

Lecture 35

Note Title

4/10/2008

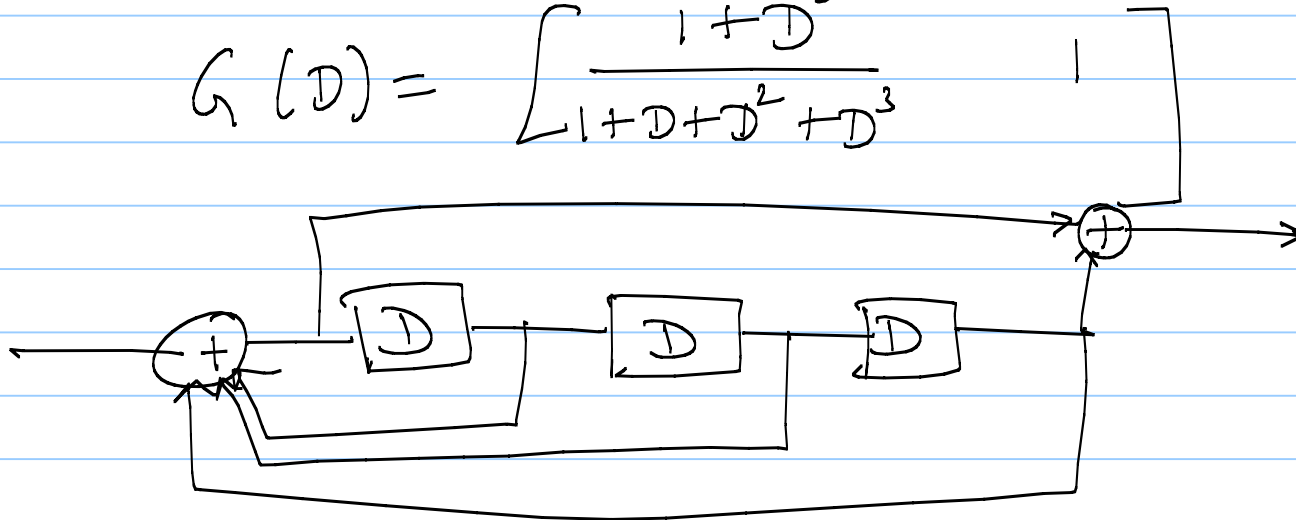
7. (F) Consider a convolutional code with the generator matrix

$$G(D) = [1 + D^3 \quad 1 + D + D^2 + D^3].$$

- (a) Draw the circuit for a non-systematic encoder for the code.
- (b) Draw the circuit for a systematic encoder for the code.

(b)

