -APA update equation is the solution for the following optimization problem.

$$\min_{\underline{w}(n+1)} \|\underline{w}(n+1) - \underline{w}(n)\|^2 \quad \text{subject to}$$

$$\underline{e}_p(n) = \left( \underline{I} - \mu \underline{Y}_n^* \underline{Y}_n (\epsilon \underline{I} + \underline{Y}_n^* \underline{Y}_n)^{-1} \right) \underline{e}_a(n)$$

Variation of APA.

Partial Rank Algorithm.

\* Weight vector is updated once out of  $k$  iterations

$$\underline{w}(n+1) = \underline{w}(n) + \mu Y_n (\epsilon I + Y_n^T Y_n)^{-1} (x_n - Y_n \underline{w}(n))$$

$$\underline{w}(n+1) = \underline{w}(n+2) = \underline{w}(n+3) = \dots = \underline{w}(n+k)$$

$\underline{w}(n+k)$  is computed as per update eqn.

\* ————— \*

24/9

## Recursive Least Squares (RLS)

- RLS uses more sophisticated approximation of Covariance matrix  $R_y$  using all the past data
- RLS is exact solution for a least squares estimation problem. Will see more details later. Right now we see RLS as stochastic Gradient method.

Newton's Method (with regularization)

$$\underline{w}(n+1) = \underline{w}(n) + \mu_n (\epsilon_n I + R_y)^{-1} (R_{y_x} - R_y \underline{w}(n))$$

gradient term

For Gradient term,

$$R_{yx} - R_y \underline{w}(n) \approx \underline{y}(n) \underline{x}(n) - \underline{y}(n) \underline{y}(n) \underline{w}(n)$$

For computing  $(\epsilon_n \mathbf{I} + R_y)^{-1}$  we use

the following approximation for  $R_y$ .

$$R_y \approx \frac{1}{n+1} \sum_{i=0}^n \lambda^{n-i} \underline{y}(i) \underline{y}(i) = \hat{R}_y$$

We choose  $0 < \lambda < 1$

$\Downarrow$

This helps in providing higher weightage to recent data in averaging.

$\lambda = 1 \Rightarrow$

all data

gets equal weightage

$\lambda \rightarrow$  forgetting factor

Choose  $\mu_n = \frac{1}{n+1}$        $\epsilon_n = \frac{\lambda^{n+1}}{n+1} \epsilon$

with these choices, newton method approximation

becomes

$$\underline{w}(n+1) = \underline{w}(n) + \frac{1}{n+1} \left( \frac{\lambda^{n+1}}{n+1} \epsilon \mathbf{I} + \frac{1}{n+1} \sum_{i=0}^{n+1} \lambda^{n-i} \underline{y}(i) \underline{y}(i) \right)^{-1} \left( \underline{y}(n) \underline{x}(n) - \underline{y}(n) \underline{y}(n) \underline{w}(n) \right)$$

$$= \underline{\omega}(n) + \left( \lambda^{n+1} \varepsilon \mathbf{I} + \sum_{i=0}^n \lambda^{n-i} \underline{y}(i) \underline{y}^*(i) \right)^{-1} \left( \underline{y}(n) \underline{x}(n) - \underline{y}(n) \underline{y}^*(n) \underline{\omega}(n) \right)$$

$$\Phi_n = \lambda^{n+1} \varepsilon \mathbf{I} + \sum_{i=0}^n \lambda^{n-i} \underline{y}(i) \underline{y}^*(i)$$

We want  $\Phi_n^{-1}$ ; Let  $P_n = \Phi_n^{-1}$

We will compute  $P_n$  using matrix inversion lemma.

First write  $\Phi_n$  using a recursion

$$\Phi_n = \lambda \left( \lambda^n \varepsilon \mathbf{I} + \sum_{i=0}^{n-1} \lambda^{n-1-i} \underline{y}(i) \underline{y}^*(i) \right) + \underline{y}(n) \underline{y}^*(n)$$

$$= \lambda \Phi_{n-1} + \underline{y}(n) \underline{y}^*(n)$$

$$\Phi_n^{-1} = P_n = \left( \lambda \Phi_{n-1} + \underline{y}(n) \underline{y}^*(n) \right)^{-1}$$

using matrix  
inversion  
lemma

$$\rightarrow = \lambda^{-1} \Phi_{n-1}^{-1} - \lambda^{-2} \Phi_{n-1}^{-1} \underline{y}(n)$$

$$\left( 1 + \lambda^{-1} \underline{y}(n) \Phi_{n-1}^{-1} \underline{y}(n) \right)^{-1} \underline{y}(n) \Phi_{n-1}^{-1}$$

Note

$$\phi_{n-1}^{-1} = P_{n-1}$$

$$P_n = \lambda^{-1} \left[ P_{n-1} - \frac{\lambda^{-1} P_{n-1} \underline{y}(n) \underline{y}(n)^T P_{n-1}}{1 + \lambda^{-1} \underline{y}(n)^T P_{n-1} \underline{y}(n)} \right]$$

Initialization  $P_{-1} = \varepsilon^{-1} I$

RLS algorithm:

$$\underline{w}(n+1) = \underline{w}(n) + P_n \underline{y}(n) [x(n) - \underline{y}(n)^T \underline{w}(n)]$$

$$P_n = \lambda^{-1} \left[ P_{n-1} - \frac{\lambda^{-1} P_{n-1} \underline{y}(n) \underline{y}(n)^T P_{n-1}}{1 + \lambda^{-1} \underline{y}(n)^T P_{n-1} \underline{y}(n)} \right]$$

$$P_{-1} = \varepsilon^{-1} I \quad \text{and} \quad 0 < \lambda < 1$$

Using Matrix inversion lemma, we have

converted a matrix inversion  $(\phi_n^{-1})$  into

a scalar inversion  $[1 + \lambda^{-1} \underline{y}(n)^T P_{n-1} \underline{y}(n)]^{-1}$