## EE512: Error Control Coding

## Solution for Assignment on Linear Block Codes

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1. Code 1: n = 4, n - k = 2Parity Check Equations:  $x_1 + x_3 = 0$ ,  $x_1 + x_2 + x_4 = 0$ Parity Bits:  $x_3 = x_1$ ,  $x_4 = x_1 + x_2$  $C_1 = \{0000,0101,1011,1110\}$ 

Code 2: n = 4, n - k = 2Parity Check Equations:  $x_2 + x_3 + x_4 = 0$ ,  $x_1 + x_2 + x_4 = 0$ Parity Bits:  $x_3 = x_1, x_4 = x_1 + x_2$ , Hence  $C_2 = C_1$ 

Systematic form: Generator Matrix:

Parity Check Matrix:

$\mathbf{G} =$	$\left[\begin{array}{c}1\\0\end{array}\right]$	$\begin{array}{c} 0 \\ 1 \end{array}$	$\begin{array}{c} 1 \\ 0 \end{array}$	1 1	]
	г.				-

<b>н</b> _[	1	0	1	0
$\mathbf{n} = \begin{bmatrix} \mathbf{n} \\ \mathbf{n} \end{bmatrix}$	. 1	1	0	1

2. After Gaussian Elimination:

C	[1]	0	0	1	0	1	1	0
	0	1	0	1	0	1	0	1
G =	0	0	1	1	0	0	1	1
	0	0	0	0	1	1	1	1

Blocklength, n = 8, Dimension k = 4, Minimum Distance  $d_{min} = 4$  (found by enumerating all codewords).

3. (a)

$$d_H(\mathbf{u}, \mathbf{v}) = \{1 \le i \le n : u_i \ne v_i\}$$
$$= \{1 \le i \le n : u_i + v_i \ne 0\}$$
$$= \operatorname{wt}(\mathbf{u} + \mathbf{v}).$$

- (b)  $\operatorname{wt}(\mathbf{u}) + \operatorname{wt}(\mathbf{v})$  gives the number of ones in  $\mathbf{u}$  and  $\mathbf{v}$  with the common positions counted twice. Hence, subtract  $2\operatorname{wt}(\mathbf{u} * \mathbf{v})$ , which is the number of nonzero positions common to both  $\mathbf{u}$  and  $\mathbf{v}$ , from  $\operatorname{wt}(\mathbf{u}) + \operatorname{wt}(\mathbf{v})$  to get  $d_H(\mathbf{u}, \mathbf{v})$ .
- (c) Follows from (3b).
- (d) Since  $d_H(\mathbf{u}, \mathbf{v}) = \sum_{i=1}^n d(u_i, v_i)$ , it is enough to prove

$$d(u_i, v_i) \le d(u_i, w_i) + d(w_i, v_i).$$
(1)

If  $u_i = v_i = w_i$ , then (1) evaluates to  $0 \le 0$ . If  $u_i \ne v_i$  and  $u_i = w_i$ ,  $v_i \ne w_i$ , 1 = 1. If  $u_i \ne v_i$ and  $u_i \ne w_i$ ,  $v_i = w_i$ , (1) evaluates to  $1 \le 1$ . If  $u_i \ne v_i$  and  $u_i \ne w_i$ ,  $v_i \ne w_i$ , (1) evaluates to  $1 \le 2$ . In all cases, inequality is satisified. Hence,  $d_H(\mathbf{u}, \mathbf{v}) \le d_H(\mathbf{u}, \mathbf{w}) + d_H(\mathbf{w}, \mathbf{v})$ .

- (e) From (3d), we know that  $wt(\mathbf{u} + \mathbf{w}) \leq wt(\mathbf{u} + \mathbf{v}) + wt(\mathbf{v} + \mathbf{w})$ . Let  $\mathbf{w} = \mathbf{0}$ . It follows  $wt(\mathbf{u}) \leq wt(\mathbf{u} + \mathbf{v}) + wt(\mathbf{v})$
- 4. (a) (n, 8, 4) Code

By Hamming bound,  $n \ge 12$ . By Singleton bound,  $n \ge 11$ . GV Bound evaluates to  $n \ge 15$ . For n = 11 or n = 12, we cannot construct a (n - 1, 8, 3) code with minimum distance 3, since we do not have n - 1 distinct columns. A (13, 8, 4) code is constructed by extending a (12, 8, 3) code. One sample parity-check matrix is shown below.

