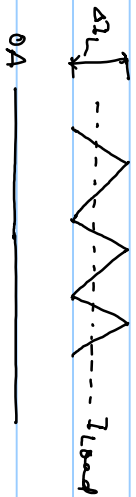
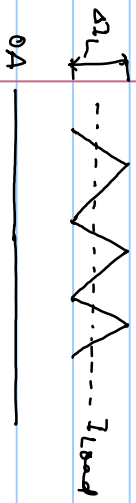
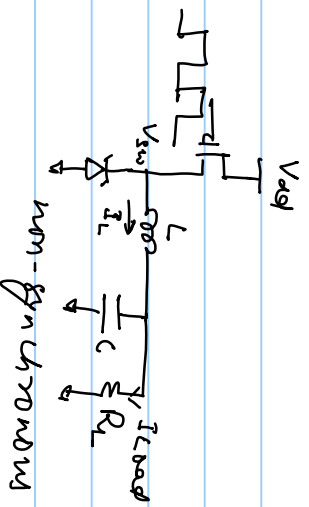
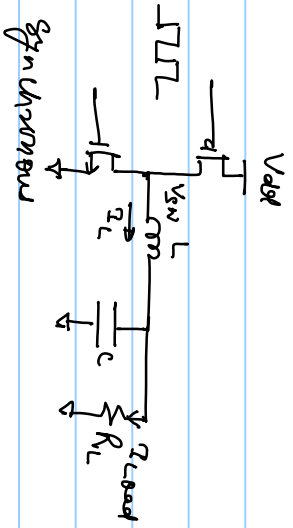


# CCM vs. DCM operation

CCM  $\rightarrow$  continuous conduction mode

DCM  $\rightarrow$  discontinuous conduction mode



$$I_{load} = \frac{\Delta I_L}{2}$$

$$I_{load} = \frac{\Delta I_L}{2}$$



$$I_{load} < \frac{\Delta I_L}{2}$$

$$I_{load} < \frac{\Delta I_L}{2}$$

## synchronous

- ① Always reverse current  
so inductor is always  
conducting

- ② Losses are more due to  
reverse conduction.

- ③ Works for all load currents  
(positive & negative)

- ④  $V_o = D \cdot V_{dc}$  holds  
true

## non-synchronous

- Doesn't allow reverse current  
so inductor doesn't conduct  
if  $I_{load} < \frac{\Delta I_L}{2}$

- Losses are reduced.  
→ better efficiency

- works only for positive  
load currents  
 $I_{load} > 0$

- $V_o = D \cdot V_{dc}$  doesn't hold  
true.

CCM - DCM Boundary Condition.

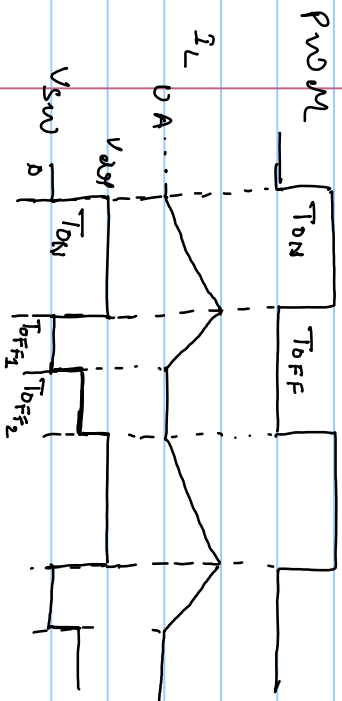
$$I_{L\text{load}} = \frac{A_i I_L}{2} = \frac{1}{2} \frac{V_0}{L} (1-D) T_{sw}$$

$$1-D = \frac{I_{L\text{load}}}{\frac{T_{sw}}{2}} \times \frac{L}{V_0}$$

$$I_{L\text{load}} = \frac{V_0}{R_L}$$

$$1-D = \frac{R_L}{L T_{sw}}$$

$$r D' = \frac{R_L}{R_L T_{sw}} = k \text{ (critical factor)} \quad \text{--- (1)}$$



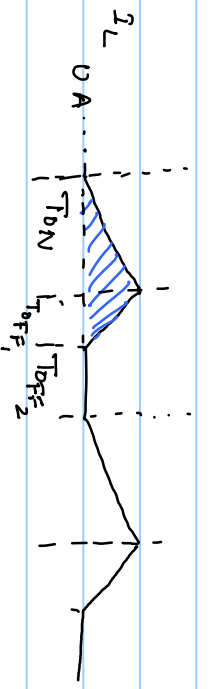
Average voltage across inductor.