$$G(D) = \begin{bmatrix} 1 + D^3 & 1 \\ 1 + D + D^2 + D^3 & 1 \end{bmatrix}$$



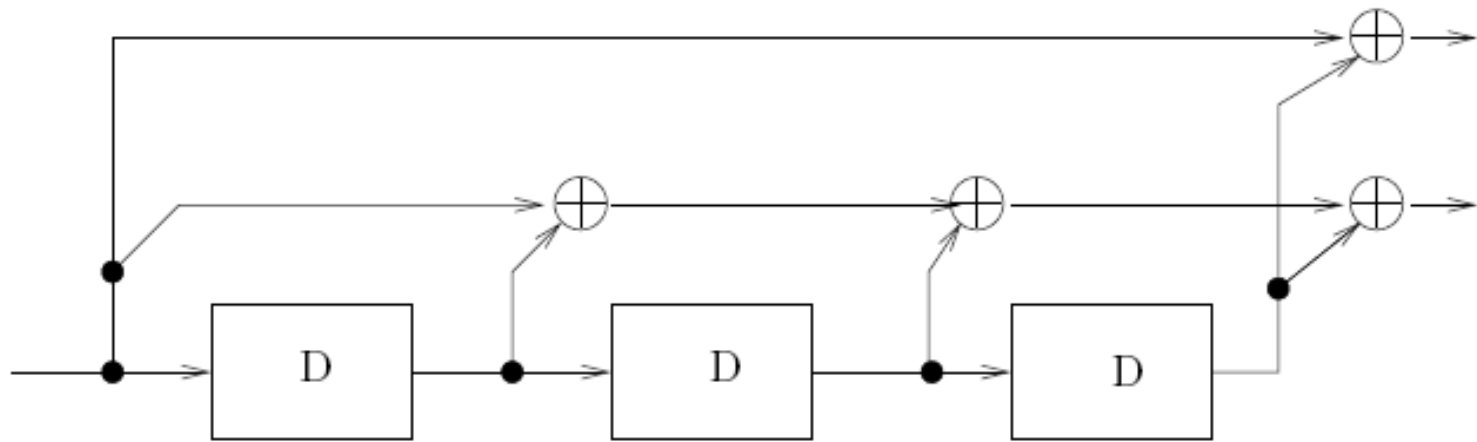


Figure 1: Encoder for Problem 8

- (c) Draw one stage of the trellis for the systematic encoder.
- 8. (F) Consider the convolutional encoder shown in Fig. 1.
 - (a) Draw one stage of the trellis for the encoder.
 - (b) Write down the transfer function matrix for the convolutional code.
 - (c) Encode the infinite message sequence (11111111.....) (all 1s).

(b) $G(D) = \begin{bmatrix} 1+D^3 & 1+D+D^2+D^3 \end{bmatrix}$
 (c)

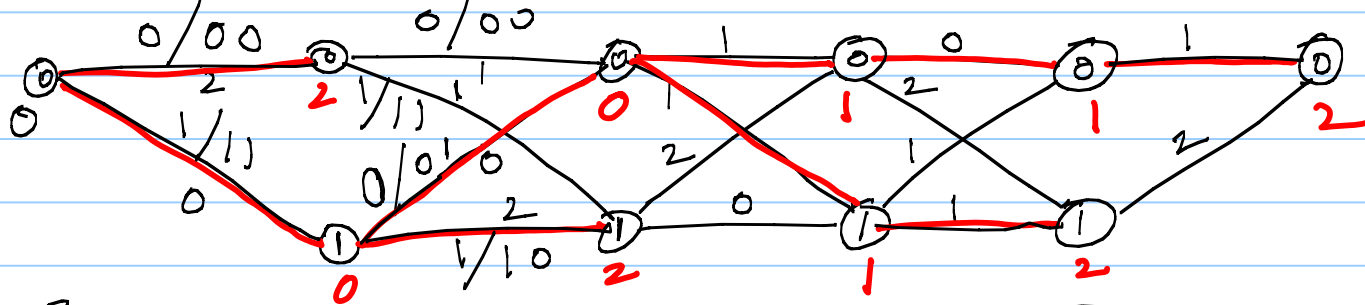
$$u(D) = \frac{1}{1+D}$$

$$v(D) = \left[\frac{1+D^3}{1+D} \quad \frac{1+D+D^2+D^3}{1+D} \right]$$

$$= \left[1+D+D^2 \quad 1+D^2 \right]$$

19) $G(D) = \left[1 \quad 1+D \right]$

$$\underline{r} = \left[\begin{array}{cccccc} 11 & 01 & 10 & 00 & 10 \\ 0/00 & 0/00 & & & & \end{array} \right]$$



$$\underline{u} = \left[1 \right]$$

$$\left[\begin{array}{cccc} 0 & 0 & 0 & 0 \end{array} \right]$$

(2) $(n, 1)$ repetition code

$$C = \left\{ \begin{array}{l} 00 \dots 0, \\ \downarrow \\ +1 +1 \dots +1, -1 -1 \dots -1 \end{array} \right\}$$

$$\underline{r} = [r_1, r_2, \dots, r_n]$$

$$\hat{c} = \arg \max_{\underline{s}} \underline{r} \cdot \underline{s}$$

$$\hat{c} = \begin{cases} 00 \dots 0 & \text{if } \sum r_i > 0 \\ 11 \dots 1 & \text{else} \end{cases}$$

bitwise-MAP:

$$l_i = e^{\frac{2r_i}{\sigma^2}}$$

$$L_1 = l_1 l_2 l_3$$

$$= e^{\sum_{i=1}^3 r_i / \sigma^2}$$

$\hat{C} =$ same as ML

$$P_r(\text{error}) = Q\left(\sqrt{\frac{2E_b}{N_0}}\right) \rightarrow \text{Same as uncoded.}$$

