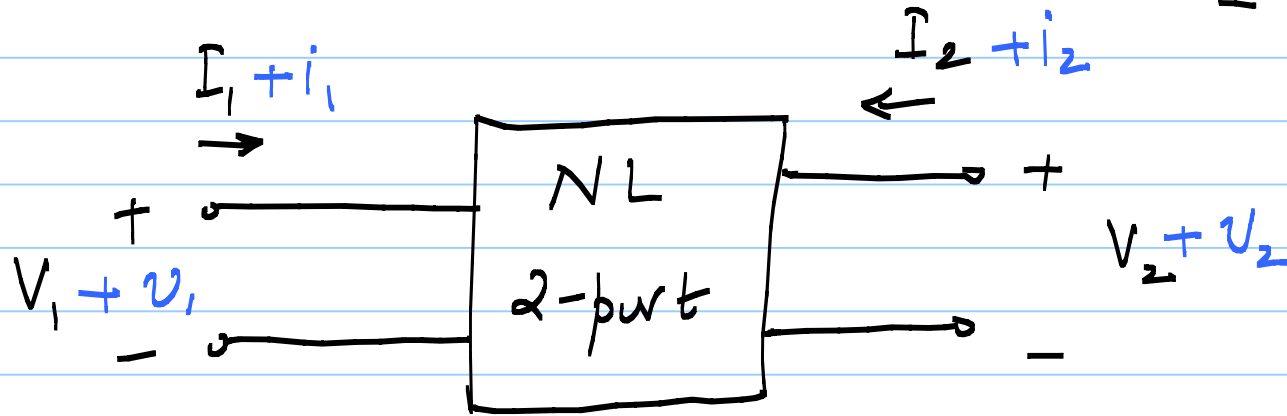


13/8/20

Lecture 6

$$[\mathbf{I}] = [\mathbf{Y}] [\mathbf{V}]$$



$$I_1 = f(V_1, V_2)$$

$$I_2 = g(V_1, V_2)$$

$$I_1' = I_1 + i_1 = f(V_1 + v_1, V_2 + v_2)$$

$$I_1 = f(V_1, V_2)$$

* If v_1 & v_2 are small, $f()$ & $g()$ can be expanded in a 2-D Taylor Series around op. pt. { approx. 3D surface by a tangential plane @ op. pt. }

$$I_1' = I_1 + i_1 \approx I_1 + \frac{\partial f}{\partial V_1} \cdot v_1 + \frac{\partial f}{\partial V_2} \cdot v_2$$

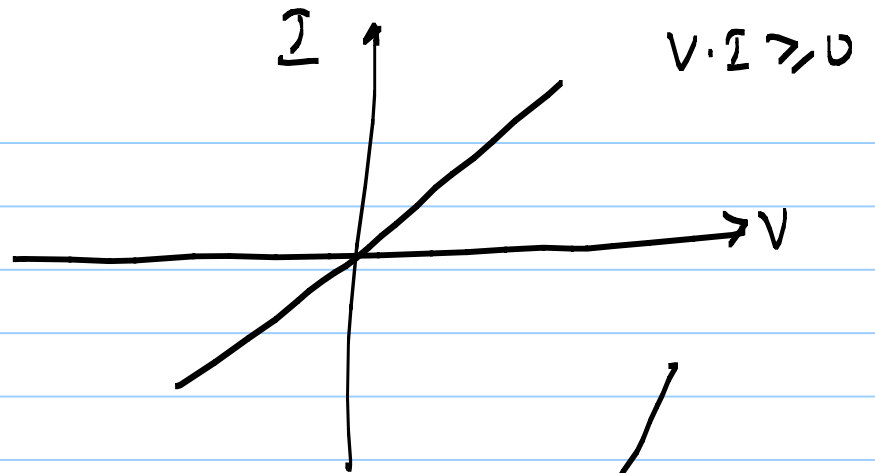
$$\Rightarrow i_1 = \frac{\partial f}{\partial V_1} \cdot v_1 + \frac{\partial f}{\partial V_2} \cdot v_2 \quad \left. \vphantom{i_1} \right\} \text{linear relationship}$$

$$i_2 = \frac{\partial g}{\partial V_1} \cdot v_1 + \frac{\partial g}{\partial V_2} \cdot v_2 \quad \left. \vphantom{i_2} \right\} \text{between } i_1, i_2, v_1, v_2$$

$$\begin{bmatrix} i_1 \\ i_2 \end{bmatrix} = \underbrace{\begin{bmatrix} \overset{y_{11}}{\partial f / \partial V_1} & \overset{y_{12}}{\partial f / \partial V_2} \\ \underset{y_{21}}{\partial g / \partial V_1} & \underset{y_{22}}{\partial g / \partial V_2} \end{bmatrix}}_{\text{incremental } y\text{-matrix}} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix}$$

