Solutions to Problem Set 5

EE419: Digital Communication Systems

Check the solutions for possible bugs!

- 1. No solution provided.
- 2. No solution provided.
- 3. (a) Assuming symbols are equally likely a priori, the probability of symbol error is given by

$$P_e = \frac{6}{8}(2Q(1/\sigma)) + \frac{2}{8}(Q(1/\sigma)) = \frac{7}{4}Q(1/\sigma).$$

(b) One possible Gray labeling is $b_0b_1b_2 = 000,001,010,011,111,110,100,101$ from left to right. The LLR for b_0 (assuming bits are *iid* uniform *a priori*) is given by

LLR(b₀) = log
$$\frac{f(r|b_0 = 0)}{f(r|b_0 = 1)}$$

= log $\frac{e^{-(r+7)^2/2\sigma^2} + e^{-(r+5)^2/2\sigma^2} + e^{-(r+3)^2/2\sigma^2} + e^{-(r+1)^2/2\sigma^2}}{e^{-(r-7)^2/2\sigma^2} + e^{-(r-5)^2/2\sigma^2} + e^{-(r-3)^2/2\sigma^2} + e^{-(r-1)^2/2\sigma^2}}.$

For b_1 ,

LLR(
$$b_1$$
) = log $\frac{f(r|b_1 = 0)}{f(r|b_1 = 1)}$
= log $\frac{e^{-(r+7)^2/2\sigma^2} + e^{-(r+5)^2/2\sigma^2} + e^{-(r-5)^2/2\sigma^2} + e^{-(r-7)^2/2\sigma^2}}{e^{-(r+3)^2/2\sigma^2} + e^{-(r+1)^2/2\sigma^2} + e^{-(r-1)^2/2\sigma^2} + e^{-(r-3)^2/2\sigma^2}}.$

For b_2 ,

LLR(b₂) = log
$$\frac{f(r|b_2 = 0)}{f(r|b_2 = 1)}$$

= log $\frac{e^{-(r+7)^2/2\sigma^2} + e^{-(r+3)^2/2\sigma^2} + e^{-(r-3)^2/2\sigma^2} + e^{-(r-5)^2/2\sigma^2}}{e^{-(r+5)^2/2\sigma^2} + e^{-(r+1)^2/2\sigma^2} + e^{-(r-1)^2/2\sigma^2} + e^{-(r-7)^2/2\sigma^2}}$

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Probability of error for b_0 is given by

$$\Pr\{\text{Error in } b_0\} = \frac{1}{4}(Q(1/\sigma) + Q(3/\sigma) + Q(5/\sigma) + Q(7/\sigma)).$$

Probability of error for b_1 is given by

$$\begin{aligned} \Pr\{\text{Error in } b_1\} &= \frac{1}{4}Q(1/\sigma) + \frac{1}{4}Q(3/\sigma) + \frac{1}{4}(Q(1/\sigma) + Q(7/\sigma)) + \frac{1}{4}(Q(3/\sigma) + Q(5/\sigma)) \\ &= 0.5(Q(1/\sigma) + Q(3/\sigma)) + 0.25(Q(5/\sigma) + Q(7/\sigma)). \end{aligned}$$

Probability of error for b_2 is given by

$$\begin{split} &\frac{1}{8}(Q(1/\sigma) - Q(3/\sigma) + Q(5/\sigma) - Q(9/\sigma) + Q(13/\sigma)) + \frac{1}{8}(Q(1/\sigma) + Q(1/\sigma) - Q(3/\sigma) + Q(7/\sigma) - Q(11/\sigma)) + \\ &\frac{1}{8}(Q(1/\sigma) - Q(3/\sigma) + Q(1/\sigma) - Q(5/\sigma) + Q(9/\sigma)) + \frac{1}{8}(Q(3/\sigma) - Q(5/\sigma) + Q(3/\sigma) - Q(7/\sigma)) + \\ &\frac{1}{8}(Q(1/\sigma) - Q(5/\sigma) + Q(3/\sigma) - Q(5/\sigma) + Q(7/\sigma)) + \frac{1}{8}(Q(3/\sigma) + Q(1/\sigma) - Q(5/\sigma) + Q(7/\sigma) - Q(9/\sigma)) + \\ &\frac{1}{8}(Q(1/\sigma) + Q(3/\sigma) - Q(7/\sigma) + Q(9/\sigma) - Q(11/\sigma)) + \frac{1}{8}(Q(1/\sigma) - Q(5/\sigma) + Q(9/\sigma) - Q(11/\sigma) + Q(1/\sigma) - Q(11/\sigma) + Q(1/\sigma)) + \\ &\frac{1}{8}Q(1/\sigma) + \frac{1}{4}Q(3/\sigma) - \frac{5}{8}Q(5/\sigma) + \frac{1}{8}Q(7/\sigma) + \frac{1}{8}Q(9/\sigma) - \frac{3}{8}Q(11/\sigma) + \frac{1}{4}Q(13/\sigma). \end{split}$$

(c) Binary labeling $b_0b_1b_2 = 000,001,010,011,100,101,110,111$ from left to right. The LLR for b_0 (assuming bits are *iid* uniform *a priori*) is given by

$$LLR(b_0) = \log \frac{f(r|b_0 = 0)}{f(r|b_0 = 1)}$$

= $\log \frac{e^{-(r+7)^2/2\sigma^2} + e^{-(r+5)^2/2\sigma^2} + e^{-(r+3)^2/2\sigma^2} + e^{-(r+1)^2/2\sigma^2}}{e^{-(r-7)^2/2\sigma^2} + e^{-(r-5)^2/2\sigma^2} + e^{-(r-3)^2/2\sigma^2} + e^{-(r-1)^2/2\sigma^2}}.$

For b_1 ,

LLR(b₁) = log
$$\frac{f(r|b_1 = 0)}{f(r|b_1 = 1)}$$

= log $\frac{e^{-(r+7)^2/2\sigma^2} + e^{-(r+5)^2/2\sigma^2} + e^{-(r-1)^2/2\sigma^2} + e^{-(r-3)^2/2\sigma^2}}{e^{-(r+3)^2/2\sigma^2} + e^{-(r+1)^2/2\sigma^2} + e^{-(r-5)^2/2\sigma^2} + e^{-(r-7)^2/2\sigma^2}}$.