(b) (n, 16, 4) Code

By Hamming bound,  $n \ge 21$ . By Singleton bound,  $n \ge 19$ . GV Bound evaluates to  $n \ge 25$ . For n = 19 or , n = 20, we cannot construct a (n - 1, 16, 3) code with minimum distance 3, since we do not have n - 1 distinct columns. A (22, 16, 4) code is constructed by extending a (21, 16, 3) code.

- (c) (n, 32, 4) Code By Hamming bound,  $n \ge 38$ . By Singleton bound,  $n \ge 33$ . GV bound evaluates to  $n \ge 42$ . A (39, 32, 4) code is constructed by extending a (38, 32, 3) code.
- 5. (a) Dual of the (n, 1, n) repetition code is the (n, n 1, 2) even weight codeword.
  - (b) Dual of the (n, n 1, 2) even weight codeword is the (n, 1, n) repetition code.
- 6.  $\mathbf{G} = \mathbf{I}_k$ . Hence  $C = \{0, 1\}^k$ .
- 7. (a) Block Length  $n = n_1 + n_2$ , Dimension k. Any codeword **x** can be written as  $\mathbf{x} = [\mathbf{x}_1 \mathbf{x}_2]$ , where  $\mathbf{x}_1 = uG_1$  and  $\mathbf{x}_2 = uG_2$ . If  $u \neq \mathbf{0}$ ,  $wt(\mathbf{x}) \ge wt(\mathbf{x}_1) + wt(\mathbf{x}_2)$ . Therefore, minimum distance  $d \ge d_1 + d_2$ .
  - (b) Block Length  $n = n_1 + n_2$ , Dimension  $k = k_1 + k_2$ , Minimum distance  $d = \min(d_1, d_2)$ . Let  $\mathbf{x}_1, \mathbf{x}_2$  be codewords generated by  $\mathbf{G}_1, \mathbf{G}_2$ . The vectors  $[\mathbf{x}_1\mathbf{0}]$ , and  $[\mathbf{0}\mathbf{x}_2]$  are codewords of the new code. Hence, minimum distance  $= \min(d_1, d_2)$ .
- 8. Let  $C = C_e \cup C_o$ , where  $C_e$  and  $C_o$  are the set of even and odd weight codewords in C respectively. Since  $\mathbf{0} \in C_e$  and sum of even weight codewords is an even weight codeword,  $C_e$  is a linear code.

Let  $\mathbf{c}_1 \in C_o$ . Consider the coset of  $C_e$  generated by  $\mathbf{c}_1$  denoted as  $\mathbf{c}_1 + C_e$ . For  $\mathbf{c}_2 \in C_o$ , we see that  $\mathbf{c}_1 + \mathbf{c}_2 \in C_e$ , which implies that  $\mathbf{c}_2 \in \mathbf{c}_1 + C_e$  and  $C_o \subseteq \mathbf{c}_1 + C_e$ . Also, for  $\mathbf{c}_3 \in C_e$ ,  $\mathbf{c}_1 + \mathbf{c}_3 \in C_o$ , which implies that  $\mathbf{c}_1 + C_e \subseteq C_o$ . Thus,  $C_e + \mathbf{c}_1 = C_o$ . Since,  $|C_e| = |C_e + \mathbf{c}_1|, |C_e| = |C_o|$ .

- 9. Let  $C_0$  and  $C_1$  be the set of codewords with 0 and 1 in any particular column.  $C_0$  is a subspace, since it includes **0** and is closed. Let  $\mathbf{c}_1, \mathbf{c}_2 \in C_1$ . Consider the coset of  $C_0$ , generated by  $\mathbf{c}_1$ . We know that  $\mathbf{c}_1 + \mathbf{c}_2 \in C_0$ . Hence any vector  $\mathbf{c}_2 \in C_1$ , also belongs to  $C_0 + \mathbf{c}_1$ . Thus,  $C_0 + \mathbf{c}_1 = C_1$ . Since,  $|C_0| = |C_0 + \mathbf{c}_1|, |C_0| + |C_1| = 2^k$ . Therefore,  $|C_0| = |C_1| = 2^k 1$ ).
- 10. The parity-check matrix for the code is

$$\mathbf{H} = \left[ \begin{array}{cc} \mathbf{H}_1 & \mathbf{0} \\ \mathbf{1} & 1 \end{array} \right],$$

where  $\mathbf{H}_1$  is the parity-check matrix of the (7,4) Hamming code. Note that  $\mathbf{H}$  is a parity-check matrix for the (8,4) extended Hamming code. The syndrome is given by  $\mathbf{s} = \mathbf{r}\mathbf{H}^T$  for a received