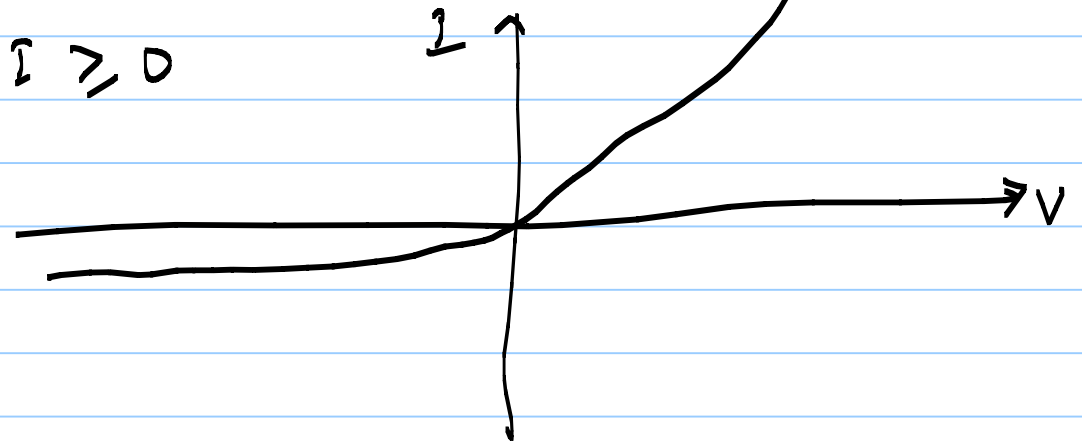
Graphical Representation

1) Linear v 1-port
passive

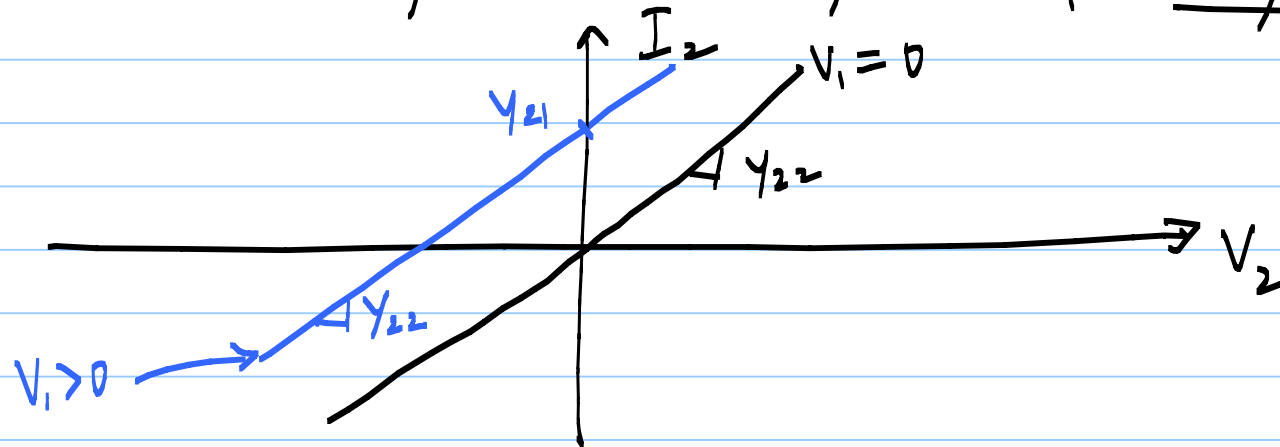


2) NL \wedge 1-port
passive

$$V \cdot I \geq 0$$



3) Linear 2-port : Input & output characteristics