For b_2 ,

LLR(b₂) = log
$$\frac{f(r|b_2 = 0)}{f(r|b_2 = 1)}$$

= log $\frac{e^{-(r+7)^2/2\sigma^2} + e^{-(r+3)^2/2\sigma^2} + e^{-(r-1)^2/2\sigma^2} + e^{-(r-5)^2/2\sigma^2}}{e^{-(r+5)^2/2\sigma^2} + e^{-(r+1)^2/2\sigma^2} + e^{-(r-3)^2/2\sigma^2} + e^{-(r-7)^2/2\sigma^2}}.$

Probability of error for b_0 is given by

$$\Pr\{\text{Error in } b_0\} = \frac{1}{4}(Q(1/\sigma) + Q(3/\sigma) + Q(5/\sigma) + Q(7/\sigma)).$$

Probability of error for b_1 is given by

$$\Pr\{\text{Error in } b_1\} = \frac{1}{4}(Q(3/\sigma) - Q(7/\sigma) + Q(11/\sigma)) + \frac{1}{4}(Q(1/\sigma) - Q(5/\sigma) + Q(9/\sigma)) + \frac{1}{4}(Q(1/\sigma) + Q(3/\sigma) - Q(7/\sigma)) + \frac{1}{4}(Q(3/\sigma) + Q(1/\sigma) - Q(5/\sigma)) \\ = \frac{3}{4}(Q(1/\sigma) + Q(3/\sigma)) - \frac{1}{2}(Q(5/\sigma) + Q(7/\sigma)) + \frac{1}{4}(Q(9/\sigma) + Q(11/\sigma)).$$

Probability of error for b_2 is given by

$$\begin{split} &\frac{1}{4}(Q(1/\sigma) - Q(3/\sigma) + Q(5/\sigma) - Q(7/\sigma) + Q(9/\sigma) - Q(11/\sigma) + Q(13/\sigma)) + \\ &\frac{1}{4}(Q(1/\sigma) + Q(1/\sigma) - Q(3/\sigma) + Q(5/\sigma) - Q(7/\sigma) + Q(9/\sigma) - Q(11/\sigma)) + \\ &\frac{1}{4}(Q(1/\sigma) - Q(3/\sigma) + Q(1/\sigma) - Q(3/\sigma) + Q(5/\sigma) - Q(7/\sigma) + Q(9/\sigma)) + \\ &\frac{1}{4}(Q(1/\sigma) - Q(3/\sigma) + Q(5/\sigma) - Q(7/\sigma) + Q(1/\sigma) - Q(3/\sigma) + Q(5/\sigma) - Q(7/\sigma)) \\ &= \frac{7}{4}Q(1/\sigma) - \frac{3}{2}Q(3/\sigma) + \frac{5}{4}Q(5/\sigma) - \frac{5}{4}Q(7/\sigma) + \frac{3}{4}Q(9/\sigma) - \frac{1}{2}Q(11/\sigma) + \frac{1}{4}Q(13/\sigma). \end{split}$$

- (d) Compare the $Q(1/\sigma)$ terms in each expression.
- 4. Similar to the previous problem.
- 5. (a) When A is a known constant, we have BPSK transmission. This has been discussed extensively in class.
 - (b) When $A \in \{\pm 1\}$ is a discrete random variable (independent of X and N) with $p = \Pr\{A = 1\}$, the received signal constellation is 1D with two points $\{-1, 1\}$. To find the decision threshold y, the condition is that

$$f(y|X = -1) = f(y|X = 1)$$

$$pe^{-(y+1)^2/2\sigma^2} + (1-p)e^{-(y-1)^2/2\sigma^2} = pe^{-(y-1)^2/2\sigma^2} + (1-p)e^{-(y+1)^2/2\sigma^2}$$

$$(1-2p)e^{-(y-1)^2/2\sigma^2} = (1-2p)e^{-(y+1)^2/2\sigma^2}$$

$$y = 0.$$

- (c) When A is Rayleigh, the received signal constellation is the entire x-axis. In this case, the decision threshold will work out to be 0 as well.
- 6. (a) The BPSK case has been done in class in detail.
 - (b) The LLR in this case will work out to

$$\log \frac{p e^{-(y+1)^2/2\sigma^2} + (1-p) e^{-(y-1)^2/2\sigma^2}}{p e^{-(y-1)^2/2\sigma^2} + (1-p) e^{-(y+1)^2/2\sigma^2}} = \log \frac{p + (1-p) e^{2y/\sigma^2}}{1-p + p e^{2y/\sigma^2}}.$$

- (c) No solution provided yet.
- 7. Here is a simple method. Since Y = AX + N and $X \in \{\pm 1\}$, $Y^2 = A^2 + 2AXN + N^2$. Taking expected values, $E[Y^2] = A^2 + \sigma^2$. Hence, a simple estimate for A is $\sqrt{E[Y^2] \sigma^2}$ assuming A is positive. A MSE estimate can be found by using ideas from constrained complexity (1-tap) equalization.