$\mathbf{s}$ Syndrome	$\mathbf{e}$ Error Pattern
0000	00000000
0001	00000001
1111	00000010
0111	00000100
1011	00001000
1101	00010000
1001	00100000
0101	01000000
0011	1000000
1110	00000011
1000	00000110
1100	00001100
0110	00011000
0100	00110000
0010	10000001
1010	00010100

Table 1: Syndrome table.

vector  $\mathbf{r}$ . The code has a total of 16 cosets represented by the 16 syndromes. Table 1 presents a syndrome table for the maximum likelihood decoder over a BSC.

The probability of codeword error is given by

 $\Pr(\text{Codeword Error}) = 1 - \Pr(\mathbf{e} = \text{Correctable Error Vector}) = 1 - 8p(1-p)^7 - 7p^2(1-p)^6 - (1-p)^8.$ 

11. Message Bit Error: In general, finding accurate probability of bit error involves more drudgery than probability of block error. Typically, simple bounds such as

$$\frac{\Pr(\text{block error})}{\text{message length}} \le \Pr(\text{bit error}) \le \Pr(\text{block error})$$

are sufficient in practice. We will illustrate the entire calculation for this problem.

The list of codewords is seen to be  $C = \{000000, 101001, 111010, 010011, 110100, 011101, 001110, 100111\}$ . The sydrome table is given in Table 2.

s Syndrome	e Error Pattern
000	000000
001	001000
010	010000
011	100010
100	100000
101	000001
110	000100
111	000010

Table 2: Syndrome Table.

The syndrome table is not sufficient for our calculation. Each coset needs to be explicitly determined in the form of a standard array as shown in Table 3. Notice that the first row of the standard array is the code itself, and the first column is the list of coset leaders.

Let us assume that the all-zero codeword [000000] is transmitted and the last three bits of a codeword are set to be the message bits. We will compute the average number of message-bit

Syndrome	Words in coset
000	$\{000000, 101001, 111010, 010011, 110100, 011101, 001110, 100111\}$
001	$\{001000, 100001, 110010, 011011, 111100, 010101, 000110, 101111\}$
010	$\{010000, 111001, 101010, 000011, 100100, 001101, 011110, 110111\}$
011	$\{100010, 001011, 011000, 110001, 010110, 111111, 101100, 000101\}$
100	$\{100000, 001001, 011010, 110011, 010100, 111101, 101110, 000111\}$
101	$\{000001, 101000, 111011, 010010, 110101, 011100, 001111, 100110\}$
110	$\{000100, 101101, 111110, 010111, 110000, 011001, 001010, 100011\}$
111	$\{000010, 101011, 111000, 010001, 110110, 011111, 001100, 100101\}$

Table 3: Standard array.

errors in one decoded codeword. More precisely, if X denotes the number of message-bit errors in a decoded codeword, we will compute the PMF of X given that [000000] is the transmitted codeword. Table 4 shows the value of X corresponding to each possibility for the decoded codeword.

Decoded codeword	Decoded message vector	X
000000	000	0
101001	001	1
111010	010	1
010011	011	2
110100	100	1
011101	101	2
001110	110	2
100111	111	3

Table 4: Message-bit error table

From Table 4 and Table 3, we see the following:

$\Pr(X=0)$	=	$\Pr(\mathbf{e} \text{ is in 1st column of array}),$
$\Pr(X=1)$	=	Pr(e  is in 2nd, 3rd, or 5th column of array),
$\Pr(X=2)$	=	$\Pr(\mathbf{e} \text{ is in 4th, 6th or 7th column of array}),$
$\Pr(X=3)$	=	$\Pr(\mathbf{e} \text{ is in 8th column of array}).$

Now,

$$\begin{split} E[X] &= & \Pr(X=1) + 2\Pr(X=2) + 3\Pr(X=3), \\ &= & (7p^2(1-p)^4 + 8p^3(1-p)^3 + 7p^4(1-p)^2 + 2p^5(1-p)) \\ &+ 2(6p^2(1-p)^4 + 8p^3(1-p)^3 + 7p^4(1-p)^2 + 2p^5(1-p) + p^6) \\ &+ 3(p^2(1-p)^4 + 4p^3(1-p)^3 + p^4(1-p)^2 + 2p^5(1-p)), \\ &= & 22p^2(1-p)^4 + 36p^3(1-p)^3 + 24p^4(1-p)^2 + 12p^5(1-p) + 2p^6. \end{split}$$

Assuming the transmission of each codeword is independent, we can see that the bit-error rate (or expected fraction of message bits in error) will evaluate to E[X]/k = E[X]/3.