$$I_1 = Y_{11}V_1 + Y_{12}V_2$$

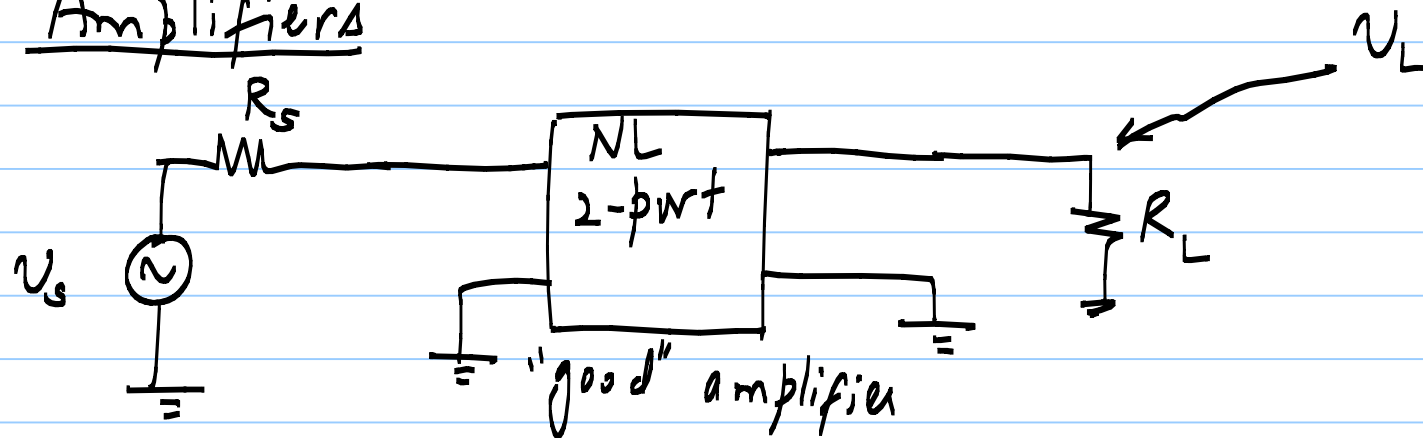
$$I_2 = Y_{21}V_1 + Y_{22}V_2$$

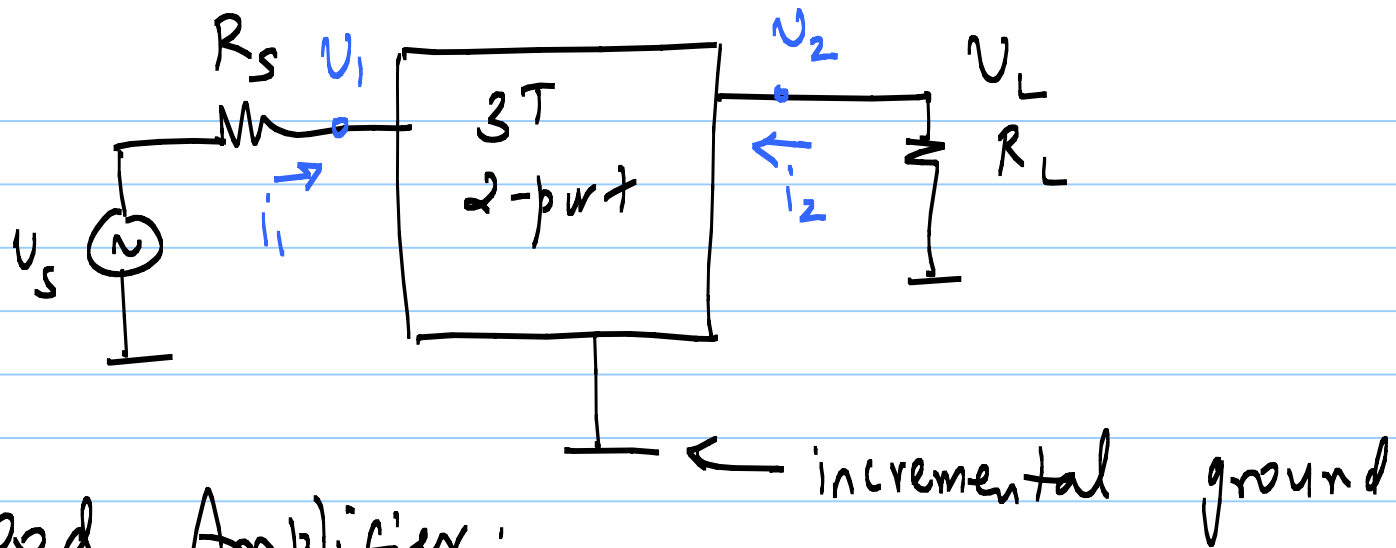
Input characteristics $\Rightarrow I_1$ vs. V_1 for various V_2

passivity : $V_1 I_1 + V_2 I_2 \geq 0$

(Lin. or NonLin)

