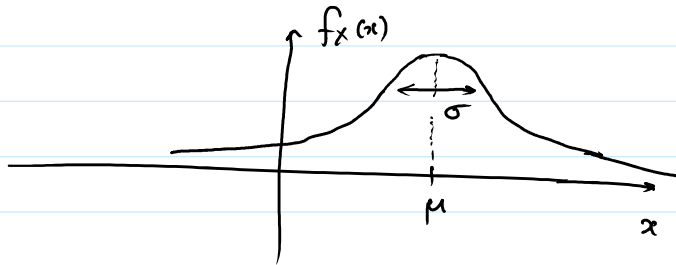


## Gaussian Basics

$X$  is a Gaussian random variable  
with mean  $\mu$  and Variance  $\sigma^2$

$$f_X(x) = \frac{1}{\sqrt{2\pi} \sigma} e^{-\frac{(x-\mu)^2}{2\sigma^2}}$$



Shorthand Notation:  $X \sim N(\mu, \sigma^2)$

Note: Suppose  $X \sim N(\mu, \sigma^2)$

$$\text{Let } Y = aX + b$$

$a, b$  constants

$$\text{Then } Y \sim N(a\mu + b, a^2\sigma^2)$$

Gaussian pdf is completely characterized  
by its Mean & Variance.

← ————— →

Definition of Q function

Suppose  $X \sim N(0, 1)$

↓  
standard normal pdf

(ccdf) Complementary cdf of  $X$

$$P(X > x) = \int_x^{\infty} f_X(t) dt$$

$$= \frac{1}{\sqrt{2\pi}} e^{-t^2/2}$$

$$= \int_x^\infty \frac{1}{\sqrt{2\pi}} e^{-t^2/2} dt$$

definition  
of Q

$$Q(x) = \int_x^\infty \frac{1}{\sqrt{2\pi}} e^{-t^2/2} dt$$

$$= 1 - P(X \leq x) = \text{cdf}$$

- $Q(x)$  is a decreasing function of  $x$
- $Q(x) \leq \frac{1}{2} e^{-x^2/2}$

x ————— x

Suppose  $Y \sim N(\mu, \sigma^2)$

$$\text{Let } z = \frac{Y - \mu}{\sigma}$$

Note that  $z \sim N(0, 1)$

Now,  $P(Y > a)$

$$= P\left(\frac{Y - \mu}{\sigma} > \frac{a - \mu}{\sigma}\right)$$

$$= P\left(z > \frac{a - \mu}{\sigma}\right)$$

$$= Q\left(\frac{a - \mu}{\sigma}\right)$$

x ————— x

Central limit theorem <sup>(CLT)</sup> (Significance of Gaussian)

Suppose  $X_1, X_2, \dots, X_n, \dots$  be

a sequence of i.i.d. random variables  
(independent and identically distributed)

with mean  $\mu$  and variance  $\sigma^2$

Consider the "average"

$$S_N = \frac{1}{\sqrt{N}} \sum_{i=1}^N \frac{x_i - \mu}{\sigma}$$

According to C.L.T.

$S_N$  converges in distribution to  $N(0,1)$

that is  $\lim_{N \rightarrow \infty} P\{S_N > x\} = Q(x)$

x ————— x

Noise in communication system

comes from various sources (independent)

- thermal noise
- quantization noise
- shot noise
- noise in physical medium

Effective noise which is sum of

the noise from various sources

can be approximated with Gaussian statistics

x ————— x

Gaussian Random Vector

$$\underline{x}_{P \times 1} = \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_P \end{bmatrix}$$

$x_1, x_2, \dots, x_p$  are called jointly

Gaussian random variables or

$\underline{x}$  is joint Gaussian random vector

if  $a_1 x_1 + a_2 x_2 + \dots + a_p x_p$  is

a Gaussian random variable

for any real values of  $a_1, a_2, \dots, a_p$

Recall mean vector

$$\underline{\mu}_x = \begin{bmatrix} E(x_1) \\ \vdots \\ E(x_p) \end{bmatrix}$$

Covariance Matrix

$$C_x = E \{ (\underline{x} - \underline{\mu}_x) (\underline{x} - \underline{\mu}_x)^T \}$$

$$= \begin{bmatrix} c_{11} & c_{12} & \dots & c_{1p} \\ \vdots & & & \\ c_{p1} & \dots & \dots & c_{pp} \end{bmatrix}$$