$$V_L(\text{avg}) = 0 = \frac{1}{T_{sw}} \left( \frac{V_{DD} - V_0}{L} \times T_{ON} + \frac{(-V_0)}{L} T_{OFF1} + 0 \times T_{OFF2} \right) = 0$$

$$\Rightarrow (V_{DD} - V_0) T_{ON} - V_0 T_{OFF1} = 0$$

$$\Rightarrow T_{ON} + T_{OFF1} = \frac{V_{DD}}{V_0} T_{ON} \quad \text{--- (2)}$$

$$I_L(\text{avg}) = I_{L(\text{load})}$$



$$D' = \frac{2L}{R_L T_{sw}} = k$$

$$I_L(\text{avg}) = \frac{1}{2} \times (T_{ON} + T_{OFF1}) \times \frac{V_{DD} - V_0}{L} \times T_{ON} = I_{L(\text{load})}$$

from (2)

$$T_{ON} + T_{OFF1} = \frac{V_{dd}}{V_0} \times T_{ON}$$

$$\left( \frac{1}{T_{sw}} \right) \frac{1}{2} \frac{V_{dd}}{V_0} T_{ON} \times \frac{V_0}{L} \left( \frac{V_{dd}}{V_0} - 1 \right) \times T_{ON} = \frac{V_0}{R_L}$$

$$\frac{1}{T_{sw}} \times \frac{1}{2} \frac{V_{dd}}{V_0} (D \cdot T_{sw}) \left( \frac{V_{dd}}{V_0} - 1 \right) \times D \cdot T_{sw}$$

$$\frac{V_{dd}}{V_0} \left( \frac{V_{dd}}{V_0} - 1 \right) D^2 = \frac{2L}{R_L T_{sw}} = k$$

$$\left( \frac{V_{dcl}}{V_0} \right)^2 - \frac{V_{dcl}}{V_0} - \frac{k}{D^2} = 0$$

$$\frac{V_{dcl}}{V_0} = x$$

$$ax^2 + bx + c$$

$$x = \frac{1 \pm \sqrt{1 + \frac{4k}{D^2}}}{2}$$

$$\frac{V_0}{V_{dcl}} = \frac{2}{1 \pm \sqrt{1 + \frac{4k}{D^2}}}$$

for buck converter,  $\frac{V_0}{V_{dcl}} < 1$

$$\frac{V_0}{V_{dcl}} \neq \frac{2}{1 - \sqrt{1 + \frac{4k}{D^2}}}$$

$$\frac{V_o}{V_{dd}} = \frac{2}{1 + \sqrt{1 + \frac{4k}{D^2}}}$$

$$D' = \frac{2L}{R_{L_{\text{eq}}}} = k$$

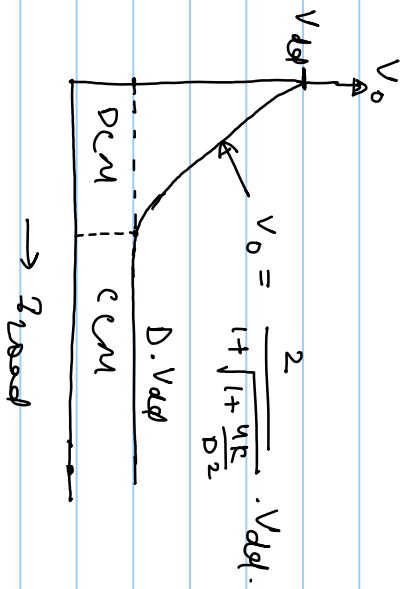
if  $R_L \rightarrow \infty$

$$k = 0$$

$$\frac{V_o}{V_{dd}} = 1 \Rightarrow V_o \approx V_{in}$$

$$I_n \text{ e } C_M; \quad V_o = D \cdot V_{dd}$$

$$\text{In DCML: } V_o = \frac{2}{1 + \sqrt{1 + \frac{4k}{D^2}}}$$



For constant  $V_o$  &  $V_{in}$

$$\frac{V_o}{V_{in}} = \frac{2}{1 + \sqrt{1 + \frac{4K}{D^2}}} = \text{constant}$$

$$\Rightarrow \frac{4K}{D^2} = \text{constant}$$

$$K = \frac{2L}{R_L \cdot T_{sw}}$$

$$\Rightarrow \frac{4}{D^2} \left( \frac{2L}{R_L T_{sw}} \right) = \text{constant}$$



$$\Rightarrow \frac{y}{D^2} \left( \frac{2L}{R_L T_{sw}} \right) = \text{constant}$$

for fixed  $L$  &  $T_{sw}$

$$R_L \uparrow \quad D \downarrow$$

For linear operation, we limit  $T_{on} \rightarrow T_{on\_min}$

$$D = \frac{T_{on}}{T_{sw}} = \frac{T_{on\_min}}{T_{sw}} \quad \text{for high } R_L \text{ (lower } I_{load})$$

$$\frac{y}{T_{on\_min}^2} \times T_{sw}^2 \left( \frac{2L}{R_L T_{sw}} \right) = \text{constant}$$

$$\frac{8L}{T_{on\_min}^2} \left( \frac{T_{sw}}{R_L} \right) = \text{constant}$$

constant

$\frac{T_{sw}}{R_L}$  should be constant in order to regulate the output.

$R_L \uparrow \quad T_{sw} \uparrow$  or  $F_{sw} \downarrow$

$F_{sw} \propto I_{load}$ .

called PWM mode

Pulse Frequency Modulation