Amplifiers





Good Amplifier:

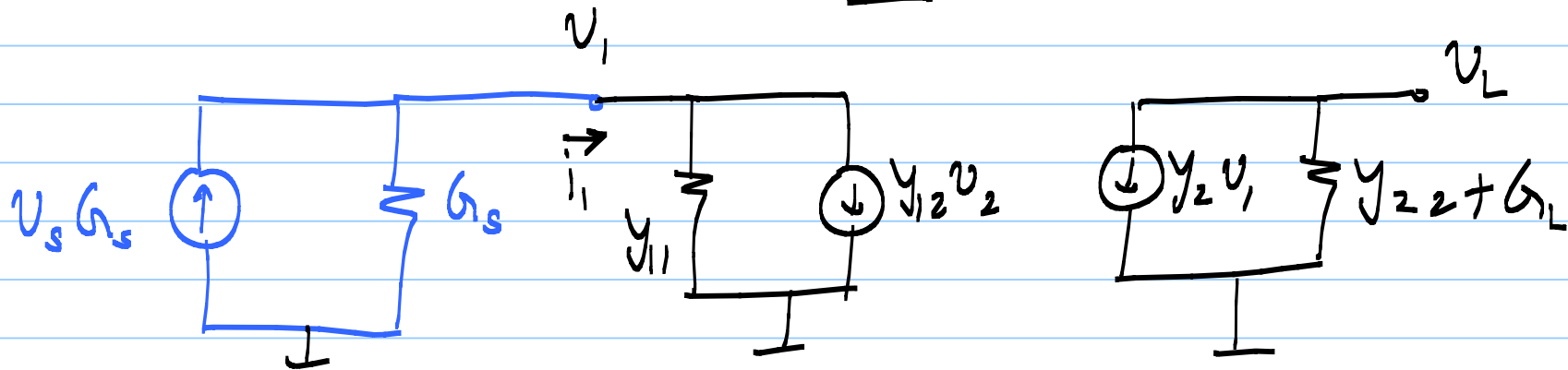
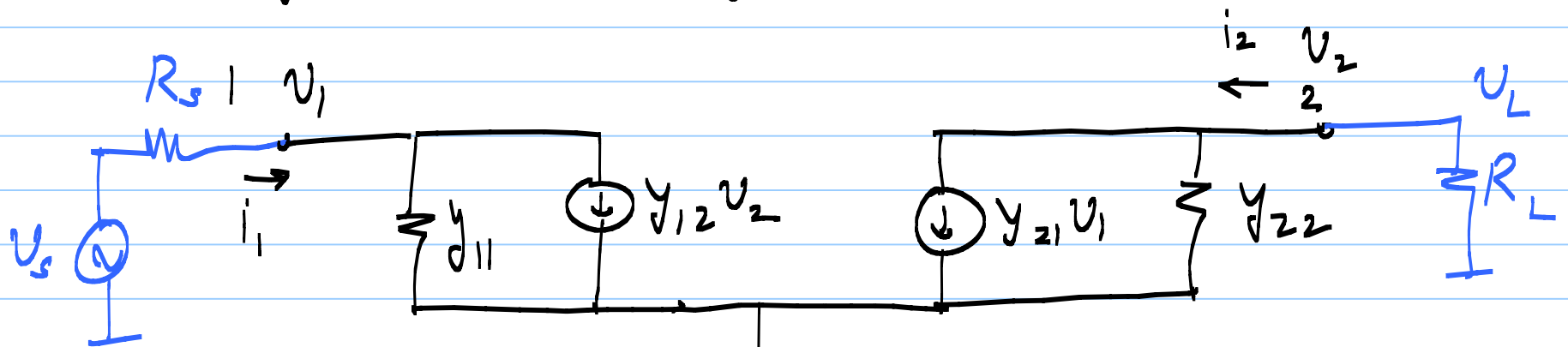
- 1) Large gain : $\frac{v_L}{v_s}$ as large as possible
- 2) Independent of source quality : v_L independent of R_s
and gain independent of R_s
- 3) gain independent of R_L too.
- 4) We want i_1 independent of v_2
"unilateral" $y_{12} = 0$ is desired