Let  $\mu_i = E(x_i)$

$$c_{ij} = E \{ (x_i - \mu_i) (x_j - \mu_j) \}$$

$$= \text{Cov}(x_i, x_j)$$

$$c_{ii} = \text{Var}(x_i) = \sigma_i^2$$

Note  $x_i \sim N(\mu_i, \sigma_i^2)$

Joint pdf of  $\underline{x}$  is completely

characterized by  $\underline{\mu}_x$  &  $C_x$

$$f_{\underline{x}}(\underline{x}) = \frac{1}{(2\pi)^{p/2} |C_x|^{1/2}} e^{-\frac{1}{2} (\underline{x} - \underline{\mu}_x)^T C_x^{-1} (\underline{x} - \underline{\mu}_x)}$$

$$f_{\underline{x}}(\underline{x}) = \frac{1}{(2\pi)^{p/2} |C_x|^{1/2}} e^{-\frac{1}{2} (\underline{x}-\underline{\mu})^T C_x^{-1} (\underline{x}-\underline{\mu})}$$

↙ determinant

$$\underline{x} = \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_p \end{bmatrix}$$

[Assume  $C_x$  is invertible]

Consider case:  $x_1, x_2, \dots, x_p$  are uncorrelated

$$\text{Cov}(x_i, x_j) = 0$$

$$C_x = \begin{bmatrix} c_{11} & & 0 \\ & c_{22} & \\ 0 & & \ddots \\ & & & c_{pp} \end{bmatrix}$$

$$\sigma_i^2 = \text{Var}(x_i) = \begin{bmatrix} \sigma_1^2 & & 0 \\ & \sigma_2^2 & \\ 0 & & \ddots \\ & & & \sigma_p^2 \end{bmatrix}$$

If  $x_i$ 's are uncorrelated,  
Covariance matrix is diagonal

Let us find joint pdf for this case

Note  $C_x^{-1} = \begin{bmatrix} 1/\sigma_1^2 & & 0 \\ & 1/\sigma_2^2 & \\ 0 & & \ddots \\ & & & 1/\sigma_p^2 \end{bmatrix}$

determinant

$$|C_x| = \prod_{i=1}^p \sigma_i^2$$

$$f_{\underline{x}}(\underline{x}) = \frac{1}{(2\pi)^{p/2} \prod_{i=1}^p \sigma_i} e^{-\frac{1}{2} (\underline{x}-\underline{\mu})^T C_x^{-1} (\underline{x}-\underline{\mu})}$$

$$f_{\underline{x}}(\underline{x}) = \frac{1}{(2\pi)^{p/2}} \prod_{i=1}^p \frac{1}{\sigma_i} e^{-\frac{1}{2} (\underline{x}-\underline{\mu})^T C_x^{-1} (\underline{x}-\underline{\mu})}$$

$$(\underline{x}-\underline{\mu})^T C_x^{-1} (\underline{x}-\underline{\mu})$$

$$= [(x_1 - \mu_1) \quad (x_2 - \mu_2) \dots (x_p - \mu_p)]$$

$$\times \begin{bmatrix} \frac{1}{\sigma_1^2} & & 0 \\ & \frac{1}{\sigma_2^2} & \\ 0 & & \ddots \\ & & & \frac{1}{\sigma_p^2} \end{bmatrix}$$

$$\times \begin{bmatrix} x_1 - \mu_1 \\ x_2 - \mu_2 \\ \vdots \\ x_p - \mu_p \end{bmatrix}$$

$$= \sum_{i=1}^p \frac{(x_i - \mu_i)^2}{\sigma_i^2}$$

$$e^{-\frac{1}{2} \sum_{i=1}^p \frac{(x_i - \mu_i)^2}{\sigma_i^2}}$$

$$f_{\underline{x}}(\underline{x}) = \frac{1}{(2\pi)^{p/2}} \prod_{i=1}^p \frac{1}{\sigma_i} e^{-\frac{1}{2} \sum_{i=1}^p \frac{(x_i - \mu_i)^2}{\sigma_i^2}}$$

$$= \prod_{i=1}^p \frac{1}{(2\pi)^{1/2} \sigma_i} e^{-\frac{1}{2} \frac{(x_i - \mu_i)^2}{\sigma_i^2}}$$

$$= \prod_{i=1}^p f_{x_i}(x_i)$$

r

$$i=1$$

$\downarrow$   
Joint pdf = product of marginal pdfs

Hence if  $x_i$ 's are jointly Gaussian  
and  $x_i$ 's are uncorrelated then  
 $x_i$ 's are independent.

x \_\_\_\_\_ x

• Any linear transformation of  
Gaussian random vector is also  
a Gaussian random vector.

$$\underline{x} = \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_p \end{bmatrix} \quad \underline{x} \sim N(\underline{\mu}_x, C_x)$$

Consider  $\underline{y} = \underline{A} \underline{x} + \underline{b}$   
 $\begin{matrix} M \times 1 & M \times p & M \times 1 \end{matrix}$   
( $\underline{A}, \underline{b}$  are constant  
matrix/vector)