3. Consider a code with parity-check matrix

$$H = \begin{bmatrix} 1 & 0 & 0 & 1 & 1 \\ 0 & 1 & 0 & 0 & 1 \\ 0 & 0 & 1 & 1 & 0 \end{bmatrix}.$$

- (a) Design soft ML and bitwise-MAP decoders over an AWGN channel under BPSK modulation.
- (b) Design a hard-decision syndrome decoder after converting the above channel to a BSC.
- (c) Is there a received word for which the above three decoders provide three different outputs?

$$C = \{00000, 11001, 10110, 01111\}$$

$$\underline{r} = [r_1 \ r_2 \ r_3 \ r_4 \ r_5]$$

$$\text{Correlations: } a_1 = r_1 + r_2 + r_3 + r_4 + r_5$$

$$a_2 = -r_1 - r_2 + r_3 + r_4 - r_5$$

$$a_3 = -r_1 + r_2 - r_3 - r_4 + r_5$$

$$a_4 = r_1 - r_2 - r_3 - r_4 - r_5$$

$$C = \{00000, 11001, 10110, 01111\}$$

$$L_4 = \frac{l_1 l_2 l_3 l_4 l_5 + l_3 l_4}{l_2 l_5 + l_1}$$

$$L_5 = \frac{l_1 l_2 l_3 l_4 l_5 + l_2 l_5}{l_3 l_4 + l_1}$$

(1) RS code $\binom{2^m-1}{(n)}, \binom{2^m-1-2t}{(k)}, \binom{2t+1}{(d)}$

(b) $p_s = 1 - (1-p)^m \rightarrow \text{as } p \rightarrow 0$
 $p_s \approx mp$

$$\begin{aligned} \Pr(\text{block error}) &= \Pr(> t \text{ symbol errors in } n \text{ symbols}) \\ &= \sum_{j=t+1}^n \binom{n}{j} p_s^j (1-p_s)^{n-j} \end{aligned}$$

BCH code $\binom{2^m-1}{(n)}, \binom{2^m-1-mt}{(k)}, \binom{2t+1}{(d)}$

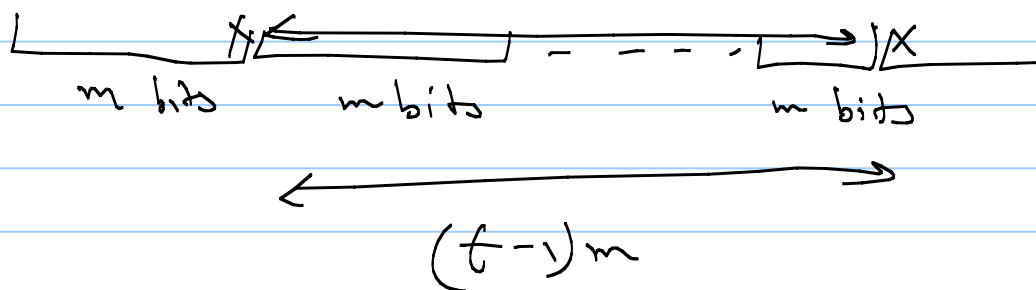
$$\Pr(\text{block error}) = \sum_{j=t+1}^n \binom{n}{j} p^j (1-p)^{n-j}$$

13. Let C be a t -error-correcting RS code of length $n = 2^m - 1$ over $\text{GF}(2^m)$.

(a) Determine the exact burst-error-correcting capability of C in bits.

(b) Let M codewords of C be symbol-interleaved by a row-column interleaver. Determine the burst-error-correcting capability after interleaving.

(a)
$$\underline{\underline{(t-1)m+1}}$$



(b)
$$(t-1)Mm + M$$

$tMm - m + 1 \rightarrow \checkmark$