We could also compute other probabilities using the standard array and the syndrome table. For

instance,

$$\begin{array}{lll} \Pr(\text{first message bit is in error}) &=& \Pr(\mathbf{e} \text{ is in 5th, 6th, 7th or 8th column of array}), \\ &=& 7p^2(1-p)^4 + 12p^3(1-p)^3 + 8p^4(1-p)^2 + 4p^5(1-p) + p^6. \\ \Pr(\text{second message bit is in error}) &=& \Pr(\mathbf{e} \text{ is in 3rd, 4th, 7th or 8th column of array}), \\ &=& 8p^2(1-p)^4 + 12p^3(1-p)^3 + 8p^4(1-p)^2 + 4p^5(1-p). \\ \Pr(\text{third message bit is in error}) &=& \Pr(\mathbf{e} \text{ is in 2nd, 4th, 6th or 8th column of array}), \\ &=& 7p^2(1-p)^4 + 12p^3(1-p)^3 + 8p^4(1-p)^2 + 4p^5(1-p). \end{array}$$

Notice that the average of the above three probabilities also works out to be equal to E[X]/3. Are these two averages always equal? How about for linear codes? What about the conditioning on the all-zero codeword?

12. (a) Parity Check Matrix

(b) The parity bits are found using the following equations so that the codeword satisfies  $\mathbf{H}[\mathbf{p} \ \mathbf{m}]^T = \mathbf{0}$ .

 $\begin{array}{rcl} x_{12} & = & x_5 + x_6 + x_7 + x_8 + x_9 + x_{10} + x_{11} \\ x_{13} & = & x_2 + x_3 + x_4 + x_8 + x_9 + x_{10} + x_{11} \\ x_{14} & = & x_1 + x_3 + x_4 + x_6 + x_7 + x_{10} + x_{11} \\ x_{15} & = & x_1 + x_2 + x_4 + x_5 + x_7 + x_9 + x_{11} \end{array}$ 

Using the given message, we get the codeword to be [111111000000100].

(c) Syndrome

 $\mathbf{s} = \mathbf{H}\mathbf{r}^T = [1010]^T$ . The syndrome equals the tenth column of **H** implying that the tenth bit is in error. Therefore, decoded message = [1110001111].

13. To show that  $C \cup (\mathbf{u} + C)$  is closed, let  $\mathbf{a}, \mathbf{b} \in C \cup (\mathbf{u} + C)$ . We consider the following cases:

case 1:  $\mathbf{a}, \mathbf{b} \in C$  implies  $\mathbf{a} + \mathbf{b} \in C$ .

case 2:  $\mathbf{a}, \mathbf{b} \in \mathbf{u} + C$  implies  $\mathbf{a} + \mathbf{b} \in C$ .

case 3:  $\mathbf{a} \in C$ ,  $\mathbf{b} \in \mathbf{u} + C$  implies  $\mathbf{a} + \mathbf{b} \in \mathbf{u} + C$ .

Hence,  $C \cup (\mathbf{u} + C)$  is a linear code.

- 14. (a) Note that a vector is in C iff it is orthogonal to all vectors in C<sup>⊥</sup>. Similarly, a vector is in C<sup>⊥</sup> iff it is orthogonal to all vectors in C.
  We need to show that (C<sup>⊥</sup>)<sup>⊥</sup> = C. Consider x ∈ C. By definition x · y = 0 for all y ∈ C<sup>⊥</sup>; hence, we conclude that x ∈ (C<sup>⊥</sup>)<sup>⊥</sup>. So, C ⊆ (C<sup>⊥</sup>)<sup>⊥</sup>.
  Now consider u ∈ (C<sup>⊥</sup>)<sup>⊥</sup>. By definition u · v = 0 for all v ∈ C<sup>⊥</sup>; hence, we conclude that u ∈ C. So, (C<sup>⊥</sup>)<sup>⊥</sup> ⊆ C.
  Hence proved.
  - (b) Similar to above part.
- 15. (a) Converting  $\mathbf{G}$  into systematic form, we get the following:

$$\mathbf{G}_s = \left[ \begin{array}{rrrr} 1 & 1 & 0 & 0 \\ 0 & 1 & 1 & 0 \\ 0 & 0 & 1 & 1 \end{array} \right]$$

Since  $\mathbf{G}_s = \mathbf{G}'$ , the two codes are identical.