We have  $\underline{y}$  is also a Gaussian vector

$$\underline{y} \sim N(\underbrace{\underline{A} \underline{\mu}_x + \underline{b}}, \underbrace{\underline{A} C_x \underline{A}^T})$$

$\mu_r$        $C_r$

$x$  —————  $x$

## Gaussian Random Process

$\{X(t)\}$  is called a Gaussian Random Process if the samples  $\{X(t_1), X(t_2), \dots, X(t_n)\}$  are jointly Gaussian for any sampling instants  $\{t_1, t_2, \dots, t_n\}$  and  $n$  is any integer

Recall Mean function

$$\mu(t) = E\{X(t)\}$$

Auto Correlation function ACF

$$R_x(t; t-\tau) = E\{X(t)X(t-\tau)\}$$

From  $\mu(t)$  &  $R_x(t; t-\tau)$  we

can find covariance of  $X(t_1), X(t_2)$  as

$$\begin{aligned} \text{Cov}(X(t_1), X(t_2)) &= E\{X(t_1)X(t_2)\} \\ &\quad - E\{X(t_1)\}E\{X(t_2)\} \\ &= R_x(t_1; t_1 - (t_1 - t_2)) \end{aligned}$$



$$- \mu(t_1) \mu(t_2)$$

Since Mean & Covariance characterize  
pdf of <sup>(joint)</sup> Gaussian random variables

$\mu(t)$  &  $R_x(t; t-\tau)$  characterize

Gaussian RP completely

• Recall for Wide Sense Stationary RP

$$\mu(t) = \mu \quad (\text{constant})$$

ACF:  $E(x(t)x(t-\tau))$  depends only on  $\tau$

$$\downarrow$$

denote  $R_x(\tau) = E(x(t)x(t-\tau))$

• ~~IP~~ Note  $\text{Cov}(x(t_1), x(t_2))$

$$= \text{Cov}(x(t_1+d), x(t_2+d))$$

for any  $t_1, t_2, d$ .

• If a Gaussian RP is  
WSS then it is also stationary

$\{x(t)\}$  &  $\{x(t-d)\}$  have

same statistics for any delay  $d$

- Recall PSD for WSS

$$S_x(f) = \int_{-\infty}^{\infty} R_x(\tau) e^{-j2\pi f\tau} d\tau$$

- For WSS Gaussian RP with

zero mean ( $\mu(t) = 0$ ) the

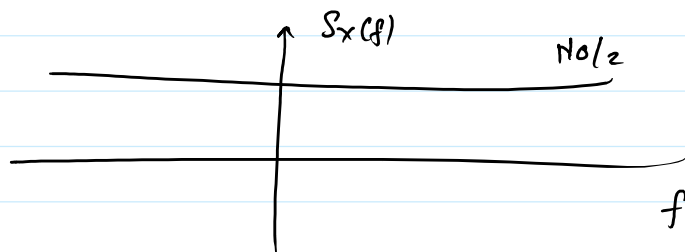
PSD is equivalent characterization  
for RP

- A white Gaussian RP is

a WSS Gaussian RP

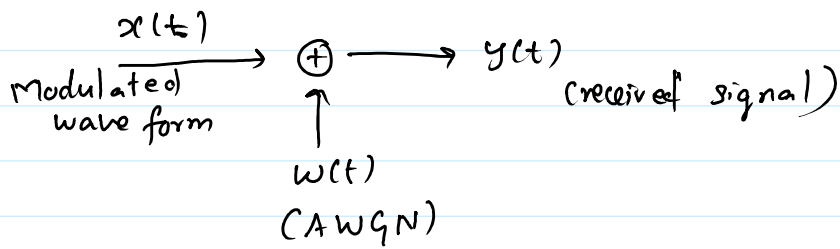
with  $\mu(t) = 0 \quad \forall$

$S_x(f) = \text{constant for all } f$



- We model noise  $w(t)$  in our communication system as white Gaussian Random process with PSD  $N_0/2 \equiv \sigma^2$

Additive white Gaussian noise model



$w(t)$  is white Gaussian with  
 (two-sided) PSD  $\frac{N_0}{2} \equiv \sigma^2$

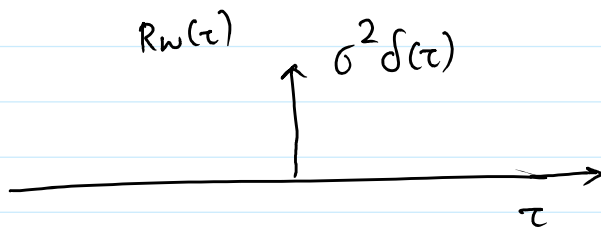
- Total noise power

$$\begin{aligned}
 \int_{-\infty}^{\infty} S_w(f) df &= \int_{-\infty}^{\infty} N_0/2 df \\
 &= \infty
 \end{aligned}$$