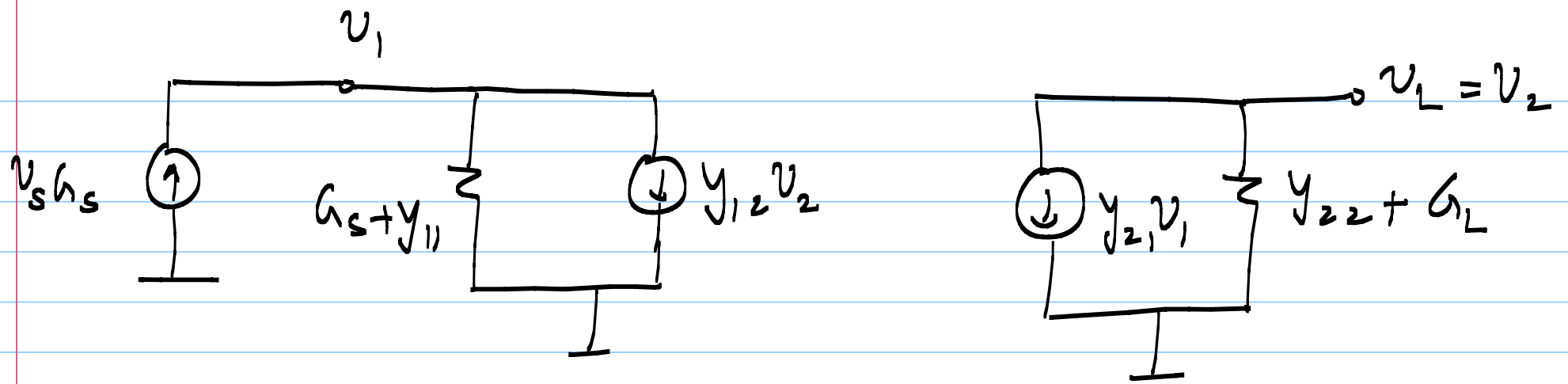
Derive constraints on $[y]$ to achieve a "good" amp.

$$i_1 = y_{11} v_1 + y_{12} v_2$$

$$i_2 = y_{21} v_1 + y_{22} v_2$$

replace by equivalent network





KCL @ input & output.

Ⓐ input: $v_s g_s = v_1 (g_s + y_{11}) + y_{12} v_2$

$$\Rightarrow v_1 = \frac{v_s g_s - y_{12} v_2}{g_s + y_{11}}$$

plug into

Ⓑ output: $y_{21} v_1 + v_2 (y_{22} + g_L) = 0$

$$y_{21} \left[\frac{v_s g_s - y_{12} v_2}{y_{11} + g_s} \right] + (y_{22} + g_L) \cdot v_2 = 0$$

$$v_s \left[\frac{y_{21} \cdot G_s}{y_{11} + G_s} \right] = v_2 \left[\frac{y_{12} y_{21}}{y_{11} + G_s} - (y_{22} + G_L) \right]$$

$$= v_2 \left[\frac{y_{12} y_{21} - (y_{22} + G_L)(y_{11} + G_s)}{y_{11} + G_s} \right]$$

$$\frac{v_2}{v_s} = \frac{v_L}{v_s} = \frac{y_{21} G_s}{y_{12} y_{21} - (y_{22} + G_L)(y_{11} + G_s)}$$

* If $y_{12} y_{21} = (y_{22} + G_L)(y_{11} + G_s)$, gain = ∞
 undesired situation, gain needs to be a
 function of y_{11} etc.

* Make amplifier unilateral: $y_{12} = 0$

$$\Rightarrow \frac{v_L}{v_S} = \frac{-y_{21} h_S}{(y_{22} + h_L)(y_{11} + h_S)}$$