(b) In systematic form, the matrix  $\mathbf{G}'$  becomes

$$\mathbf{G'}_s = \left[ \begin{array}{rrrrr} 1 & 0 & 0 & 1 & 0 & 0 \\ 0 & 1 & 0 & 0 & 1 & 0 \\ 0 & 0 & 1 & 0 & 0 & 1 \end{array} \right]$$

After a column permutation, we get

$$\mathbf{G'}_{eq} = \left[ \begin{array}{rrrrr} 1 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 1 \end{array} \right]$$

Since  $\mathbf{G} = \mathbf{G'}_{eq}$ , the two codes are equivalent.

16. Let the parity-check matrix be

$$\mathbf{H} = \begin{bmatrix} 1 & a & b & 1 & 0 & 0 & 0 \\ 0 & c & d & 0 & 1 & 0 & 0 \\ 1 & e & f & 0 & 0 & 1 & 0 \\ 1 & g & h & 0 & 0 & 0 & 1 \end{bmatrix},$$

where the variables a through g are bits that need to be determined. Since  $\mathbf{H}[0110011]^T = \mathbf{0}$ , we get the following equations:

$$\begin{array}{lll} a+b=0 & \Longrightarrow & a=b \\ c+d=0 & \Longrightarrow & c=d \\ e+f=1 & \Longrightarrow & e\neq f \\ g+h=1 & \Longrightarrow & g\neq h \end{array}$$

For a minimum distance of 4, we need to have atleast 3 linearly independent columns. This implies atleast 3 ones must be present in each of the columns. Therefore the parity check matrix is

$$\mathbf{H} = \begin{bmatrix} 1 & 1 & 1 & 1 & 0 & 0 & 0 \\ 0 & 1 & 1 & 0 & 1 & 0 & 0 \\ 1 & 0 & 1 & 0 & 0 & 1 & 0 \\ 1 & 1 & 0 & 0 & 0 & 0 & 1 \end{bmatrix}$$

Note that we still need to check that the minimum distance is actually 4!

17. (a) Notice that the generator matrix can be written as  $G = [G_1 G_2]$ , where

$$G_1 = \begin{bmatrix} 1 & 0 & 0 & 0 & 1 \\ 0 & 1 & 0 & 0 & 1 \\ 0 & 0 & 1 & 0 & 1 \\ 0 & 0 & 0 & 1 & 1 \end{bmatrix}$$

is a generator matrix of the (5,4,2) even-weight code (why? find its parity-check matrix in systematic form) and  $G_2$  is a generator matrix of the (7,4,3) Hamming code (why? find parity-check matrix). By Problem (7a), we see that  $d \ge 2+3=5$ . Since the first row of G is a codeword of weight 5, we see that d = 5.

- Alternatively, by enumerating all 16 codewords, we find  $d_{min} = 5$ .
- (b) By careful inspection, we can quickly see that the codeword closest to [11111111111] will be [111101111111]. Alternatively, we need to calculate the syndrome and find a suitable coset leader!
- 18. (a) Linear: Impossible

For  $\mathbf{a} \in C$ ,  $\mathbf{a} + \mathbf{a} = 0$  has even weight and does not belong to C. Hence, the code is nonlinear.

- (b) Minimum Distance is 5: Impossible Let  $\mathbf{u}$ ,  $\mathbf{v}$  belong to C.  $d_H(\mathbf{u}, \mathbf{v}) = \operatorname{wt}(\mathbf{u}) + \operatorname{wt}(\mathbf{v}) - 2\operatorname{wt}(\mathbf{u}\mathbf{v}) =$  some even number. Hence minimum distance cannot be an odd number.
- (c) C is self-orthogonal: Impossible For a self-orthogonal code, any codeword is orthogonal to itself. For  $\mathbf{a} \in C$ ,  $\mathbf{a} \cdot \mathbf{a} = \text{wt}(\mathbf{a})$ mod 2 = 1. Hence, C cannot be self-orthogonal.