- ACF of white noise

$\hat{=}$  Inverse F.T. of PSD

$$R_w(\tau) = \frac{N_0}{2} \delta(\tau) = \sigma^2 \delta(\tau)$$



- $E\{w(t)w(t-\tau)\} = 0$   
if  $\tau \neq 0$

$$\Rightarrow E(w(t_1)w(t_2)) = 0$$

if  $t_1 \neq t_2$

$$\Rightarrow \text{Cov}(w(t_1), w(t_2)) = 0$$

if  $t_1 \neq t_2$

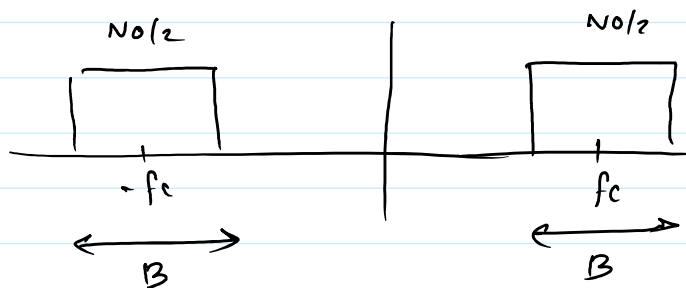
$\Rightarrow$  Samples of white noise at two different instants are independent.

• white noise with infinite power

is a simplified description of noise. Noise gets

filtered thru the receiver processing and the filtered noise has finite power

• After (Passband) filtering



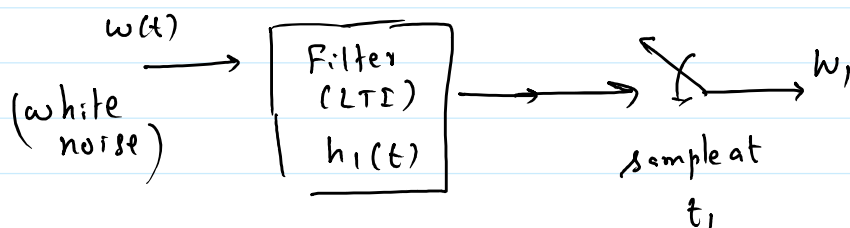
Total noise power (after filtering)

$$P_n = \frac{N_0}{2} (2B)$$

$$= N_0 B$$

(finite power)

- Properties of filtering white noise (and sampling)



$$w_1 = \int_{-\infty}^{\infty} w(t) \underbrace{h_1(t_1 - t)} dt$$

$$\text{Let } u_1(t) = h_1(t_1 - t)$$

filtering + sampling  
 ↓ is equivalent  
 to inner product

$$w_1 = \int_{-\infty}^{\infty} w(t) u_1(t) dt = \langle w(t), u_1(t) \rangle$$

Similarly let  $u_2(t)$  be  
 another function

$$\text{Let } w_2 = \int_{-\infty}^{\infty} w(t) u_2(t) dt = \langle w(t), u_2(t) \rangle$$

Claims:

- $w_1$  &  $w_2$  are jointly Gaussian
- $E(w_1) = E(w_2) = 0$
- $\text{Var}(w_1) = \frac{N_0}{2} \|u_1(t)\|^2$   
 $= \frac{N_0}{2} \int_{-\infty}^{\infty} |u_1(t)|^2 dt$
- $\text{Var}(w_2) = \frac{N_0}{2} \|u_2(t)\|^2$

- Covariance  $\text{Cov}(W_1, W_2) = \frac{N_0}{2} \langle u_1(t), u_2(t) \rangle$

- If  $u_1(t), u_2(t)$  are orthogonal then  $W_1$  &  $W_2$  are independent.

Proof:

due to zero mean

$$\text{Cov}(W_1, W_2) \stackrel{\downarrow}{=} E(W_1, W_2)$$

$$= E \left\{ \left( \int_{-\infty}^{\infty} w(t_1) u_1(t_1) dt_1 \right) \left( \int_{-\infty}^{\infty} w(t_2) u_2(t_2) dt_2 \right) \right\}$$

$$= E \left\{ \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} w(t_1) w(t_2) u_1(t_1) u_2(t_2) dt_1 dt_2 \right\}$$

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \underbrace{E \{ w(t_1) w(t_2) \}}_{\frac{N_0}{2} \delta(t_1 - t_2)} u_1(t_1) u_2(t_2) dt_1 dt_2$$

$$= \int_{t_1 = -\infty}^{\infty} \frac{N_0}{2} u_1(t_1) u_2(t_1) dt_1$$

$$= \frac{N_0}{2} \langle u_1(t), u_2(t) \rangle$$